



# Commercial and Industrial Mobile DDR 512Mb SDRAM

## Features

- **JEDEC LPDDR Compliant**
  - Low Power Consumption
  - 2n Prefetch Architecture
  - Differential clock inputs (CK and  $\overline{\text{CK}}$ )
  - Double-data rate on DQs, DQS and DM
  - Commands entered on each positive CK edge
  - DQS edge-aligned with data for READs; center-aligned with data for WRITEs
  - Status Register Read (SRR)
- **Signal Integrity**
  - Configurable DS for system compatibility
- **Data Integrity**
  - DRAM built-in Temperature Sensor for Temperature Compensated Self Refresh (TCSR)
  - Auto Refresh and Self Refresh Modes
- **Power Saving Mode**
  - Deep Power Down Mode (DPD)
  - Partial Array Self Refresh (PASR)
  - Clock Stop capability during idle period
- **LVC MOS Interface and Power Supply**
  - VDD/VDDQ=1.70 to 1.95V

## Options

- **Speed Grade (CL-TRCD-TRP) <sup>1</sup>**
  - 333 Mbps / 3-3-3
  - 400 Mbps / 3-3-3
- **Temperature Range ( $T_c$ )**
  - Commercial Grade = -25°C ~85°C
  - Industrial Grade = -40°C ~85°C

## Programmable Functions

- **CAS Latency (2, 3)**
- **Burst Length (2, 4, 8, 16)**
- **Burst Type (Sequential, Interleaved)**
- **Driver Strength (full, 1/2, 3/4, 1/4)**

## Packages / Density Information

### Lead-free RoHS compliance and Halogen-free

512Mb (Org / Package)		Length x Width (mm)	Ball pitch (mm)
32Mx16	60-ball VFBGA	8.00 x 9.00	0.80
16Mx32	90-ball VFBGA	8.00 x 13.00	0.80

### Density and Addressing

Item	512Mb		
	Standard		Reduced Page Size
Addressing Organization	32M x 16	16M x 32	16M x 32
Number of banks	4	4	4
Bank Address	BA0,BA1	BA0,BA1	BA0,BA1
Auto precharge	A10/AP	A10/AP	A10/AP
Row Address	A0-A12	A0-A12	A0-A13
Column Address	A0-A9	A0-A8	A0-A7
tRFC(ns) <sup>2</sup>	110	110	110
tREFI (μs) <sup>3</sup>	7.8	7.8	7.8

Notes:

1. The timing specification of high speed bin is backward compatible with low speed bin.
2. Violating tRFC specification will induce malfunction.
3. tREFI values for all bank refresh is within temperature specification(<= 85°C).

## Descriptions

The 512Mb Mobile LPDDR SDRAM is a high-speed CMOS, dynamic random-access memory containing 536,870,912 bits. It is internally configured as a quad-bank DRAM.

The 512Mb chip is organized as 8Mbit x 4 banks x 16 I/O or 4Mbit x 4 banks x 32 I/O device. Each of the x16's 134,217,728-bit banks is organized as 8,192 rows by 1,024 columns by 16 bits. Each of the x32's 134,217,728-bit banks is organized as 8,192 rows by 512 columns by 32 bits. In the reduced page-size option, each of the x32's 134,217,728-bit banks are organized as 16,384 rows by 256 columns by 32 bits. To achieve high-speed operation, our LPDDR SDRAM uses the double data rate architecture and adopt 2n-prefetch interface designed to transfer two data per clock cycle at the I/O pins.

The chip is designed to comply with all key Mobile Double-Data-Rate SDRAM key features. All of the control and address inputs are synchronized with a pair of externally supplied differential clocks, and latched at the cross point of differential clocks (CK rising and  $\overline{CK}$  falling). The input data is registered at both edges of DQS, and the output data is referenced to both edges of DQS, as well as to both edges of CK. DQS is a bidirectional data strobe signal, transmitted by the LPDDR SDRAM during READs (edge-aligned with data), and by the memory controller during WRITEs (center-aligned with data).

LPDDR SDRAM, Read and Write access are burst oriented. The address bits registered coincident with the ACTIVE command to select the row in the specific bank. And then the address bits registered with the READ or WRITE command to select the starting column location in the bank for the burst access. The burst length can be programmed as 2, 4, 8 or 16. An Auto Precharge function may be enabled to provide a self-timed row precharge that is initiated at the end of burst access.

LPDDR SDRAM with Auto Refresh mode, and the Power-down mode for power saving. And the Deep Power Down Mode can achieve the maximum power reduction by removing the memory array power within Low Power DDR SDRAM. With this feature, the system can cut off almost all DRAM power without adding the cost of a power switch and giving up month-board power-line layout flexibility. Self Refresh mode with Temperature Compensated Self Refresh (TCSR) and Partial Array Self Refresh (PASR) options, which allow users to achieve additional power saving. The TCSR and PASR options can be programmed via the extended mode register. The two features may be combined to achieve even greater power saving. The DLL that is typically used on standard DDR devices is not necessary on the Mobile DDR SDRAM. It has been omitted to save power.

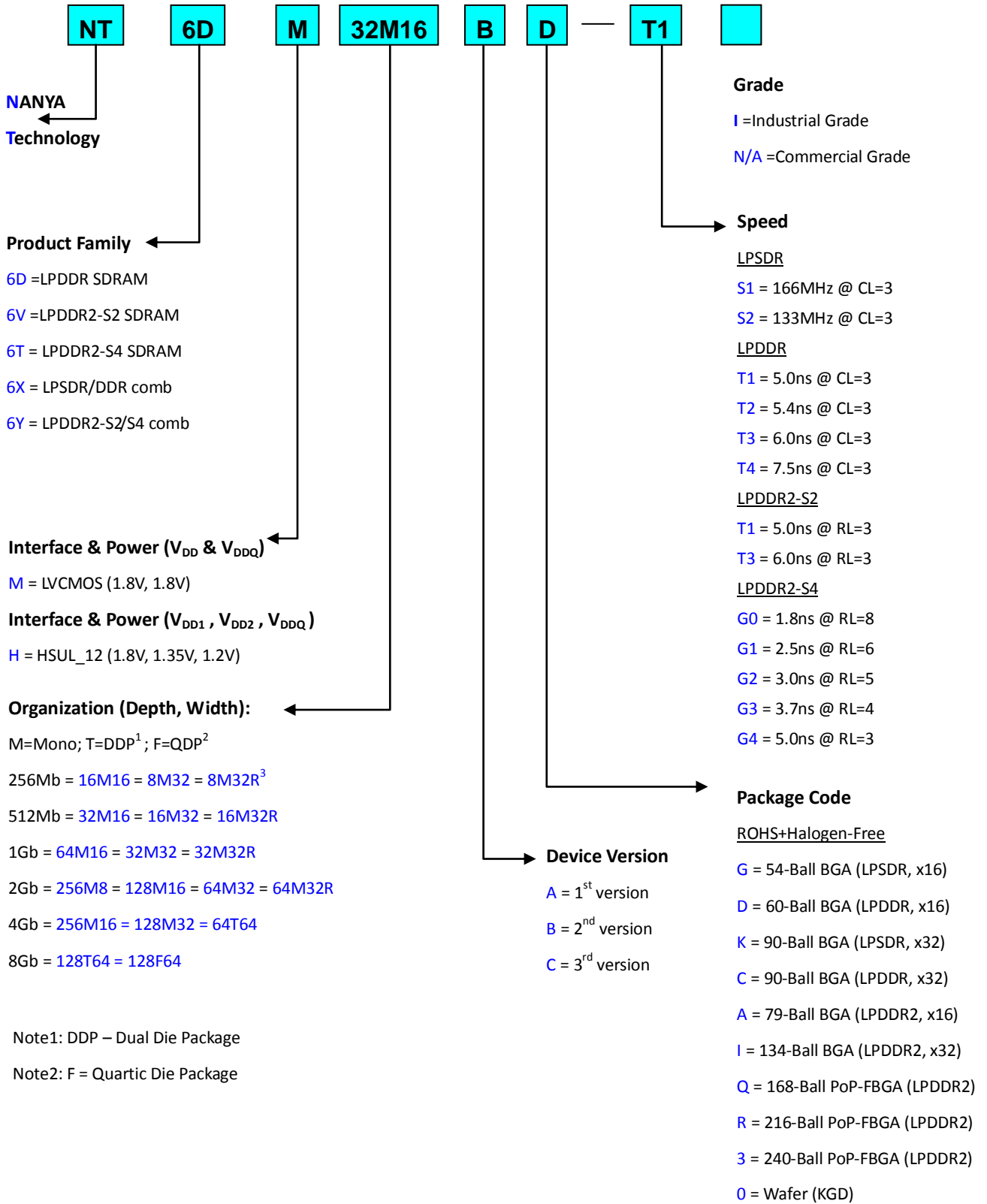
All inputs are LVCMOS compatible. Devices will have a  $V_{DD}$  and  $V_{DDQ}$  supply of 1.8V (nominal).

## Ordering Information

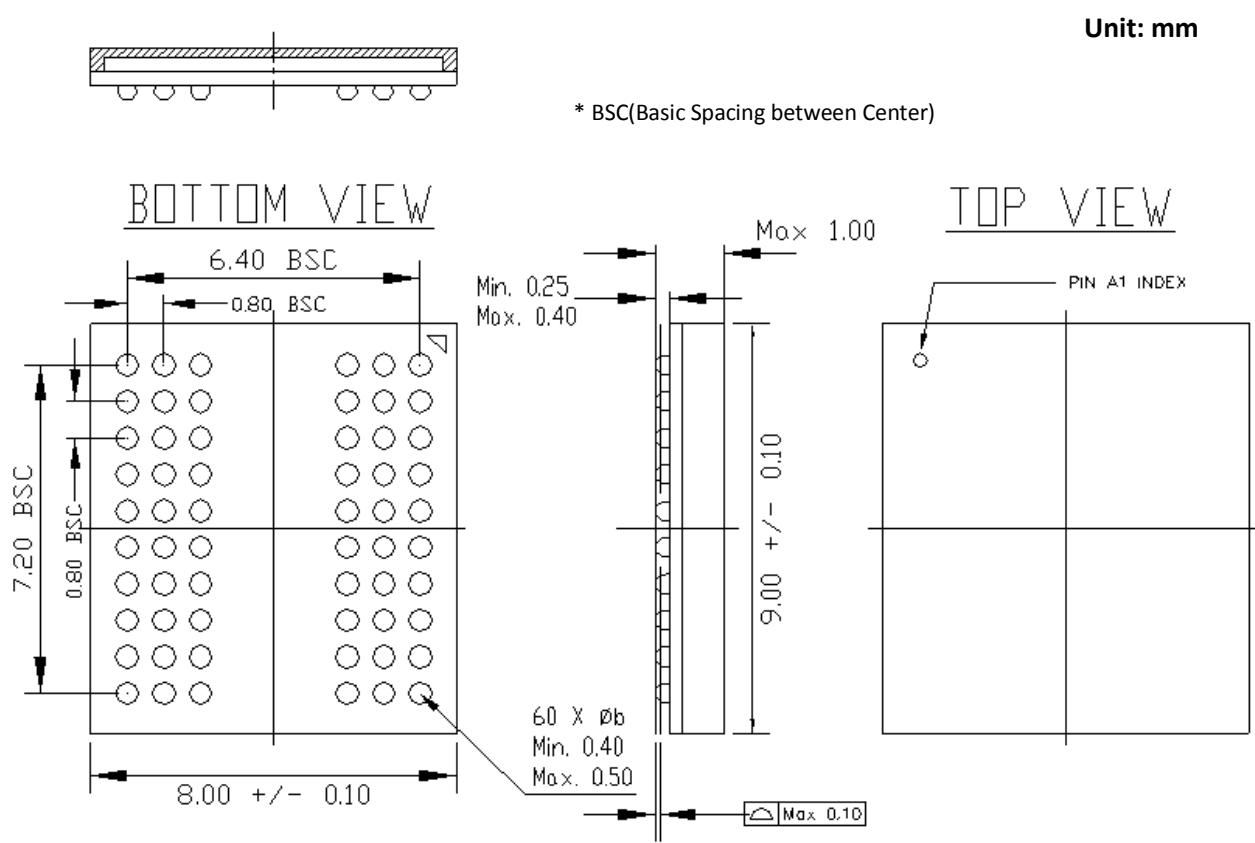
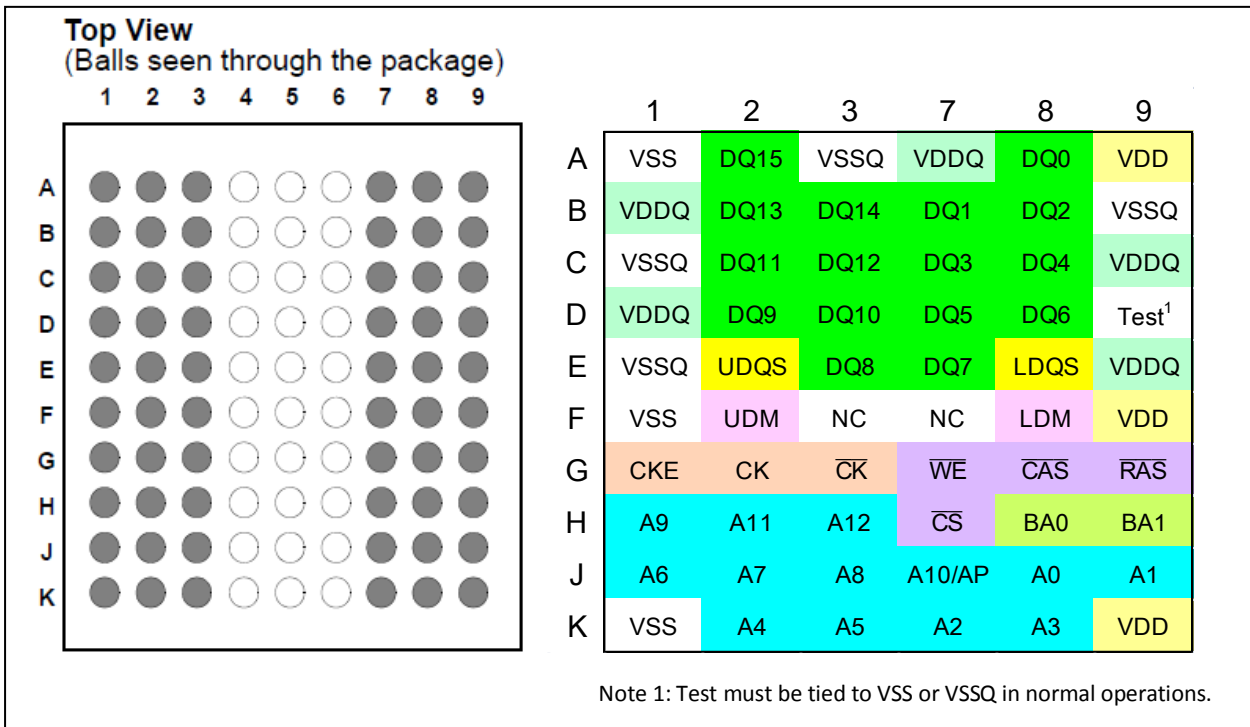
Lead-free RoHS compliance and Halogen-free

Organization	Part Number	Package	Speed			
			TCK (ns)	Clock (MHz)	Data Rate (Mb/s/pin)	CL
<b>Commercial Grade</b>						
32MX16	NT6DM32M16BD-T1	60 ball	5.0	200	400	3
	NT6DM32M16BD-T3		6.0	166	333	3
16MX32	NT6DM16M32BC-T1	90 ball	5.0	200	400	3
	NT6DM16M32BC-T3		6.0	166	333	3
<b>Industrial Grade</b>						
32MX16	NT6DM32M16BD-T1I	60 ball	5.0	200	400	3
	NT6DM32M16BD-T3I		6.0	166	333	3
16MX32	NT6DM16M32BC-T1I	90 ball	5.0	200	400	3
	NT6DM16M32BC-T3I		6.0	166	333	3

## NANYA Mobile Component/Wafer Part Numbering Guide



### Ball Assignments and Package Outline Drawing LPDDR SDRAM X16 in VFBGA-60 (8mm x 9mm)

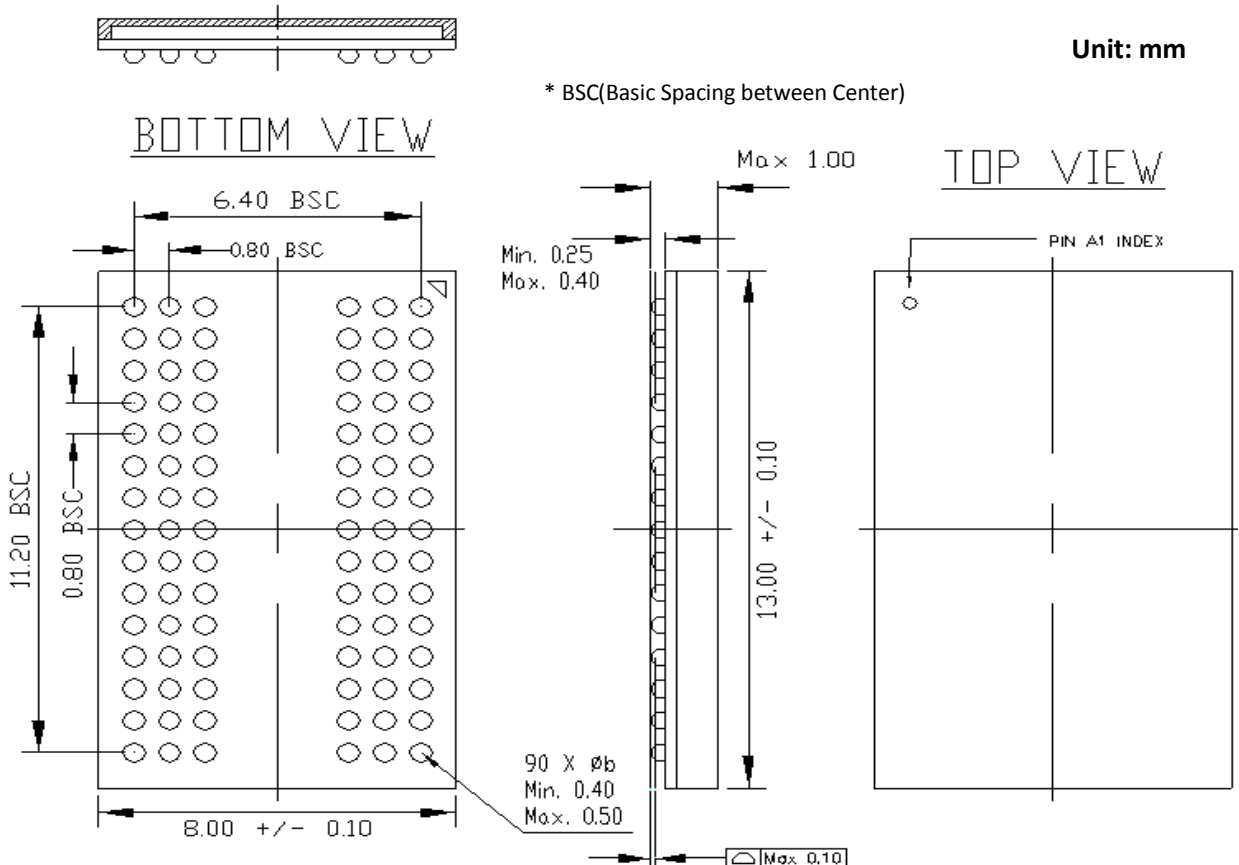


### Ball Assignments and Package Outline Drawing LPDDR SDRAM X32 in VFBGA-90 (8mm x 13mm)

**Top View**  
(Balls seen through the package)

	1	2	3	7	8	9
A	VSS	DQ31	VSSQ	VDDQ	DQ16	VDD
B	VDDQ	DQ29	DQ30	DQ17	DQ18	VSSQ
C	VSSQ	DQ27	DQ28	DQ19	DQ20	VDDQ
D	VDDQ	DQ25	DQ26	DQ21	DQ22	Test <sup>2</sup>
E	VSSQ	DQS3	DQ24	DQ23	DQS2	VDDQ
F	VDD	DM3	NC	A13 <sup>1</sup>	DM2	VSS
G	CKE	CK	CK	WE	CAS	RAS
H	A9	A11	A12	CS	BA0	BA1
J	A6	A7	A8	A10/AP	A0	A1
K	A4	DM1	A5	A2	DM0	A3
L	VSSQ	DQS1	DQ8	DQ7	DQS0	VDDQ
M	VDDQ	DQ9	DQ10	DQ5	DQ6	VSSQ
N	VSSQ	DQ11	DQ12	DQ3	DQ4	VDDQ
P	VDDQ	DQ13	DQ14	DQ1	DQ2	VSSQ
R	VSS	DQ15	VSSQ	VDDQ	DQ0	VDD

Note 1: A13 is only available for reduced page-size configuration.  
Note 2: Test must be tied to VSS or VSSQ in normal operations.



### Ball Descriptions

Symbol <sup>1</sup>	Type	Function
CK, $\overline{CK}$	Input	<b>Clock:</b> CK and $\overline{CK}$ are differential clock inputs. All address and control input signals are sampled on the crossing of the positive edge of CK and negative edge of $\overline{CK}$ . Input and output data is referenced to the crossing of CK and $\overline{CK}$ (both directions of crossing). Internal clock signals are derived from CK, $\overline{CK}$ .
CKE	Input	<b>Clock Enable:</b> CKE HIGH activates, and CKE LOW deactivates internal clock signals, and device input buffers and output drivers. Taking CKE LOW provides PRECHARGE POWER-DOWN and SELF REFRESH operation (all banks idle), or ACTIVE POWERDOWN (row ACTIVE in any bank). CKE is synchronous for all functions except for SELF REFRESH EXIT, which is achieved asynchronously. Input buffers, excluding CK, $\overline{CK}$ and CKE, are disabled during power-down and self refresh mode which are contrived for low standby power consumption.
$\overline{CS}$	Input	<b>Chip Select:</b> $\overline{CS}$ enables (registered LOW) and disables (registered HIGH) the command decoder. All commands are masked when $\overline{CS}$ is registered HIGH. $\overline{CS}$ provides for external bank selection on systems with multiple banks. $\overline{CS}$ is considered part of the command code.
$\overline{RAS}$ , $\overline{CAS}$ , $\overline{WE}$	Input	<b>Command Inputs:</b> $\overline{RAS}$ , $\overline{CAS}$ and $\overline{WE}$ (along with $\overline{CS}$ ) define the command being entered.
DM For x16, LDM, UDM For x32, DM0-DM3	Input	<b>Input Data Mask:</b> DM is an input mask signal for write data. Input data is masked when DM is sampled HIGH along with that input data during a WRITE access. DM is sampled on both edges of DQS. Although DM pins are input-only, the DM loading matches the DQ and DQS loading.  For x16 devices, LDM corresponds to the data on DQ0-DQ7, UDM corresponds to the data on DQ8-DQ15.  For x32 devices, DM0 corresponds to the data on DQ0-DQ7, DM1 corresponds to the data on DQ8-DQ15, DM2 corresponds to the data on DQ16-DQ23, and DM3 corresponds to the data on DQ24-DQ31.
DQ For x16: DQ0-DQ15 For x32: DQ0-DQ31	Input/output	<b>Data Bus:</b> Bi-directional Input / Output data bus.
DQS For x16: LDQS, UDQS For x32: DQS0-DQS3	Input/output	<b>Data Strobe:</b> Output with read data, input with write data. Edge-aligned with read data. Centered with write data to capture write data.  For x16 device, LDQS corresponds to the data on DQ0-DQ7, UDQS corresponds to the data on DQ8-DQ15.  For x32 device, DQS0 corresponds to the data on DQ0-DQ7, DQS1 corresponds to the data on DQ8-DQ15, DQS2 corresponds to the data on DQ16-DQ23, and DQS3 corresponds to the data on DQ24-DQ31.
BA0, BA1	Input	<b>Bank Address Inputs:</b> BA0 and BA1 define to which bank an ACTIVE, READ, WRITE, or PRECHARGE command is being applied. BA0 and BA1 also determine which mode register is loaded during a LOAD MODE REGISTER command.

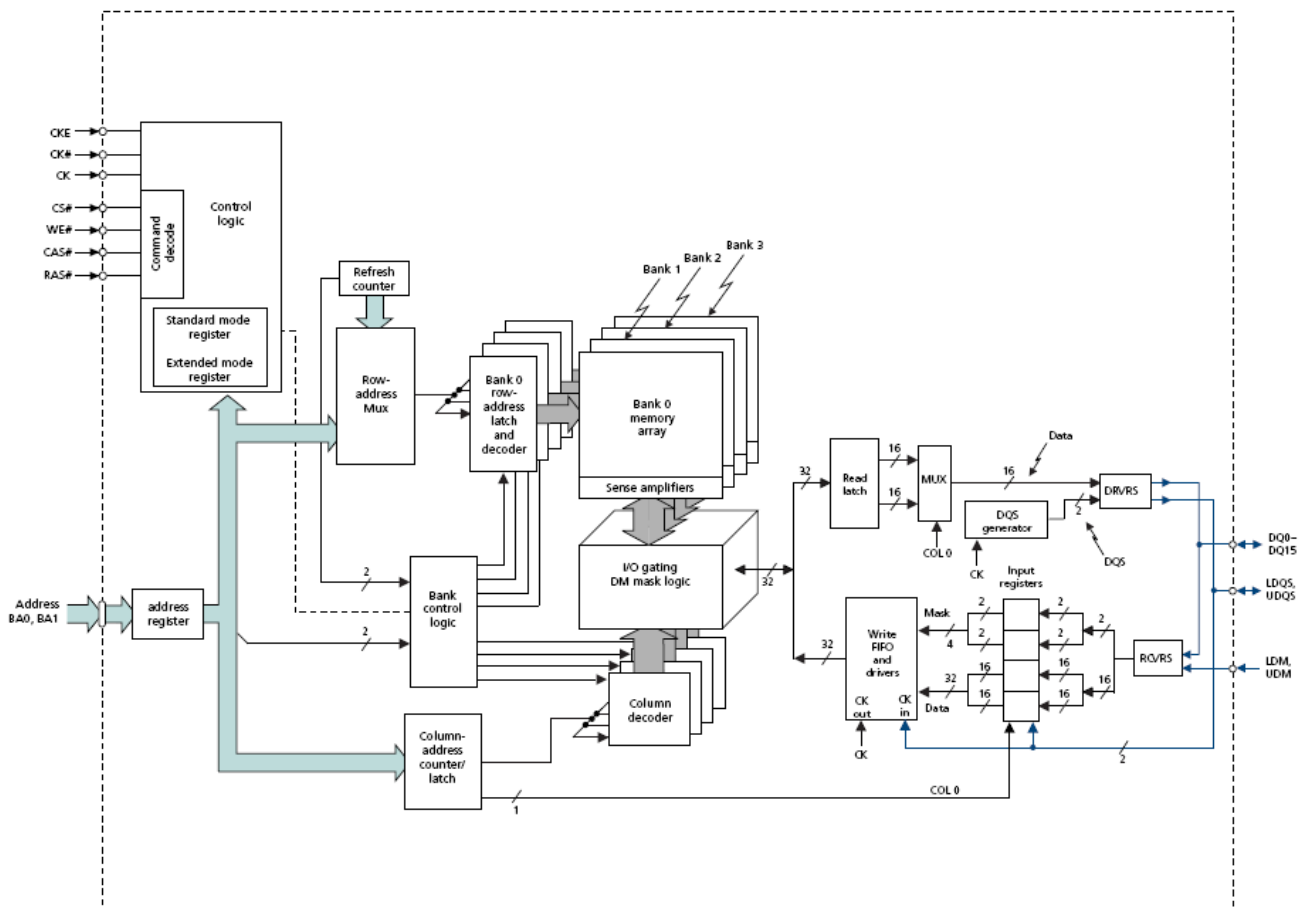
<b>A13 - A0</b>	Input	<b>Address Inputs:</b> provide the row address for ACTIVE commands, and the column address and auto precharge bit(A10) for READ or WRITE commands, to select one location out of the memory array in the respective bank. During a PRECHARGE command, A10 determines whether the PRECHARGE applies to one bank (A10 LOW, bank selected by Bank Address Inputs) or all banks (A10 HIGH). The address inputs also provide the opcode during a MODE REGISTER SET command.
<b>NC</b>	-	<b>No Connect:</b> These pins should be left unconnected.
<b>VDDQ</b>	Supply	<b>DQ Power Supply:</b> Isolated on the die for improved noise immunity.
<b>VSSQ</b>	Supply	<b>DQ Ground:</b> Provide isolated ground to DQs for improved noise immunity.
<b>VDD</b>	Supply	<b>Power Supply</b>
<b>VSS</b>	Supply	<b>Ground</b>

Notes :

- The signal may show up in a different symbol but it indicates to the same thing. e.g., /CK = CK# =  $\overline{\text{CK}}$  = CKb,  
/DQS = DQS# =  $\overline{\text{DQS}}$  = DQsb

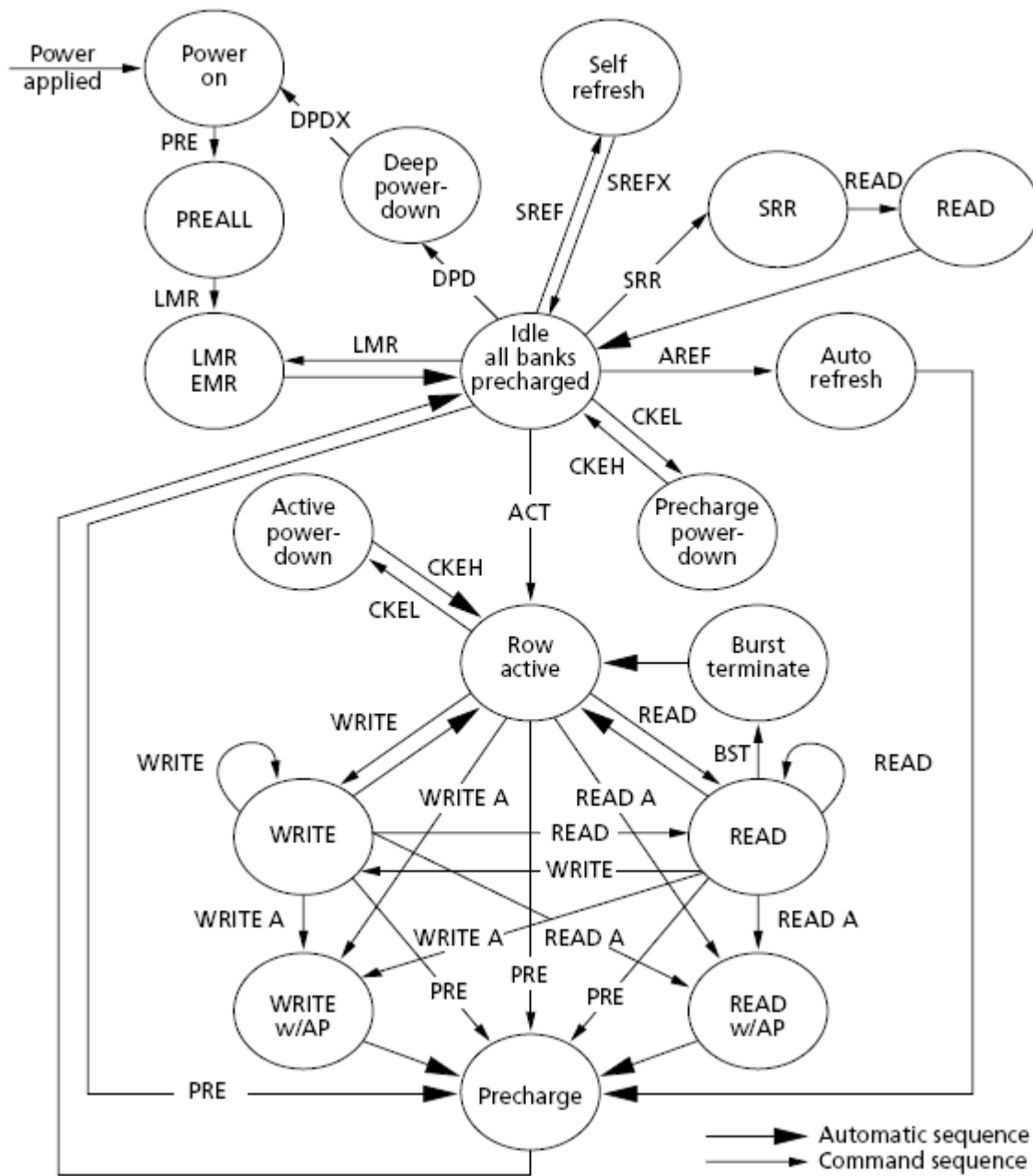


### Functional Block Diagram – LPDDR 32Mx16





### Simplified State Diagram



Abbr.	Function	Abbr.	Function	Abbr.	Function
<b>ACT</b>	Active	<b>LMR</b>	Load mode register	<b>PRE</b>	Precharge
<b>READ</b>	Read (w/o Autoprecharge)	<b>CKEH</b>	Exit power-down	<b>PREALL</b>	Precharge all banks
<b>READ A</b>	Read (w/ Autoprecharge)	<b>CKEL</b>	Enter power-down	<b>AREF</b>	Auto Refresh
<b>WRITE</b>	Write (w/o Autoprecharge)	<b>DPD</b>	Enter Deep Power Down	<b>SREF</b>	Enter self refresh
<b>WRITE A</b>	Write (w/ Autoprecharge)	<b>DPDX</b>	Exit Deep Power Down	<b>SREFX</b>	Exit self refresh
<b>EMR</b>	Load extended mode register	<b>BST</b>	Burst Terminate	<b>SRR</b>	Status Register Read

### Electrical Specifications

#### Absolute Maximum DC Ratings

Symbol	Parameter	Min	Max	Units
$V_{DD} / V_{DDQ}$	$V_{DD} / V_{DDQ}$ supply voltage relative to $V_{SS}$	-1.0	2.4	V
$V_{in}$	Voltage on any pin relative to $V_{SS}$	-0.5	2.4 or ( $V_{DDQ} + 0.3V$ ), Whichever is less	V
Tstg	Storage Temperature (plastic)	-55	+150	°C

Notes:

- Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability.
- Storage Temperature is the case surface temperature on the center/top side of the DRAM.
- $V_{DD}$  and  $V_{DDQ}$  must be within 300mV of each other at all times.  $V_{DDQ}$  must not exceed  $V_{DD}$ .

#### Input / Output Capacitance

Symbol	Parameter	Min	Max	Unit	Notes
CCK	Input capacitance: CK, $\overline{CK}$	1.5	3.0	pF	
CDCK	Input capacitance delta: CK, $\overline{CK}$	-	0.25	pF	2
CI	Input capacitance, all other input-only pins	1.5	3.0	pF	
CDI	Input capacitance delta, all other input-only pins	-	0.5	pF	2
CIO	Input/output capacitance, DQ, DM, DQS	3.0	5.0	pF	
CDIO	Input/output capacitance delta, DQ, DM, DQS	-	0.5	pF	3

Notes:

- These values are guaranteed by design and are tested on a sample base only.
- These capacitance values are for single monolithic devices only. Multiple die packages will have parallel capacitive loads.
- Input capacitance is measured according to JEP147 procedure for measuring capacitance using a vector network analyzer.  $V_{DD}$ ,  $V_{DDQ}$  are applied and all other pins (except the pin under test) floating. DQs should be in high impedance state. This may be achieved by pulling CKE to low level.
- Although DM is an input-only pin, the input capacitance of this pin must model the input capacitance of the DQ and DQS pins. This is required to match signal propagation times of DQ, DQS and DM in the system.

## AC/DC Electrical Characteristics and Operating Conditions

Apply Note 1-3 to whole the table.

Symbol	Parameter	Min	Max	Unit	Notes
V <sub>DD</sub>	Supply voltage	1.70	1.95	V	-
V <sub>DDQ</sub>	I/O Supply voltage	1.70	1.95	V	-
<b>Address and Command inputs</b>					
V <sub>IH</sub>	Input voltage high	0.8 x V <sub>DDQ</sub>	V <sub>DDQ</sub> + 0.3	V	-
V <sub>IL</sub>	Input voltage low	-0.3	0.2 x V <sub>DDQ</sub>	V	-
<b>Clock inputs (CK, <math>\overline{CK}</math>)</b>					
V <sub>IN</sub>	DC input voltage	-0.3	V <sub>DDQ</sub> + 0.3	V	-
V <sub>ID(DC)</sub>	DC input differential voltage	0.4 x V <sub>DDQ</sub>	V <sub>DDQ</sub> + 0.6	V	2
V <sub>ID(AC)</sub>	AC Input Differential Voltage	0.6 x V <sub>DDQ</sub>	V <sub>DDQ</sub> + 0.6	V	2
V <sub>IX</sub>	AC Differential Crosspoint Voltage	0.4 x V <sub>DDQ</sub>	0.6 x V <sub>DDQ</sub>	V	3
<b>Data inputs</b>					
V <sub>IH(DC)</sub>	DC input high voltage	0.7 x V <sub>DDQ</sub>	V <sub>DDQ</sub> + 0.3	V	-
V <sub>IL(DC)</sub>	DC input low voltage	-0.3	0.3 x V <sub>DDQ</sub>	V	-
V <sub>IH(AC)</sub>	AC input high voltage	0.8 x V <sub>DDQ</sub>	V <sub>DDQ</sub> + 0.3	V	-
V <sub>IL(AC)</sub>	AC input low voltage	-0.3	0.2 x V <sub>DDQ</sub>	V	-
<b>Data outputs</b>					
V <sub>OH</sub>	DC output high voltage: Logic 1 (I <sub>OH</sub> = -0.1mA)	0.9 x V <sub>DDQ</sub>	-	V	-
V <sub>OL</sub>	DC output low voltage: Logic 0 (I <sub>OL</sub> = -0.1mA)	-	0.1 x V <sub>DDQ</sub>	V	-
<b>Leakage current</b>					
I <sub>I</sub>	Input leakage current Any input $0 \leq V_{IN} \leq V_{DD}$ , All other pins not under test = 0V	-1	1	uA	
I <sub>OZ</sub>	Output leakage current DQs are disabled; $0 \leq V_{OUT} \leq V_{DDQ}$	-5	5	uA	
Notes:					
1.All voltages referenced to VSS and VSSQ must be same potential.					
2.VID(DC) and VID(AC) are the magnitude of the difference between the input level on CK and the input level on $\overline{CK}$ .					
3.The value of VIX is expected to be 0.5 * VDDQ and must track variations in the DC level of the same.					

**IDD Specifications and Measurement Conditions (32Mx16)**

Notes 1 – 5 apply to all the parameters/conditions in this table

Symbol	Parameter/Condition	T1(-5)	T3(-6)	Unit	Notes	
		LPDDR400	LPDDR333			
IDD0	<b>Operating one bank active-precharge current:</b> tRC = tRCmin; tCK = tCKmin; CKE is HIGH; $\overline{CS}$ is HIGH between valid commands; address inputs are SWITCHING; data bus inputs are STABLE	70	60	mA	6	
IDD2P	<b>Precharge power-down standby current:</b> all banks idle, CKE is LOW; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	300		uA	7,8	
IDD2PS	<b>Precharge power-down standby current with clock stopped:</b> all banks idle, CKE is LOW; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	300		uA	7	
IDD2N	<b>Precharge non power-down standby current:</b> all banks idle, CKE is HIGH; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	15	15	mA	9	
IDD2NS	<b>Precharge non power-down standby current with clock stopped:</b> all banks idle, CKE is HIGH; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	10	10	mA	9	
IDD3P	<b>Active power-down standby current:</b> one bank active, CKE is LOW; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	5		mA	8	
IDD3PS	<b>Active power-down standby current with clock stopped:</b> one bank active, CKE is LOW; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	4		mA		
IDD3N	<b>Active non power-down standby current:</b> one bank active, CKE is HIGH; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	15	15	mA	6	
IDD3NS	<b>Active non power-down standby current with clock stopped:</b> one bank active, CKE is HIGH; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	10	10	mA	6	
IDD4R	<b>Operating burst read current:</b> one bank active; BL=4; CL=3; tCK = tCKmin; continuous read bursts; IOUT = 0 mA address inputs are SWITCHING; 50% data change each burst transfer	115	105	mA	6	
IDD4W	<b>Operating burst write current:</b> one bank active; BL=4; tCK = tCKmin; continuous write bursts; address inputs are SWITCHING; 50% data change each burst transfer	115	105	mA	6	
IDD5	<b>Auto Refresh current:</b> tCK = tCKmin; burst refresh; CKE is HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	tRC = tRFC	95	95	mA	10
IDD5A		tRC = tREFI	3	3	mA	10,11
IDD8	<b>Deep power-down current:</b> Address and control inputs are STABLE; data bus inputs are STABLE	25°C	10		uA	7,13
		85°C	20		uA	7

### IDD Specifications and Measurement Conditions (16MX32)

Notes 1 – 5 apply to all the parameters/conditions in this table

Symbol	Parameter/Condition	T1(-5)	T3(-6)	Unit	Notes	
		LPDDR400	LPDDR333			
IDD0	<b>Operating one bank active-precharge current:</b> tRC = tRCmin; tCK = tCKmin; CKE is HIGH; $\overline{CS}$ is HIGH between valid commands; address inputs are SWITCHING; data bus inputs are STABLE	70	60	mA	6	
IDD2P	<b>Precharge power-down standby current:</b> all banks idle, CKE is LOW; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	300		uA	7,8	
IDD2PS	<b>Precharge power-down standby current with clock stopped:</b> all banks idle, CKE is LOW; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	300		uA	7	
IDD2N	<b>Precharge non power-down standby current:</b> all banks idle, CKE is HIGH; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	15	15	mA	9	
IDD2NS	<b>Precharge non power-down standby current with clock stopped:</b> all banks idle, CKE is HIGH; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	10	10	mA	9	
IDD3P	<b>Active power-down standby current:</b> one bank active, CKE is LOW; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	5		mA	8	
IDD3PS	<b>Active power-down standby current with clock stopped:</b> one bank active, CKE is LOW; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	4		mA		
IDD3N	<b>Active non power-down standby current:</b> one bank active, CKE is HIGH; $\overline{CS}$ is HIGH, tCK = tCKmin; address and control inputs are SWITCHING; data bus inputs are STABLE	15	15	mA	6	
IDD3NS	<b>Active non power-down standby current with clock stopped:</b> one bank active, CKE is HIGH; $\overline{CS}$ is HIGH, CK = LOW, CK = HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	10	10	mA	6	
IDD4R	<b>Operating burst read current:</b> one bank active; BL=4; CL=3; tCK = tCKmin; continuous read bursts; IOUT = 0 mA address inputs are SWITCHING; 50% data change each burst transfer	115	105	mA	6	
IDD4W	<b>Operating burst write current:</b> one bank active; BL=4; tCK = tCKmin; continuous write bursts; address inputs are SWITCHING; 50% data change each burst transfer	115	105	mA	6	
IDD5	<b>Auto Refresh current:</b> tCK = tCKmin; burst refresh; CKE is HIGH; address and control inputs are SWITCHING; data bus inputs are STABLE	tRC = tRFC	95	95	mA	10
IDD5A		tRC = tREFI	3	3	mA	10,11
IDD8	<b>Deep power-down current:</b> Address and control inputs are STABLE; data bus inputs are STABLE	25°C	10		uA	7,13
		85°C	20		uA	7,13

**IDD Specifications and Measurement Conditions**

**IDD6 Self-refresh and Partial Array Refresh current**

Notes 1 – 5, 7, and 12 apply to all the parameters/conditions in this table

Symbol	Parameter/Condition	Temperature	PASR	Typical	Unit
IDD6	<b>Self refresh current:</b> CKE=LOW; $t^{CK} = t^{CK(min)}$ ; Address and control inputs are stable; Data bus inputs are stable.	85°C	Full Array	600	uA
			1/2 Array	480	uA
			1/4 Array	420	uA
			1/8 Array	420	uA
			1/16 Array	400	uA
		45°C	Full Array	300	uA
			1/2 Array	260	uA
			1/4 Array	250	uA
			1/8 Array	250	uA
			1/16 Array	250	uA

**IDD Notes:**

- All voltages referenced to  $V_{SS}$ .
- Tests for  $I_{DD}$  may be conducted at nominal supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage and temperature range specified.
- Timing and  $I_{DD}$  tests may use a  $V_{IL}$ -to- $V_{IH}$  swing of up to 1.5V in the test environment, but input timing is still referenced to  $V_{DDQ}/2$  (or, to the crossing point for CK and  $\overline{CK}$ ). The output timing reference voltage level is  $V_{DDQ}/2$ .
- $I_{DD}$  is dependent on output loading and cycle rates. Specified values are obtained with minimum cycle time with the outputs open.
- $I_{DD}$  specifications are tested after the device is properly initialized, and are averaged at the defined cycle rate.
- MIN ( $t^{RC}$  or  $t^{RFC}$ ) for IDD measurements is the smallest multiple of  $t^{CK}$  that meets the minimum absolute value for the respective parameter.  $t^{RAS}$  (MAX) for IDD measurements is the largest multiple of  $t^{CK}$  that meets the maximum absolute value for  $t^{RAS}$ .
- Measurement is taken 500ms after entering into this operating mode to provide settling time for the tester.
- $V_{DD}$  must not vary more than 4 % if CKE is not active while any bank is active.
- IDD2N specifies DQ, DQS, and DM to be driven to a valid HIGH or LOW logic level.
- CKE must be active (HIGH) during the entire time a REFRESH command is executed. From the time the AUTO REFRESH command is registered, CKE must be active at each rising clock edge until  $t^{RFC}$  later.
- This limit is a nominal value and does not result in a fail. CKE is HIGH during REFRESH command period ( $t^{RFC}$  [MIN]) else CKE is LOW (for example, during standby).
- Values for IDD6 85°C are guaranteed for the entire temperature range.
- IDD8 are typical values.



### AC Timings

Note 1-9 apply to all of the parameters in this table

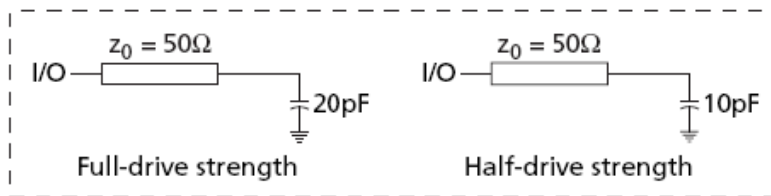
Symbol	Parameter		T3 (-6): LPDDR333		T1 (-5): LPDDR400		Unit	Notes
			Min	Max	Min	Max		
tAC	Access window of DQs from CK, $\overline{CK}$		2.0	5.5	2.0	5.0	ns	
tDQSK	Access window of DQS from CK, $\overline{CK}$		2.0	5.5	2.0	5.0	ns	
tCH	CK high-level width		0.45	0.55	0.45	0.55	tCK	
tCL	CK low-level width		0.45	0.55	0.45	0.55	tCK	
tHP	Half-clock period		min(tCH,tCL)	-	min(tCH,tCL)	-	ns	10, 11
tCK	Clock cycle time	CL=3	6.0	100	5.0	100	ns	12
		CL=2(optional)	12	100	12	100		
tDSf	DQ and DM input setup time relative to DQS (fast slew rate)		0.6	-	0.48	-	ns	13,14,15
tDSs	DQ and DM input setup time relative to DQS (slow slew rate)		0.7	-	0.58	-	ns	13,14,16
tDHf	DQ and DM input hold time relative to DQS (fast slew rate)		0.6	-	0.48	-	ns	13,14,15
tDHs	DQ and DM input hold time relative to DQS (slow slew rate)		0.7	-	0.58	-	ns	13,14,16
tDIPW	DQ and DM input pulse width		2.1	-	1.8	-	ns	17
tISf	Address and Control input setup time (fast slew rate)		1.1	-	0.9	-	ns	15,18
tISs	Address and Control input setup time (slow slew rate)		1.3	-	1.1	-	ns	16,18
tIHf	Address and Control input hold time (fast slew rate)		1.1	-	0.9	-	ns	15,18
tIHs	Address and Control input hold time (slow slew rate)		1.3	-	1.1	-	ns	16,18
tIPW	Address and Control input pulse width		2.7	-	2.3	-	ns	17
tLZ	Data-out Low-z window from CK, $\overline{CK}$		1.0	-	1.0	-	ns	19
tHZ	Data-out high-z window from CK, $\overline{CK}$		-	5.5	-	5.0	ns	19
tDQSQ	DQS-DQ skew, DQS to last DQ valid, per group, per access		-	0.5	-	0.4	ns	20
tQH	DQ-DQS hold, DQS to first DQ to go non-valid, per access		tHP - tQHS	-	tHP - tQHS	-	ns	11
tQHS	Data Hold Skew Factor		-	0.65	-	0.5	ns	11

tDQSS	WRITE command to first DQS latching transition	0.75	1.25	0.75	1.25	tCK		
tDQSH	DQS input high pulse width	0.4	-	0.4	-	tCK		
tDQSL	DQS input low pulse width	0.4	-	0.4	-	tCK		
tDSS	DQS falling edge to CK setup time	0.2	-	0.2	-	tCK		
tDSH	DQS falling edge hold time from CK	0.2	-	0.2	-	tCK		
tMRD	Load MODE Register command cycle time	2	-	2	-	tCK		
tWPRES	DQS write preamble setup time	0	-	0	-	ns	21	
tWPST	DQS write postamble	0.4	0.6	0.4	0.6	tCK	22	
tWPRE	DQS write preamble	0.25	-	0.25	-	tCK		
tRPRE	DQS read preamble	CL=3	0.9	1.1	0.9	1.1	tCK	23
		CL=2(optional)	0.5	1.1	0.5	1.1	tCK	23
tRPST	DQS read postamble	0.4	0.6	0.4	0.6	tCK		
tRAS	ACTIVE to PRECHARGE command	42	70,000	40	70,000	ns		
tRC	ACTIVE to ACTIVE / ACTIVE to AUTO REFRESH command period	60	-	55	-	ns		
tRFC	Auto Refresh command period	110	-	110	-	ns		
tRCD	ACTIVE to READ or WRITE delay	18	-	15	-	ns	24	
tRP	PRECHARGE command period	18	-	15	-	ns	24	
tRRD	ACTIVE <i>bank-a</i> to ACTIVE <i>bank-b</i> command	12	-	10	-	ns		
tWR	Write recovery time	15	-	15	-	ns		
tDAL	Auto precharge write recovery + precharge time	-	-	-	-	-	26	
tWTR	Internal WRITE to READ command delay	2	-	2	-	tCK		
tXSR	Exit SELF REFRESH to first valid command	200	-	200	-	ns	27	
tXP	Exit power-down mode to first valid command	25	-	25	-	ns	28	
tCKE	CKE min. pulse width (high and low pulse width)	2	-	2	-	tCK		
tREF	Refresh period	-	64	-	64	ms		
tREFI	Average periodic refresh interval	-	7.8	-	7.8	us	29, 30	
tSRR	SRR-to-READ	2	-	2	-	tCK		
tSRC	Read of SRR to next valid command	CL+1	-	CL+1	-	tCK		

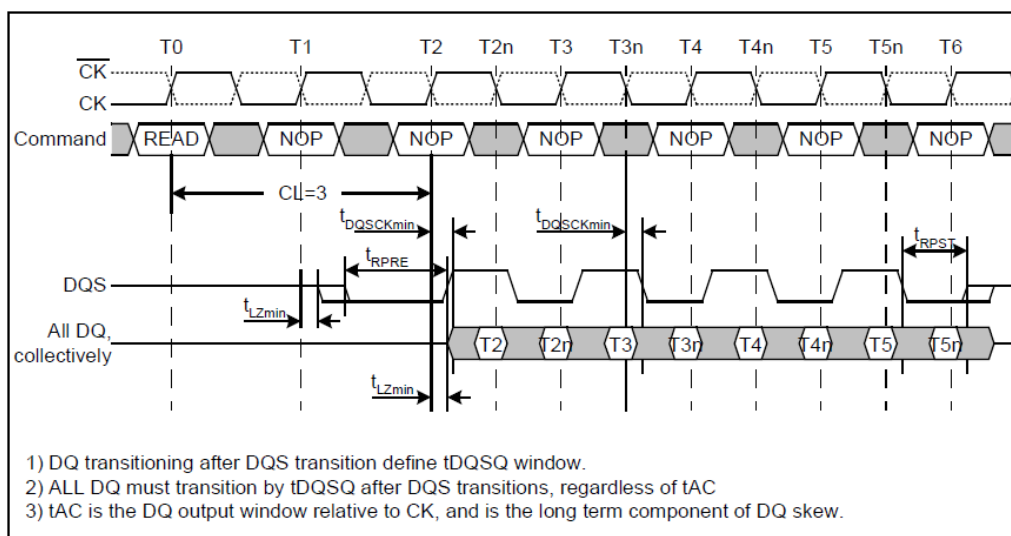
tTQ	Internal temperature sensor valid temperature output enable	2	-	2	-	ms	31
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Notes:

1. All voltages referenced to Vss.
2. All parameters assume proper device initialization.
3. Tests for AC timing, and electrical AC and DC characteristics may be conducted at nominal supply voltage levels, but the related specifications and device operation are guaranteed for the full voltage range specified.
4. The circuit shown below represents the timing reference load used in defining the relevant timing parameters of the device. It is not intended to be either a precise representation of the typical system environment or a depiction of the actual load presented by a production tester. System designers will use IBIS or other simulation tools to correlate the timing reference load to system environment. Specifications are correlated to production test conditions (generally a coaxial transmission line terminated at the tester electronics). For the half-strength driver with a nominal 10pF load, parameters  $t_{AC}$  and  $t_{QH}$  are expected to be in the same range. However, these parameters are not subject to production test but are estimated by design/characterization. Use of IBIS or other simulation tools for system design validation is suggested.



5. The CK,  $\overline{CK}$  input reference voltage level (for timing referenced to CK,  $\overline{CK}$ ) is the point at which CK and  $\overline{CK}$  cross; the input reference voltage level for signals other than CK,  $\overline{CK}$  is  $V_{DDQ}/2$ .
6. The timing reference voltage level is  $V_{DDQ}/2$ .
7. AC and DC input and output voltage levels are defined in the section for Electrical Characteristics and AC/DC operating conditions.
8. A CK/ $\overline{CK}$  differential slew rate of 2.0 V/ns is assumed for all parameters.
9. CAS latency definition: with CL = 3 the first data element is valid at  $(2 * t_{CK} + t_{AC})$  after the clock at which the READ command was registered (see figure); with CL = 2 the first data element is valid at  $(t_{CK} + t_{AC})$  after the clock at which the READ command was registered; with CL = 4 the first data element is valid at  $(3 * t_{CK} + t_{AC})$  after the clock at which the READ command was registered.



10. Min (tCL, tCH) refers to the smaller of the actual clock low time and the actual clock high time as provided to the device (i.e. this value can be greater than the minimum specification limits for tCL and tCH)
11. tQH = tHP - tQHS, where tHP = minimum half clock period for any given cycle and is defined by clock high or clock low (tCL, tCH). tQHS accounts for 1) the pulse duration distortion of on-chip clock circuits; and 2) the worst case push-out of DQS on one transition followed by the worst case pull-in of DQ on the next transition, both of which are, separately, due to data pin skew and output pattern effects, and p-channel to n-channel variation of the output drivers.
12. The only time that the clock frequency is allowed to change is during clock stop, power-down or self-refresh modes.
13. The transition time for DQ, DM and DQS inputs is measured between VIL(DC) to VIH(AC) for rising input signals, and VIH(DC) to VIL(AC) for falling input signals.
14. DQS, DM and DQ input slew rate is specified to prevent double clocking of data and preserve setup and hold times. Signal transitions through the DC region must be monotonic.
15. Input slew rate  $\geq 1.0$  V/ns.
16. Input slew rate  $\geq 0.5$  V/ns and  $< 1.0$  V/ns.
17. These parameters guarantee device timing but they are not necessarily tested on each device.
18. The transition time for address and command inputs is measured between VIH and VIL.
19. tHZ and tLZ transitions occur in the same access time windows as valid data transitions. These parameters are not referred to a specific voltage level, but specify when the device is no longer driving (HZ), or begins driving (LZ).
20. tDQSQ consists of data pin skew and output pattern effects, and p-channel to n-channel variation of the output drivers for any given cycle.
21. The specific requirement is that DQS be valid (HIGH, LOW, or some point on a valid transition) on or before the corresponding CK edge. A valid transition is defined as monotonic and meeting the input slew rate specifications of the device. When no writes were previously in progress on the bus, DQS will be transitioning from Hi-Z to logic LOW. If a previous write was in progress, DQS could be HIGH, LOW, or transitioning from HIGH to LOW at this time, depending on tDQSS.
22. The maximum limit for this parameter is not a device limit. The device operates with a greater value for this parameter, but system performance (bus turnaround) will degrade accordingly.
23. A low level on DQS may be maintained during High-Z states (DQS drivers disabled) by adding a weak pull-down element in the system. It is recommended to turn off the weak pull-down element during read and write bursts (DQS drivers enabled).
24. Speed bin (CL - tRCD - tRP) = 3 - 3 - 3
25. Speed bin (CL - tRCD - tRP) = 3 - 4 - 4 (all speed bins except LPDDR200)
26. tDAL = (tWR/tCK) + (tRP/tCK): for each of the terms, if not already an integer, round to the next higher integer.
27. There must be at least two clock pulses during the tXSR period.
28. There must be at least one clock pulse during the tXP period.
29. tREFI values are dependent on density and bus width.
30. A maximum of 8 Refresh commands can be posted to any given LPDDR, meaning that the maximum absolute interval between any Refresh command and the next Refresh command is  $8 * tREFI$ .
31. It's not supported for package level.

**OUTPUT SLEW RATE CHARACTERISTICS**

PARAMETER	MIN	MAX	UNIT	NOTES
Pull-up and Pull-Down Slew Rate for Full Strength Driver	0.7	2.5	V/ns	1,2
Pull-up and Pull-Down Slew Rate for Three-Quarters Strength Driver	0.5	1.75	V/ns	1,2
Pull-up and Pull-Down Slew Rate for Half Strength Driver	0.3	1.0	V/ns	1,2
Output Slew rate Matching ratio (Pull-up to Pull-down)	0.7	1.4	-	3

NOTES:

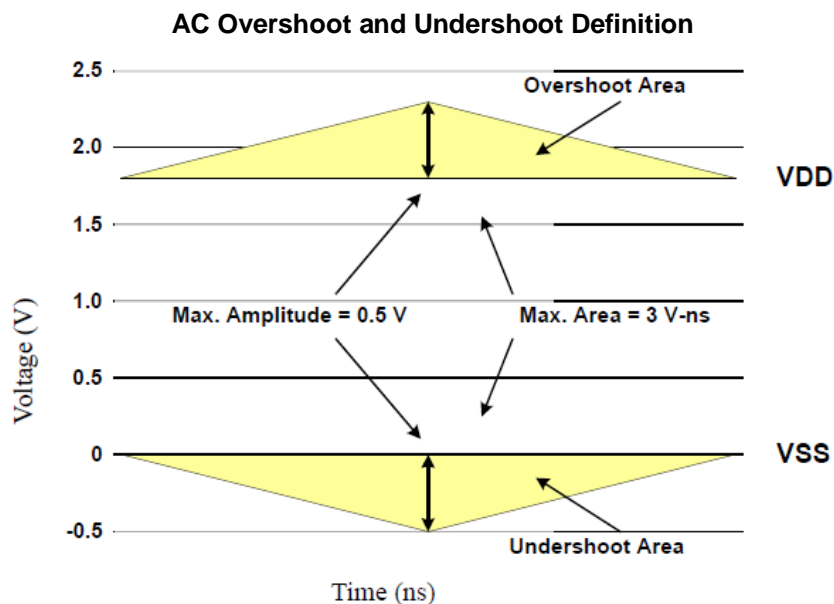
1. Measured with a test load of 20 pF connected to VSSQ.
2. Output slew rate for rising edge is measured between VILD(DC) to VIH(DC) and for falling edge between VIH(DC) to VILD(DC).
3. The ratio of pull-up slew rate to pull-down slew rate is specified for the same temperature and voltage, over the entire temperature and voltage range. For a given output, it represents the maximum difference between pull-up and pull-down drivers due to process variation.

**AC Overshoot/Undershoot Specification**

PARAMETER	SPECIFICATION
Maximum peak amplitude allowed for overshoot	0.5 V
Maximum peak amplitude allowed for undershoot	0.5 V
The area between overshoot signal and VDD must be less than or equal to	3 V-ns
The area between undershoot signal and GND must be less than or equal to	3 V-ns

NOTES:

1. This specification is intended for devices with no clamp protection and is guaranteed by design.



**OUTPUT DRIVE STRENGTH CHARACTERISTICS**

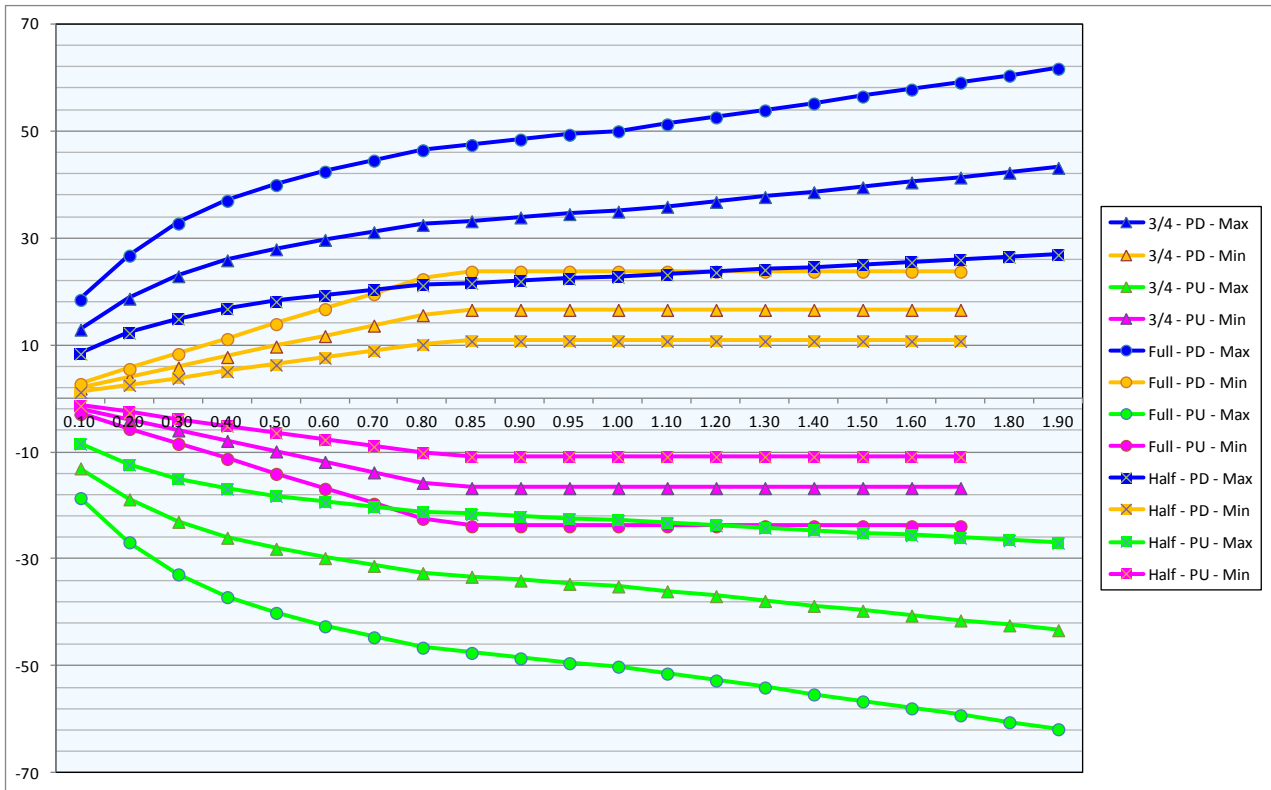
VOLTAGE [V]	FULL DRIVE STRENGTH				HALF DRIVE STRENGTH				THREE-QUARTERS DRIVE STRENGTH			
	PULL-DOWN CURRENT [mA]		PULL-UP CURRENT [mA]		PULL-DOWN CURRENT [mA]		PULL-UP CURRENT [mA]		PULL-DOWN CURRENT [mA]		PULL-UP CURRENT [mA]	
	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX
0	0	0	0	0	0	0	0	0	0	0	0	0
0.1	2.8	18.53	-2.8	-18.53	1.27	8.42	-1.27	-8.42	1.96	12.97	-1.96	-12.97
0.2	5.6	26.8	-5.6	-26.8	2.55	12.3	-2.55	-12.3	3.92	18.76	-3.92	-18.76
0.3	8.4	32.8	-8.4	-32.8	3.82	14.95	-3.82	-14.95	5.88	22.96	-5.88	-22.96
0.4	11.2	37.05	-11.2	-37.05	5.09	16.84	-5.09	-16.84	7.84	25.94	-7.84	-25.94
0.5	14	40	-14	-40	6.36	18.2	-6.36	-18.2	9.8	28	-9.8	-28
0.6	16.8	42.5	-16.8	-42.5	7.64	19.3	-7.64	-19.3	11.76	29.75	-11.76	-29.75
0.7	19.6	44.57	-19.6	-44.57	8.91	20.3	-8.91	-20.3	13.72	31.2	-13.72	-31.2
0.8	22.4	46.5	-22.4	-46.5	10.16	21.2	-10.16	-21.2	15.68	32.55	-15.68	-32.55
0.85	23.8	47.48	-23.8	-47.48	10.8	21.6	-10.8	-21.6	16.66	33.24	-16.66	-33.24
0.9	23.8	48.5	-23.8	-48.5	10.8	22	-10.8	-22	16.66	33.95	-16.66	-33.95
0.95	23.8	49.4	-23.8	-49.4	10.8	22.45	-10.8	-22.45	16.66	34.58	-16.66	-34.58
1	23.8	50.05	-23.8	-50.05	10.8	22.73	-10.8	-22.73	16.66	35.04	-16.66	-35.04
1.1	23.8	51.35	-23.8	-51.35	10.8	23.21	-10.8	-23.21	16.66	35.95	-16.66	-35.95
1.2	23.8	52.65	-23.8	-52.65	10.8	23.67	-10.8	-23.67	16.66	36.86	-16.66	-36.86
1.3	23.8	53.95	-23.8	-53.95	10.8	24.14	-10.8	-24.14	16.66	37.77	-16.66	-37.77
1.4	23.8	55.25	-23.8	-55.25	10.8	24.61	-10.8	-24.61	16.66	38.68	-16.66	-38.68
1.5	23.8	56.55	-23.8	-56.55	10.8	25.08	-10.8	-25.08	16.66	39.59	-16.66	-39.59
1.6	23.8	57.85	-23.8	-57.85	10.8	25.54	-10.8	-25.54	16.66	40.5	-16.66	-40.5
1.7	23.8	59.15	-23.8	-59.15	10.8	26.01	-10.8	-26.01	16.66	41.41	-16.66	-41.41
1.8	—	60.45	—	-60.45	—	26.48	—	-26.48	—	42.32	—	-42.32
1.9	—	61.75	—	-61.75	—	26.95	—	-26.95	—	43.23	—	-43.23

NOTES:

1. Based on nominal impedance of 25 Ohms (Full Drive), 55 Ohms (Half Drive) and 36 Ohms(Three-Quarters) at VDDQ/2
2. The full variation in driver current from minimum to maximum due to process, temperature and voltage will lie within the outer bounding lines of the I-V curve.
3. The I-V current for the optional quarter drive strength is approximately 50% of the half drive strength.
4. The IV current for the Three-Quarters Strength Driver is approximately 70% of the full drive strength current.
5. Implementation and availability of Three-Quarters Strength Driver is optional for speed bins LPDDR333 and below.

### I-V Curves for Full, Three-Quarters and Half Drive Strength

Characteristics are specified under best and worst process variation/conditions



## Basic Functionality

The LPDDR SDRAM is a high-speed CMOS, dynamic random access memory internally configured as a four-bank DRAM. The double data rate architecture is essentially a 2n prefetch with an interface designed to transfer two data words per clock cycle at the I/O pins. A single read or write access for the LPDDR SDRAM effectively consists of a single 2n-bit wide, one clock cycle data transfer at the internal DRAM core and two corresponding n-bit wide, one-half-clock-cycle data transfers at the I/O pins.

Read and write access to the LPDDR SDRAM are burst oriented; access start at a selected location and continue for a programmed number of locations in a programmed sequence. Operation begins with the registration of an ACTIVE command, which is then followed by a READ or WRITE command. The address bits registered coincident with the ACTIVE command are used to select the bank and the row to be activated (BA0-BA1 select the bank; A0-A13 select the row). The address bits registered coincident with the READ or WRITE command are used to select the bank and the starting column location for the burst operation. The Mobile DDR SDRAM provides for programmable READ or WRITE burst lengths of 2, 4, 8, or 16. An auto precharge function may be enabled to provide a self-timed row precharge that is initiated at the end of the burst access. As with standard DDR SDRAMs, the pipelined, multibank architecture of the Mobile DDR SDRAMs supports concurrent operation, thereby providing high effective bandwidth by hiding row precharge and activation time.

An auto refresh mode is provided, along with a power saving power-down mode. Deep power-down mode is offered to achieve maximum power reduction by eliminating the power of the memory array. Data will not be retained after device enters deep power-down mode. Two self refresh features, temperature-compensated self refresh (TCSR) and partial array self refresh (PASR), offer additional power saving. TCSR is controlled by the automatic on-chip temperature sensor. The PASR can be customized using the extended mode register settings. The two features may be combined to achieve even greater power saving. The DLL that is typically used on standard DDR devices is not necessary on the Mobile DDR SDRAM. It has been omitted to save power.

Prior to normal operation, the LPDDR SDRAM must be initialized. The following sections provide detailed information covering device initialization, register definition, command descriptions and device operation.

## Initialization

LPDDR SDRAMs must be powered up and initialized in a predefined manner. Operations procedures other than those specified may result in undefined operation. If there is any interruption to the device power, the initialization routine should be followed. The steps to be followed for device initialization are listed below.

The Mode Register and Extended Mode Register do not have default values. If they are not programmed during the initialization sequence, it may lead to unspecified operation. The clock stop feature is not available until the device has been properly initialized from Steps 1 through 11.



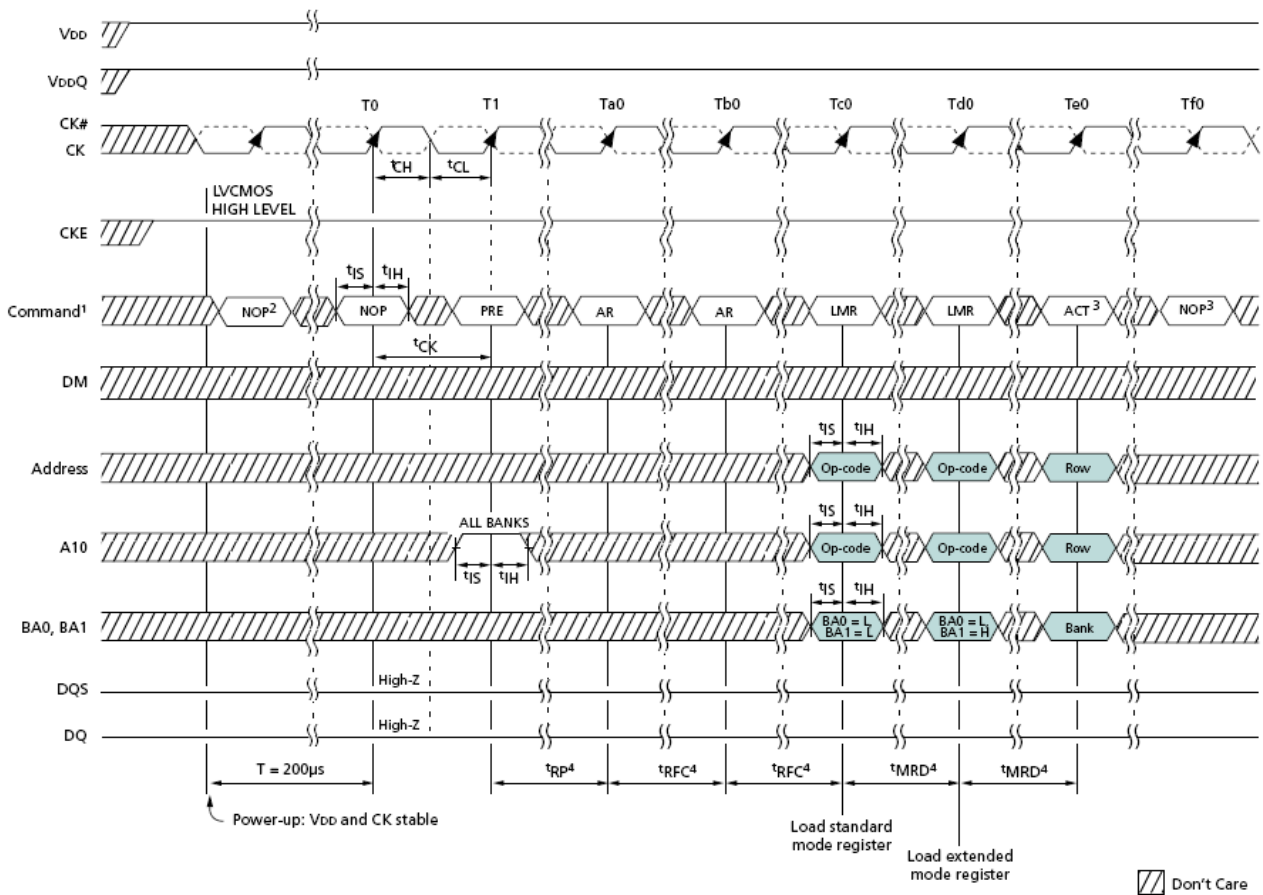
The Following sequence is required for POWER UP and Initialization

1. Provide power, the device core power (VDD) and the device I/O power (VDDQ) must be brought up simultaneously to prevent device latch-up. Although not required, it is recommended that VDD and VDDQ are from the same power source. Also assert and hold Clock Enable (CKE) to a LV-CMOS logic high level
2. Once the system has established consistent device power and CKE is driven high, it is safe to apply stable clock
3. There must be at least 200  $\mu$ s of valid clocks before any command may be given to the DRAM. During this time NOP or DESELECT commands must be issued on the command bus.
4. Issue a PRECHARGE ALL command.
5. Provide NOPs or DESELECT commands for at least tRP time.
6. Issue an AUTO REFRESH command followed by NOPs or DESELECT command for at least tRFC time. Issue the second AUTO REFRESH command followed by NOPs or DESELECT command for at least tRFC time. Note as part of the initialization sequence there must be two auto refresh commands issued. The typical flow is to issue them at Step 6, but they may also be issued between steps 10 and 11.
7. Using the MRS command, load the base mode register. Set the desired operating modes.
8. Provide NOPs or DESELECT commands for at least tMRD time.
9. Using the MRS command, program the extended mode register for the desired operating modes. Note the order of the base and extended mode register programming is not important.
10. Provide NOP or DESELCT commands for at least tMRD time.
11. The DRAM has been properly initialized and is ready for any valid command.

### Brief Description of Initialization Sequence

Step	Description for Initialization
1	VDD and VDDQ Ramp: CKE must be held high
2	Apply stable clocks
3	Wait at least 200 $\mu$ s with NOP or DESELECT on command bus
4	PRECHARGE ALL
5	Assert NOP or DESELCT for tRP time
6	Issue two AUTOREFRESH commands each followed by NOP or DESELECT commands for tRFC time
7	Configure Mode Register
8	Assert NOP or DESELECT for tMRD time
9	Configure Extended Mode Register
10	Assert NOP or DESELECT for tMRD time
11	LPDDR SDRAM is ready for any valid command

### Initialization Sequence Diagram



#### NOTES:

1. PRE = PRECHARGE command; LMR = LOAD MODE REGISTER command; AR = AUTO REFRESH command; ACT = ACTIVE command.
2. NOP or DESELECT commands are required for at least 200us.
3. Other valid commands are possible.
4. NOPs or DESELECTs are required during this time.

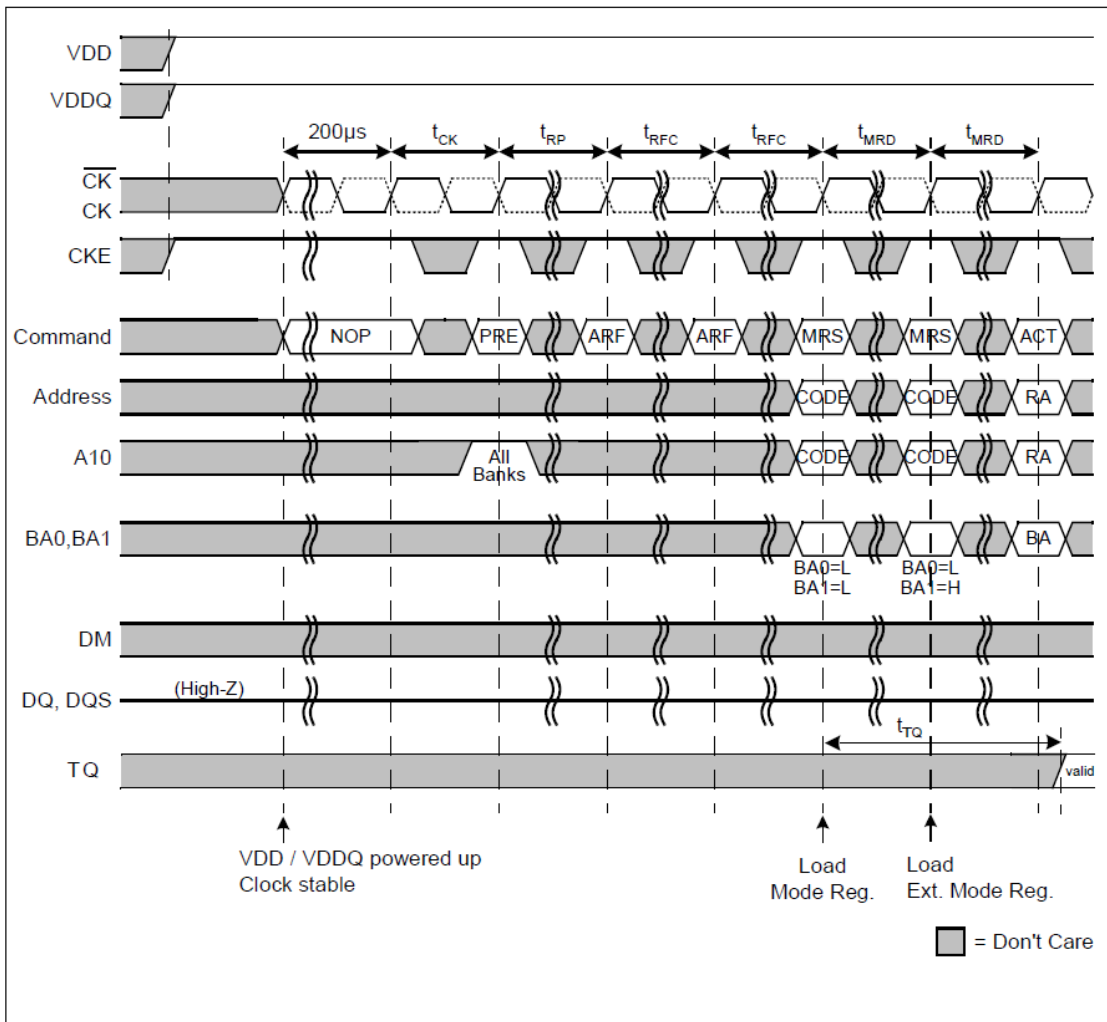
**Initialization Sequence with Temperature Output Signal (Unsupported on package level)**

**TQ Signal Initialization**

During device initialization the TQ signal output will be invalid until  $t_{TQ}$  after the first MRS command. Following  $t_{TQ}$  the TQ signal will output logic-HIGH when the device temperature is greater than, or equal to, 85°C, and logic-LOW when the device temperature is less than 85°C. There is no high-impedance state for this output signal.

**Temperature Output Signal**

The LPDDR-SDRAM device may include an optional temperature output signal (TQ). This signal is an asynchronous LVCMOS output which outputs a logic-HIGH when the device temperature is greater than, or equal to, 85°C and a logic-LOW when the device temperature is less than 85°C. There is no high-impedance state output from this signal. The TQ output signal activates even during clock stop, power down, and self refresh modes. The signal is not valid during initialization and becomes valid after  $t_{TQ}$  following the first MRS command. When TQ output is logic-HIGH,  $t_{REF}$  is specified to be 16ms. Additionally, AC parameters shall be de-rated to 20% and DC parameters shall not be guaranteed.



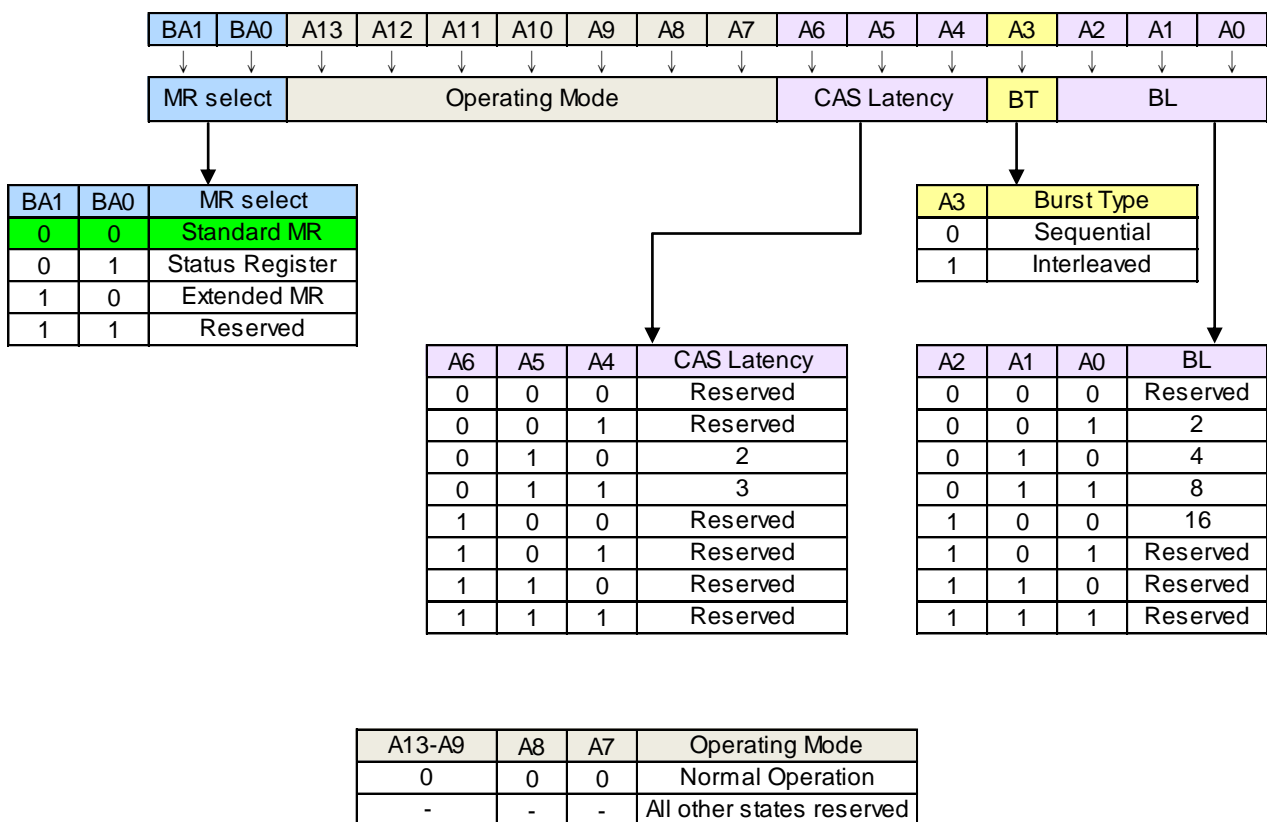
## Register Definition

### Mode Registers and Extended Mode Registers

The Mode Registers are used to define the specific mode of operation of the LPDDR SDRAM. This define includes the definition of a burst length, a burst type, a CAS latency. Additionally, driver strength, Temperature Compensated Self Refresh (TCSR), and Partial Array Self Refresh (PASR) are also user defined variables and must be programmed with an Extended Mode Register Set (EMRS) command. The default value of the mode register is not defined, therefore the mode register must be written after power up for proper operation. The Mode Register must be loaded when all banks are idle and no bursts are progress, and the controller must wait the specific time <sup>t</sup>MRD before initiating any subsequent operation. Violating either of these requirements will result in unspecified operation. The MRS contents won't be changed until it is reprogrammed, the device goes into Deep Power-Down, or the device loses power.

The mode register is written by asserting low on  $\overline{CS}$ ,  $\overline{RAS}$ ,  $\overline{CAS}$ ,  $\overline{WE}$ , BA0 and BA1, while controlling the state of address pins A0~A13. The mode register contents can be changed using the same command and clock cycle requirements during normal operation as long as all banks are in the precharge state. The mode register is divided into various fields depending on the functionality. Burst length is defined by A0~A2 with options of 2, 4, 8 and 16 bit burst length. Burst address sequence type is defined by A3 and CAS latency is defined by A4~A6. A7~A13 must be set to low to ensure future compatibility.

### Standard Mode Register definition



NOTE 1 : A logic 0 should be programmed to all unused / undefined address bits to ensure future compatibility.

### Burst Length, Type, and Order

Accesses within a given burst may be programmed to sequential or interleaved order. The burst type is selected via bit A3 as above figure. The ordering of access within a burst is determined by the burst length, burst type, and the starting column address. The burst length determines the maximum number of column locations that can be accessed for a given READ or WRITE command. The burst length is defined by bits A0–A2. Burst length options include 2, 4, 8 or 16 for both the sequential and the interleaved burst types.

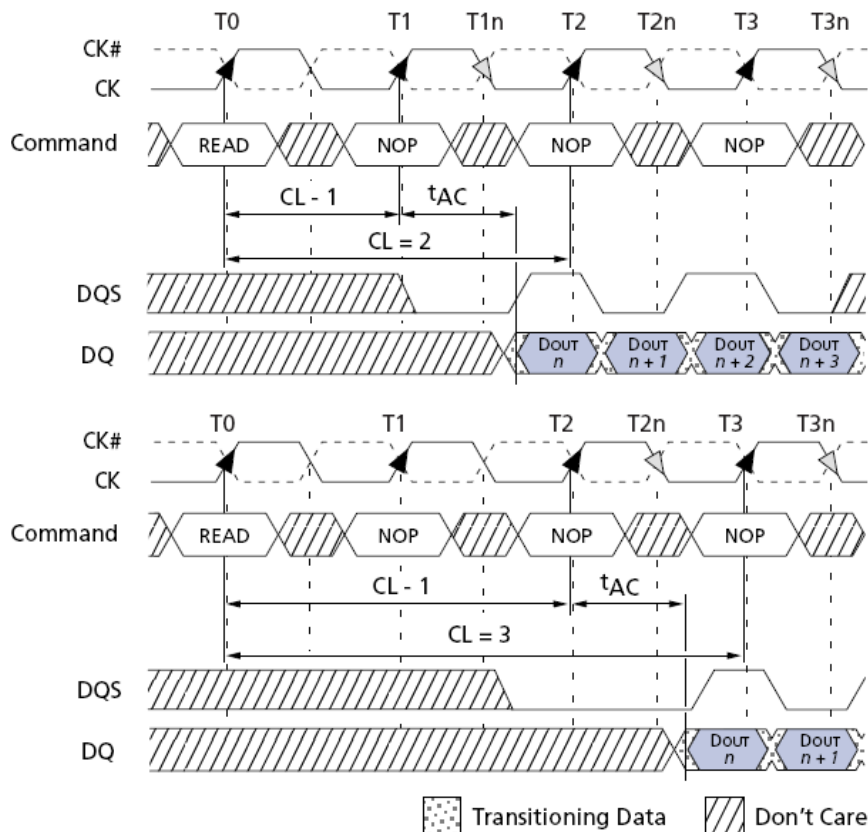
When a READ or WRITE command is issued, a block of columns equal to the BL is effectively selected. All accesses for that burst take place within this block, meaning that the burst will wrap when a boundary is reached. The block is uniquely selected by A1–Ai when BL = 2, by A2–Ai when BL = 4, by A3–Ai when BL = 8, and by A4–Ai when BL = 16 (where Ai is the most significant column address bit for a given configuration). The remaining (least significant) address bits are used to specify the starting location within the block. The programmed BL applies to both READ and WRITE bursts. Accesses within a given burst may be programmed to be either sequential or interleaved via the standard mode register.

### Burst Type and Burst Order

Burst Length	Starting Column Address				Burst Type	
	A3	A2	A1	A0	Sequential	Interleaved
2	-	-	-	0	0,1	0,1
	-	-	-	1	1,0	1,0
4	-	-	0	0	0,1,2,3	0,1,2,3
	-	-	0	1	1,2,3,0	1,0,3,2
	-	-	1	0	2,3,0,1	2,3,0,1
	-	-	1	1	3,0,1,2	3,2,1,0
8	-	0	0	0	0,1,2,3,4,5,6,7	0,1,2,3,4,5,6,7
	-	0	0	1	1,2,3,4,5,6,7,0	1,0,3,2,5,4,7,6
	-	0	1	0	2,3,4,5,6,7,0,1	2,3,0,1,6,7,4,5
	-	0	1	1	3,4,5,6,7,0,1,2	3,2,1,0,7,6,5,4
	-	1	0	0	4,5,6,7,0,1,2,3	4,5,6,7,0,1,2,3
	-	1	0	1	5,6,7,0,1,2,3,4	5,4,7,6,1,0,3,2
	-	1	1	0	6,7,0,1,2,3,4,5	6,7,4,5,2,3,0,1
	-	1	1	1	7,0,1,2,3,4,5,6	7,6,5,4,3,2,1,0
16	0	0	0	0	0,1,2,3,4,5,6,7,8,9,A,B,C,D,E,F	0,1,2,3,4,5,6,7,8,9,A,B,C,D,E,F
	0	0	0	1	1,2,3,4,5,6,7,8,9,A,B,C,D,E,F,0	1,0,3,2,5,4,7,6,9,8,B,A,D,C,F,E
	0	0	1	0	2,3,4,5,6,7,8,9,A,B,C,D,E,F,0,1	2,3,0,1,6,7,4,5,A,B,8,9,E,F,C,D
	0	0	1	1	3,4,5,6,7,8,9,A,B,C,D,E,F,0,1,2	3,2,1,0,7,6,5,4,B,A,9,8,F,E,D,C
	0	1	0	0	4,5,6,7,8,9,A,B,C,D,E,F,0,1,2,3	4,5,6,7,0,1,2,3,C,D,E,F,8,9,A,B
	0	1	0	1	5,6,7,8,9,A,B,C,D,E,F,0,1,2,3,4	5,4,7,6,1,0,3,2,D,C,F,E,9,8,B,A
	0	1	1	0	6,7,8,9,A,B,C,D,E,F,0,1,2,3,4,5	6,7,4,5,2,3,0,1,E,F,C,D,A,B,8,9
	0	1	1	1	7,8,9,A,B,C,D,E,F,0,1,2,3,4,5,6	7,6,5,4,3,2,1,0,F,E,D,C,B,A,9,8
	1	0	0	0	8,9,A,B,C,D,E,F,0,1,2,3,4,5,6,7	8,9,A,B,C,D,E,F,0,1,2,3,4,5,6,7
	1	0	0	1	9,A,B,C,D,E,F,0,1,2,3,4,5,6,7,8	9,8,B,A,D,C,F,E,1,0,3,2,5,4,7,6
	1	0	1	0	A,B,C,D,E,F,0,1,2,3,4,5,6,7,8,9	A,B,8,9,E,F,C,D,2,3,0,1,6,7,4,5
	1	0	1	1	B,C,D,E,F,0,1,2,3,4,5,6,7,8,9,A	B,A,9,8,F,E,D,C,3,2,1,0,7,6,5,4
	1	1	0	0	C,D,E,F,0,1,2,3,4,5,6,7,8,9,A,B	C,D,E,F,8,9,A,B,4,5,6,7,0,1,2,3
	1	1	0	1	D,E,F,0,1,2,3,4,5,6,7,8,9,A,B,C	D,C,F,E,9,8,B,A,5,4,7,6,1,0,3,2
	1	1	1	0	E,F,0,1,2,3,4,5,6,7,8,9,A,B,C,D	E,F,C,D,A,B,8,9,6,7,4,5,2,3,0,1
	1	1	1	1	F,0,1,2,3,4,5,6,7,8,9,A,B,C,D,E	F,E,D,C,B,A,9,8,7,6,5,4,3,2,1,0

### CAS Latency (CL)

The CAS Latency, or READ latency is the delay, in clock cycles, between the registration of a Read command and the availability of the first bit of output data. CAS Latency is defined by bit A6~A4 in the standard mode register. If a READ command is registered at a clock edge  $n$ , and the CAS latency is 3 clocks, the first data element will be valid at  $(n + 2tCK + tAC)$ . If a READ command is registered at a clock edge  $n$ , and the CAS latency is 2 clocks, the first data element will be valid at  $(n + 1tCK + tAC)$ .



### Extended Mode Register definition

The Extended Mode Register controls functions beyond those controlled by the Mode Register; these additional functions include output drive strength selection, Temperature Compensated Self Refresh (TCSR) and Partial Array Self Refresh (PASR). TCSR and PASR are effective in Self Refresh mode only. The extended mode register is programmed via the LOAD MODE REGISTER command with  $BA0=0$  and  $BA1=1$ , and the information won't be changed until it is reprogrammed, the device goes into deep power-down mode, or the device loses power. The EMRS must be loaded when all banks are idle and no bursts are in progress, and the controller must wait the specified time  $tMRD$  before initiating any subsequent operation. Violating either of these requirements will result in unspecified operation. Address bits A0-A2 specify PASR, A3-A4 the TCSR, A5-A6 the Drive Strength. A logic 0 should be programmed to all the undefined addresses bits to ensure future compatibility.

### Temperature Compensated Self Refresh (TCSR)

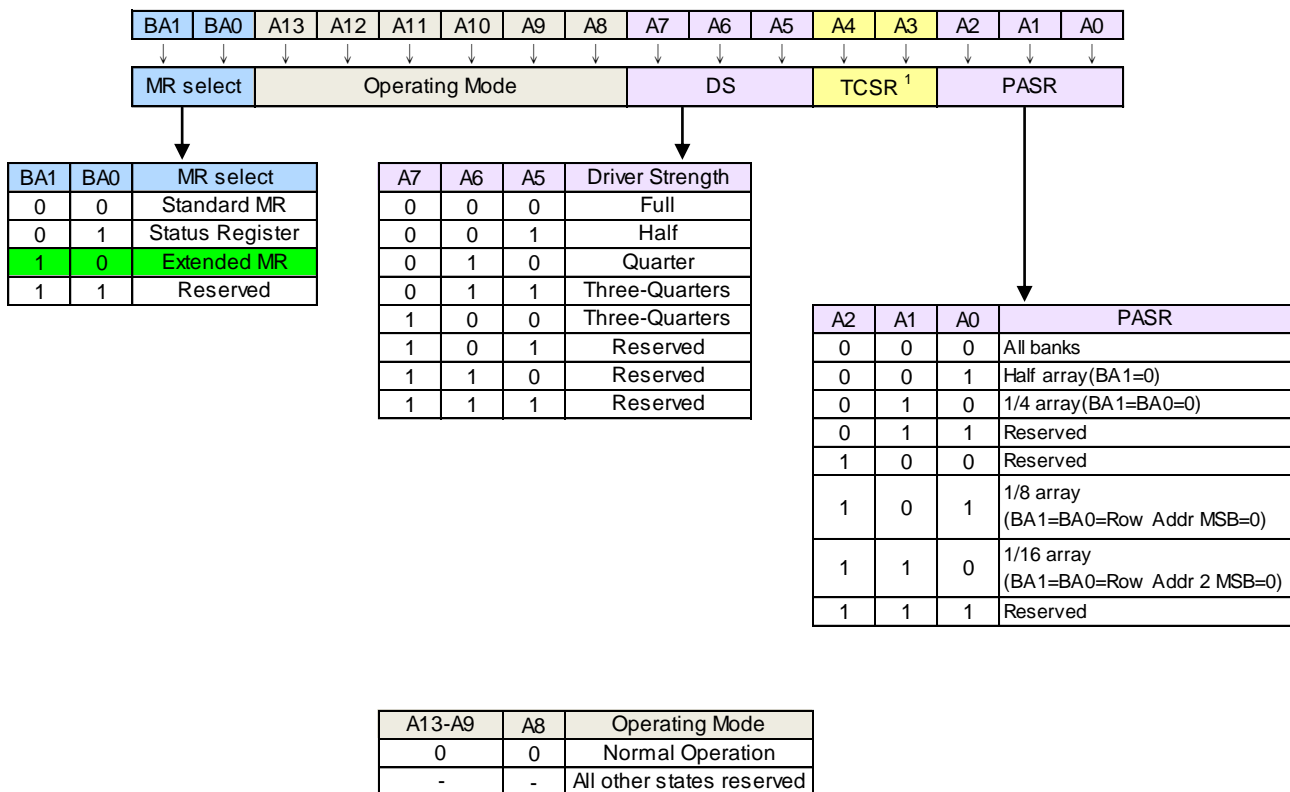
On this version of the LPDDR SDRAM, the internal temperature sensor is implemented to adjust the self refresh oscillator automatically base on the case temperature. To maintain backward compatibility, the programming of TCSR bits no effect on the device so the address bits, A3 and A4 are ignored (don't care) during EMRS programming.

### Partial-Array Self Refresh (PASR)

For further power savings during self refresh, the PASR feature may allow the self refresh to be restricted to a variable portion of the total array. They are full array (default: banks 0, 1, 2, and 3), 1/2 array (banks 0 and 1), 1/4 array (bank 0), 1/8 array (bank 0 with row address MSB=0), and 1/16 array (bank 0 with row address MSB=0, and row address MSB-1=0). Data outside the defined area will be lost. Address bits A0 to A2 are used to set PASR.

### Output Drive Strength

LPDDR SDRAM provides the option to control the drive strength of the output buffers for the smaller systems or point-to-point environments. The value was selected based on the expected loading of the memory bus. Total four values provided, and they are 25 ohm, 36ohm, 55ohm, and 80ohm internal impedance. They are full, three-quarter, one-half, and one-quarter drive strengths, respectively.



### Extended Mode Register

NOTE 1: On-die temperature sensor is used in place of TCSR. Setting these bits will have no effect.

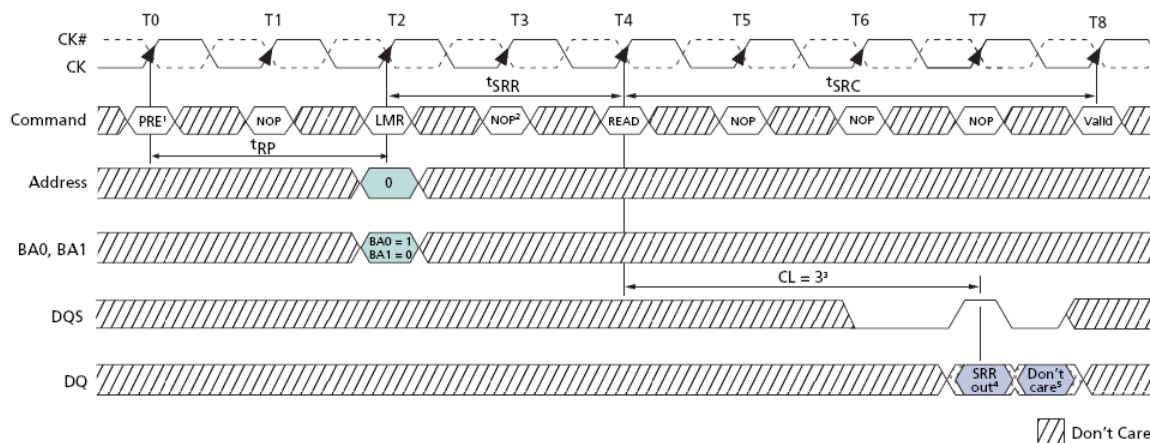
NOTE 2: A logic 0 should be programmed to all unused / undefined address bits to ensure future compatibility.

NOTE 3: Implementation and availability of Three-Quarters Strength Driver is optional for speed bins LPDDR333 and below.

### Status Read Register (SRR)

The status read register (SRR) is only for READ, and contains the specific die information such as density, device type, data bus width, refresh rate, revision ID and manufactures. The SRR is read via the LOAD MODE REGISTER command with BA0=1 and BA1=0. The sequence to perform an SRR command is as follows:

- The device had been properly initialized and in the idle or all banks precharge state.
- Issue a LMR command with BA [1:0] = "01".
- Wait t<sub>SRR</sub>; only NOP or DESELECT commands are supported during this period.
- Issue a READ command with all address pins set to "0".
- CAS latency cycles later, the device returns the registers data. The SRR read with fixed burst length 2, first bit of the burst output SRR data, and second bit of the burst is "Don't Care".
- The next command to the SDRAM must be issued t<sub>SRC</sub> after the SRR READ command is issued; only NOP or DESELECT commands are supported during this period.

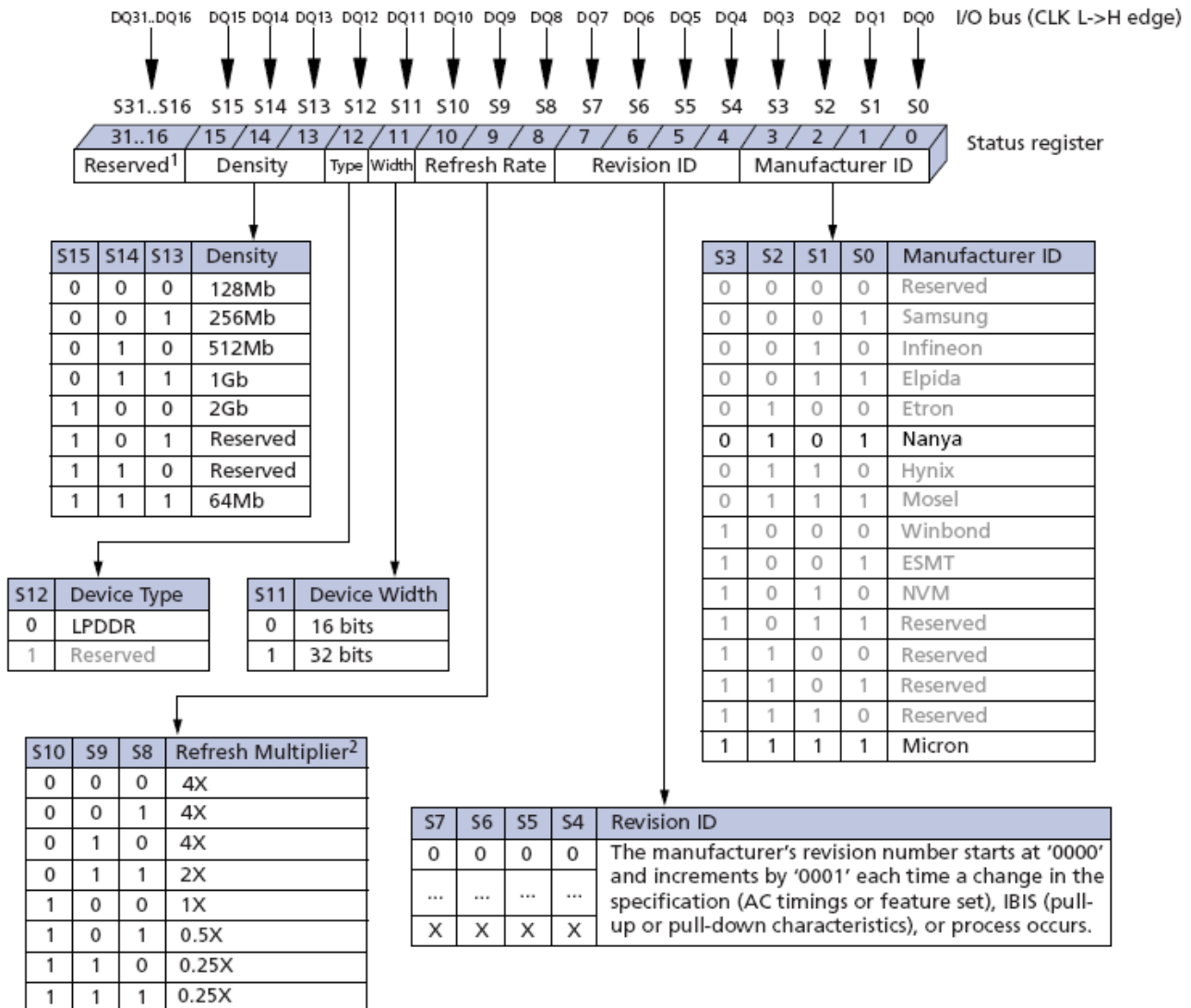


#### NOTES:

1. SRR can only be issued after power-up sequence is complete, and all banks are precharged and in the idle state.
2. NOP or DESELECT commands are required between LMR and READ command (t<sub>SRR</sub>) and between READ and next VALID command (t<sub>SRC</sub>)
3. CAS latency is predetermined by the programming of the mode register. Here CL=3 as an example only.
4. Burst length is fixed to 2 for SRR regardless of the value programmed by the mode register.
5. The second bit of the data-out burst is a "Don't Care".



Register Definition



Status Register Definition

NOTES:

1. Reserved bits should be set to zero for future compatibility.
2. Refresh multiplier is based on the memory device's on-board temperature sensor. Required average periodic refresh interval = tREFI x multiplier.

### LPDDR SDRAM Command Description and Operation

#### Command Truth Table

NANE (Function)	Abbr.	$\overline{CS}$	RAS	CAS	WE	BA	A10/AP	ADDR	NOTES
DESELECT	DESELECT	H	X	X	X	X	X	X	2
NO OPERATION	NOP	L	H	H	H	X	X	X	2
ACTIVE (select bank and active row)	ACT	L	L	H	H	Valid	Row	Row	
READ (select bank, column, and start read burst)	READ	L	H	L	H	Valid	L	Col	
READ with AP (read burst with Auto Precharge)	READA	L	H	L	H	Valid	H	Col	3
WRITE (select bank, column, and start write burst)	WRITE	L	H	L	L	Valid	L	Col	
WRITE with AP (write burst with Auto Precharge)	WRITEA	L	H	L	L	Valid	H	Col	3
BURST TERMINATE or enter Deep Power-Down	BST	L	H	H	L	X	X	X	4,5
PRECHARGE (deactive row in selected bank)	PRE	L	L	H	L	Valid	L	X	6
PRECHARGE ALL (deactive rows in all banks)	PREALL	L	L	H	L	X	H	X	6
AUTO REFRESH or enter SELF REFRESH	REFA / REFS	L	L	L	H	X	X	X	7,8,9
LOAD MODE REGISTER	LMR	L	L	L	L	Valid	Op-code		10

**NOTES:**

- All states and sequences not shown are illegal or reserved.
- DESELECT and NOP are functionally interchangeable.
- Autoprecharge is non-persistent. A10 High enables Auto Precharge, while A10 Low disables Autoprecharge
- Burst Terminate applies to only Read bursts with Auto Precharge disabled. This command is undefined and should not be used for Read with Auto Precharge enabled, and for Write bursts.
- This command is BURST TERMINATE if CKE is High and DEEP POWER DOWN entry if CKE is Low.
- If A10 is Low, bank address determines which bank is to be precharged. If A10 is High, all banks are precharged and BA0-BA1are don't care.
- This command is AUTO REFRESH if CKE is High, and SELF REFRESH if CKE is low.
- All address inputs and I/O are 'don't care' except for CKE. Internal refresh counters control bank and row addressing.
- All banks must be precharged before issuing an AUTO-REFRESH or SELF REFRESH command.
- BA0 and BA1 value select between MRS and EMRS.
- CKE is HIGH for all commands shown except SELF REFRESH and DEEP POWER-DOWN.

### DM Operation Truth Table

Function	DM	DQ	Notes
Write Enable	L	Valid	1,2
Write Inhibit	H	X	1,2

NOTES:

1. Used to mask write data, provided coincident with the corresponding data.
2. All states and sequences not shown are reserved and/or illegal.

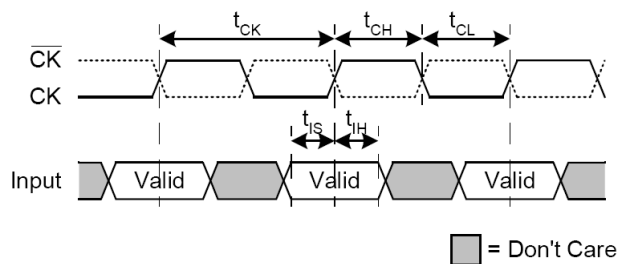
**CKE Truth Table**

Current State	CKE		Command (n) RAS, CAS, WE, CS	Action (n) -Result	Notes
	CKE n-1	CKE n			
Power Down	L	L	X	Maintain Power Down	
Self Refresh	L	L	X	Maintain Self Refresh	
Deep Power Down	L	L	X	Maintain Deep Power Down	
Power Down	L	H	NOP or DESELECT	Exit Power Down	5, 6, 9
Self Refresh	L	H	NOP or DESELECT	Exit Self Refresh	5, 7, 10
Deep Power Down	L	H	NOP or DESELECT	Exit Deep Power Down	5, 8
All Banks Idle	H	L	NOP or DESELECT	Precharge Power Down Entry	5
Bank(s) Active	H	L	NOP or DESELECT	Active Power Down Entry	5
All Banks Idle	H	L	AUTO REFRESH	Self Refresh Entry	
All Banks Idle	H	L	BURST TERMINATE	Deep Power Down Entry	
See the other Truth Tables	H	H	See the other Truth Tables		

**NOTES:**

1. CKEn is the logic state of CKE at clock edge n; CKEn-1 was the state of CKE at the previous clock edge.
2. Current state is the state of LPDDR immediately prior to clock edge n.
3. COMMANDn is the command registered at clock edge n, and ACTIONn is the result of COMMANDn.
4. All states and sequences not shown are illegal or reserved.
5. DESELECT and NOP are functionally interchangeable.
6. Power Down exit time (tXP) should elapse before a command other than NOP or DESELECT is issued.
7. SELF REFRESH exit time (tXSR) should elapse before a command other than NOP or DESELECT is issued.
8. The Deep Power-Down exit procedure must be followed as discussed in the Deep Power-Down section of the Functional Description.
9. The clock must toggle at least once during the tXP period.
10. The clock must toggle at least once during the tXSR time.

**Basic Timing Parameters for Commands**



NOTE 1: Input = A0 – An, BA0, BA1, CKE,  $\overline{CS}$ ,  $\overline{RAS}$ ,  $\overline{CAS}$ ,  $\overline{WE}$ ; An = Address bus MSB.

**Current State Bank n Truth Table (command to Bank n)**

Current State	Command				Description	Action (n) -Result	Notes
	CS	RAS	CAS	WE			
Any	H	X	X	X	DESELECT (NOP)	Continue previous operation	
	L	H	H	H	NOP	Continue previous operation	
Idle	L	L	H	H	ACTIVE	Select and Active row	
	L	L	L	H	AUTO REFRESH	Auto refresh	10
	L	L	L	L	MODE REGISTER SET	Mode register set	10
Row Active	L	H	L	H	READ	Select column & start read burst	
	L	H	L	L	WRITE	Select column & start write burst	
	L	L	H	L	PRECHARGE	Deactive row in bank or banks	4
READ (AP disable)	L	H	L	H	READ	Select column & start new read burst	5,6
	L	H	L	L	WRITE	Select column & start write burst	5,6,13
	L	L	H	L	PRECHARGE	Truncate read burst, start precharge	
	L	H	H	L	BURST TERMINTTE	Burst terminate	11
WRITE (AP disable)	L	H	L	H	READ	Select column & start read burst	5,6,12
	L	H	L	L	WRITE	Select column & start new write burst	5,6
	L	L	H	L	PRECHARGE	Truncate write burst, start precharge	12

NOTES:

- The table applies when both CKEn-1 and CKEn are HIGH, and after tXSR or tXP has been met if the previous state was Self Refresh or Power Down.
- DESELECT and NOP are functionally interchangeable.
- All states and sequences not shown are illegal or reserved.
- This command may or may not be bank specific. If all banks are being precharged, they must be in a valid state for precharging.
- A command other than NOP should not be issued to the same bank while a READ or WRITE burst with Auto Precharge is enabled.
- The new Read or Write command could be Auto Precharge enabled or Auto Precharge disabled.
- Current State Definitions:
  - Idle: The bank has been precharged, and tRP has been met.
  - Row Active: A row in the bank has been activated, and tRCD has been met. No data bursts / accesses and no register accesses are in progress.
  - Read: A READ burst has been initiated, with Auto Precharge disabled, and has not yet terminated or been terminated.
  - Write: a WRITE burst has been initiated, with Auto Precharge disabled, and has not yet terminated or been terminated.
- The following states must not be interrupted by a command issued to the same bank. DESELECT or NOP commands or allowable commands to the other bank should be issued on any clock edge occurring during these states. Allowable commands to the other bank are determined by its current state and Truth Table - Current State Bank n - Command to Bank n/m.
  - Precharging: starts with the registration of a PRECHARGE command and ends when tRP is met. Once tRP is met, the bank will be in the idle state.

- Row Activating: starts with registration of an ACTIVE command and ends when tRCD is met. Once tRCD is met, the bank will be in the 'row active' state.
  - Read with AP Enabled: starts with the registration of the READ command with Auto Precharge enabled and ends when tRP has been met. Once tRP has been met, the bank will be in the idle state.
  - Write with AP Enabled: starts with registration of a WRITE command with Auto Precharge enabled and ends when tRP has been met. Once tRP is met, the bank will be in the idle state.
9. The following states must not be interrupted by any executable command; DESELECT or NOP commands must be applied to each positive clock edge during these states.
- Refreshing: starts with registration of an AUTO REFRESH command and ends when tRFC is met. Once tRFC is met, the device will be in an 'all banks idle' state.
  - Accessing Mode Register: starts with registration of a MODE REGISTER SET command and ends when tMRD has been met. Once tMRD is met, the device will be in an 'all banks idle' state.
  - Precharging All: starts with the registration of a PRECHARGE ALL command and ends when tRP is met. Once tRP is met, the bank will be in the idle state.
10. Not bank-specific; requires that all banks are idle and no bursts are in progress.
11. Not bank-specific. BURST TERMINATE affects the most recent read burst, regardless of bank.
12. Requires appropriate DM masking.
13. A WRITE command may be applied after the completion of the READ burst; otherwise, a BURST TERMINATE must be used to end the READ prior to asserting a WRITE command.

### Current State Bank n Truth Table (command to Bank m)

Current State	Command				Description	Action (n) -Result	Notes
	$\overline{CS}$	$\overline{RAS}$	$\overline{CAS}$	$\overline{WE}$			
Any	H	X	X	X	DESELECT (NOP)	Continue previous operation	
	L	H	H	H	NOP	Continue previous operation	
Idle	X	X	X	X	ANY	Any command allowed to bank m	
Row Activating, Active, or Precharging	L	L	H	H	ACTIVE	Select and activate row	
	L	H	L	H	READ	Select column & start read burst	8
	L	H	L	L	WRITE	Select column & start write burst	8
	L	L	H	L	PRECHARGE	Precharge	
READ (AP disable)	L	L	H	H	ACTIVE	Select and activate row	
	L	H	L	H	READ	Select column & start read burst	8
	L	H	L	L	WRITE	Select column & start write burst	8,10
	L	L	H	L	PRECHARGE	Precharge	
WRITE (AP disable)	L	L	H	H	ACTIVE	Select and activate row	
	L	H	L	H	READ	Select column & start read burst	8,9
	L	H	L	L	WRITE	Select column & start write burst	8
	L	L	H	L	PRECHARGE	Precharge	
Read (AP enabled)	L	L	H	H	ACTIVE	Select and activate row	
	L	H	L	H	READ	Select column & start read burst	5,8
	L	H	L	L	WRITE	Select column & start write burst	5,8,10
	L	L	H	L	PRECHARGE	Precharge	
Write (AP enabled)	L	L	H	H	ACTIVE	Select and activate row	
	L	H	L	H	READ	Select column & start read burst	5,8
	L	H	L	L	WRITE	Select column & start write burst	5,8
	L	L	H	L	PRECHARGE	Precharge	

#### NOTES:

- The table applies when both  $\overline{CKEn-1}$  and  $\overline{CKEn}$  are HIGH, and after  $t_{XSR}$  or  $t_{XP}$  has been met if the previous state was Self Refresh or Power Down.
- DESELECT and NOP are functionally interchangeable.
- All states and sequences not shown are illegal or reserved.
- Current State Definitions:
  - Idle: the bank has been precharged, and  $t_{RP}$  has been met.
  - Row Active: a row in the bank has been activated, and  $t_{RCD}$  has been met. No data bursts/accesses and no register accesses are in progress.
  - Read: a READ burst has been initiated, with Auto Precharge disabled, and has not yet terminated or been terminated.
  - Write: a WRITE burst has been initiated, with Auto Precharge disabled, and has not yet terminated or been terminated.

5. Read with AP enabled and Write with AP enabled: the Read with Auto Precharge enabled or Write with Auto Precharge enabled states can be broken into two parts: the access period and the precharge period. For Read with AP, the precharge period is defined as if the same burst was executed with Auto Precharge disabled and then followed with the earliest possible PRECHARGE command that still accesses all the data in the burst. For Write with AP, the precharge period begins when tWR ends, with tWR measured as if Auto Precharge was disabled. The access period starts with registration of the command and ends where the precharge period (or tRP) begins. During the precharge period of the Read with AP enabled or Write with AP enabled states, ACTIVE, PRECHARGE, READ, and WRITE commands to the other bank may be applied; during the access period, only ACTIVE and PRECHARGE commands to the other banks may be applied. In either case, all other related limitations apply (e.g. contention between READ data and WRITE data must be avoided).

From Command	To Command	Min. delay (w/ concurrent Auto Precharge)
Write w/ AP	READ or READ w/ AP	$[1 + (BL/2)] tCK + tWTR$
	WRITE or WRITE w/ AP	$(BL/2) tCK$
	PRECHARGE	1 tCK
	ACTIVE	1 tCK
READ w/ AP	READ or READ w/ AP	$(BL/2) tCK$
	WRITE or WRITE w/ AP	$[CL + (BL/2)] tCK$
	PRECHARGE	1 tCK
	ACTIVE	1 tCK

6. AUTO REFRESH, SELF REFRESH, and MODE REGISTER SET commands may only be issued when all bank are idle.
7. A BURST TERMINATE command cannot be issued to another bank; it applies to the bank represented by the current state only.
8. READs or WRITEs listed in the Command column include READs and WRITEs with Auto Precharge enabled and READs and WRITEs with Auto Precharge disabled.
9. Requires appropriate DM masking.
10. A WRITE command may be applied after the completion of data output, otherwise a BURST TERMINATE command must be issued to end the READ prior to asserting a WRITE command.



## COMMAND

### NO OPERATION (NOP)

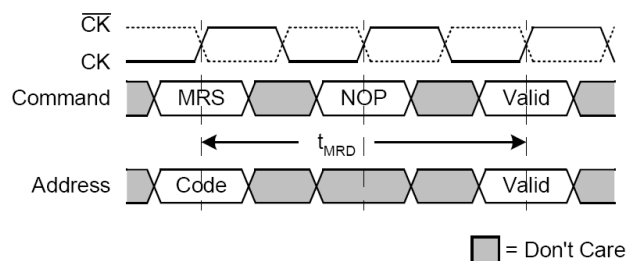
The No operation (NOP) command is used to instruct the selected LPDDR SDRAM to perform a NOP. This prevents unwanted commands from being registered during idle or wait states. Operations already in progress are not affected.

### DESELECT

The Deselect function ( $\overline{CS}$ =HIGH) prevents new commands from being executed by the LPDDR SDRAM. Operations already in progress are not affected.

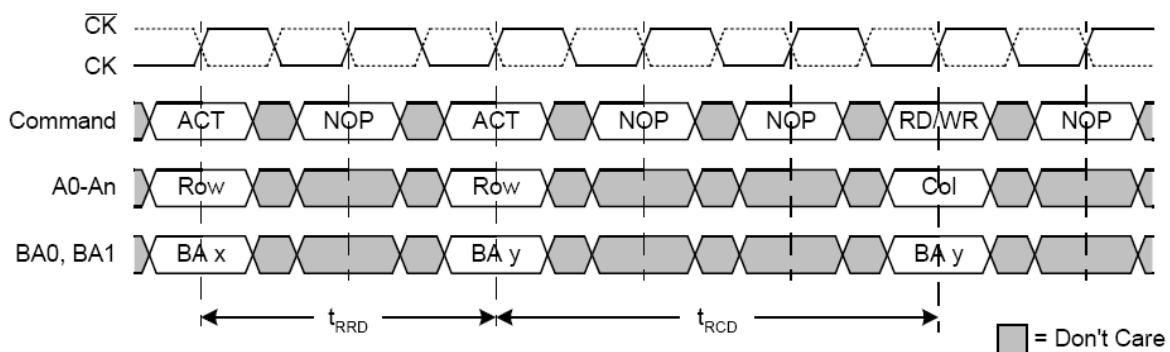
### LOAD MODE REGISTER

The mode registers are loaded via the address inputs and can only be issued when all banks are idle, no bursts are in progress. The subsequent executable command can not be issued until  $t_{MRD}$  is met.



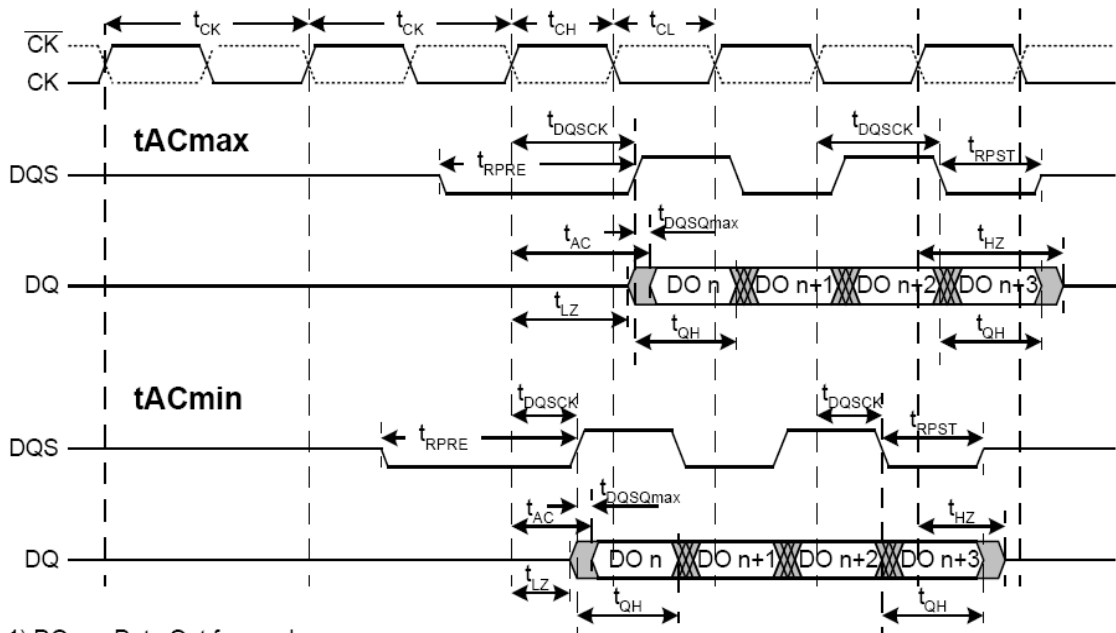
### ACTIVE

The ACTIVE command is used to open (or activate) a row in a particular bank for subsequent access. The values on the BA0 and BA1 inputs select the bank, and the addresses provided on inputs A0-A13 selects the row. Once a row is open, a READ or WRITE command could be issued to that row, subject to the  $t_{RCD}$  specification. A subsequent ACTIVE command to another row in the same bank can only be issued after the previous row has been closed. The minimum time interval between two successive ACTIVE commands on the same bank is defined by  $t_{RC}$ . The subsequent ACTIVE command to another bank can be issued while the first bank is being accessed, which results in a reduction of total row-access overhead. The minimum time interval between two successive ACTIVE commands on different banks is defined by  $t_{RRD}$ . These rows remain active (or open) for accesses until a PRECHARGE command is issued to that bank. A PRECHARGE command must be issued before opening a different row in the same bank.



**READ**

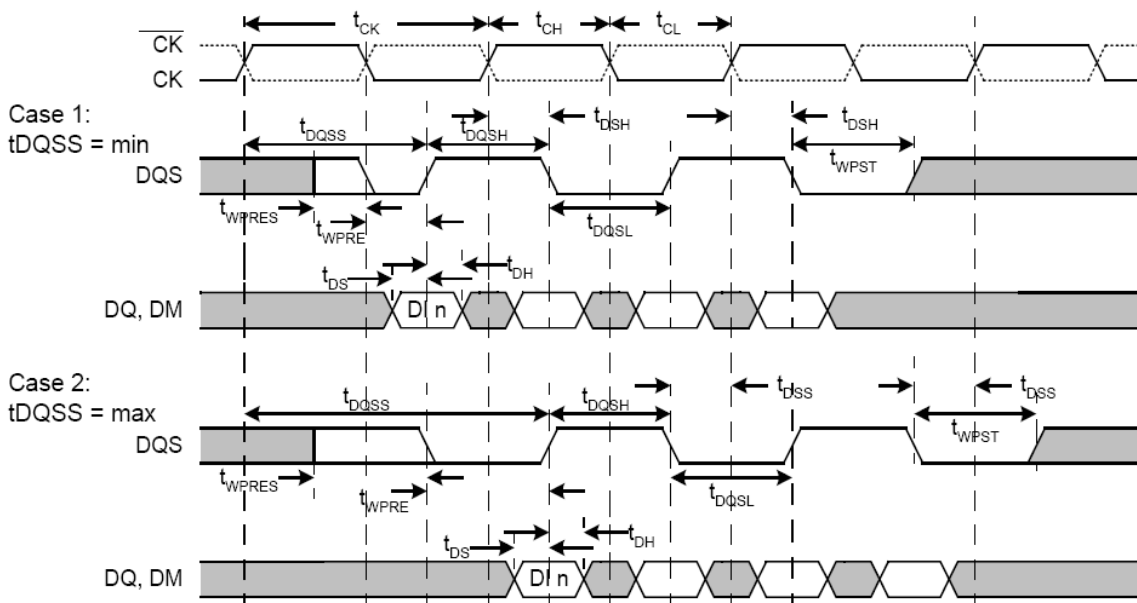
The READ command is used to initiate a burst read access to an active row, with a burst length as set in the Mode Register. BA0 and BA1 select the bank, and the address inputs select the starting column location. The value of A10 determines whether or not auto precharge is used. If auto precharge is selected, the row being accessed will be precharged at the end of the READ burst; if auto precharge is not selected, the row will remain open for subsequent accesses. During Read bursts, DQS is driven by the LPDDR SDRAM along with the output data. The initial Low state of the DQS is known as the read preamble; the Low state coincident with last data-out element is known as the read postamble. The first data-out element is edge aligned with the first rising edge of DQS and the successive data-out elements are edge aligned to successive edges of DQS.



- 1) DO n = Data Out from column n
  - 2) All DQ are valid  $t_{AC}$  after the CK edge.  
All DQ are valid  $t_{DQSQ}$  after the DQS edge, regardless of  $t_{AC}$
- = Don't Care

**WRITE**

The WRITE command is used to initiate a burst write access to an active row, with a burst length as set in the Mode Register. BA0 and BA1 select the bank, and the address inputs select the starting column location. The value of A10 determines whether or not auto precharge is used. If auto precharge is selected, the row being accessed will be precharged at the end of the WRITE burst; if auto precharge is not selected, the row will remain open for subsequent accesses. Input data appearing on the DQs is written to the memory array subject to the DM input logic level appearing coincident with the data. If a given DM signal is registered LOW, the corresponding data will be written to memory; if the DM signal is registered HIGH, the corresponding data inputs will be ignored, and a WRITE will not be executed to that byte/column location. During Write bursts, the first valid data-in element will be registered on the first rising edge of DQS following the WRITE command, and the subsequent data elements will be registered on successive edges of DQS. The Low state of DQS between the WRITE command and the first rising edge is called the write preamble; the Low state on DQS following the last data-in element is called write postamble.



- 1) DI n = Data In for column n
  - 2) 3 subsequent elements of Data In are applied in the programmed order following DI n.
  - 3) tDQSS: each rising edge of DQS must fall within the ± 25% window of the corresponding positive clock edge.
- = Don't Care

## PRECHARGE

The PRECHARGE command is used to deactivate the open row in a particular bank or the open row in all banks. The bank(s) will be available for a subsequent row access a specified time ( $t_{RP}$ ) after the PRECHARGE command is issued. Input A10 determines whether one or all banks are to be precharged. In case where only one bank is to be precharged, inputs BA0, BA1 select the bank. Otherwise BA0, BA1 are treated as “Don’t Care”. Once a bank has been precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued. A PRECHARGE command will be treated as a NOP if there is no open row in that bank, or if the previously open row is already in the process of precharging.

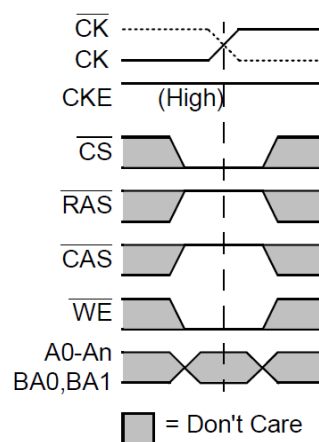
## AUTO PRECHARGE

Auto Precharge is a feature which performs the same individual bank precharge function as described above, but without requiring an explicit command. This is accomplished by using A10 (A10 = High), to enable Auto Precharge in conjunction with a specific READ or WRITE command. A precharge of the bank / row that is addressed with the READ or WRITE command is automatically performed upon completion of the read or write burst. Auto Precharge is non persistent in that it is either enabled or disabled for each individual READ or WRITE command.

Auto Precharge ensures that a precharge is initiated at the earliest valid stage within a burst. The user must not issue another command to the same bank until the precharging time ( $t_{RP}$ ) is completed. This is determined as if an explicit PRECHARGE command was issued at the earliest possible time, as described for each burst type in the Operation section of this specification.

## BURST TERMINATE

The BURST TERMINATE command is used to truncate read bursts with auto precharge disabled. The most recently registered READ command prior to the BURST TERMINATE command will be truncated. The BURST TERMINATE command is not bank specific, and should not be used to terminate write bursts.

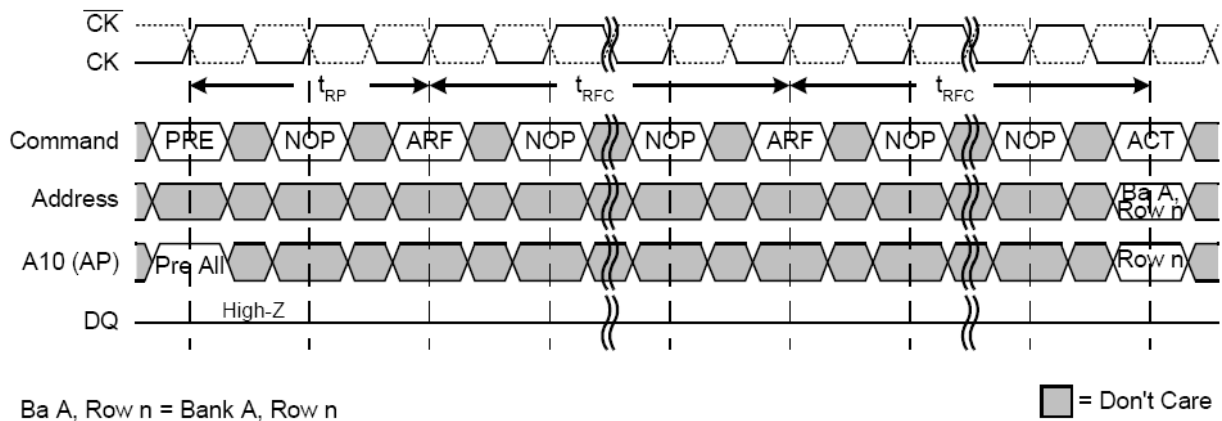


**REFRESH**

LPDDR SDRAM devices require a refresh of all rows in any rolling 64ms interval. Each refresh is generated in one of two ways: by an explicit AUTO REFRESH command, or by an internally timed event in SELF REFRESH mode:

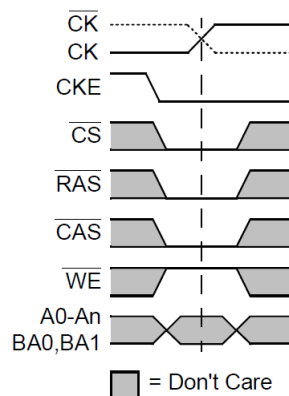
**- AUTO REFRESH**

AUTO REFRESH command is used during normal operation of the LPDDR SDRAM, and it's non-persistent, so it must be issued each time a refresh is required. The refresh addressing is generated by the internal refresh controller. The address bits become "Don't Care" during AUTO REFRESH. The LPDDR SDRAM requires AUTO REFRESH commands at an average periodic interval of  $t_{REFI}$ . To provide improved efficiency in scheduling and switching between tasks, some flexibility in the absolute interval is provided. The auto refresh period begins when the AUTO REFRESH command is registered and ends  $t_{RFC}$  later.



**- SELF REFRESH**

SELF REFRESH command can be used to retain data in the LPDDR SDRAM, even if the rest of the system is powered down. When in the self refresh mode, the LPDDR SDRAM retains data without external clocking. The LPDDR SDRAM device has a built-in timer to accommodate Self Refresh operation. The SELF REFRESH command is initiated like an AUTO REFRESH command, except CKE is LOW. Input signals except CKE are "Don't Care" during Self Refresh. The user may halt the external clock one clock after the SELF REFRESH command is registered.

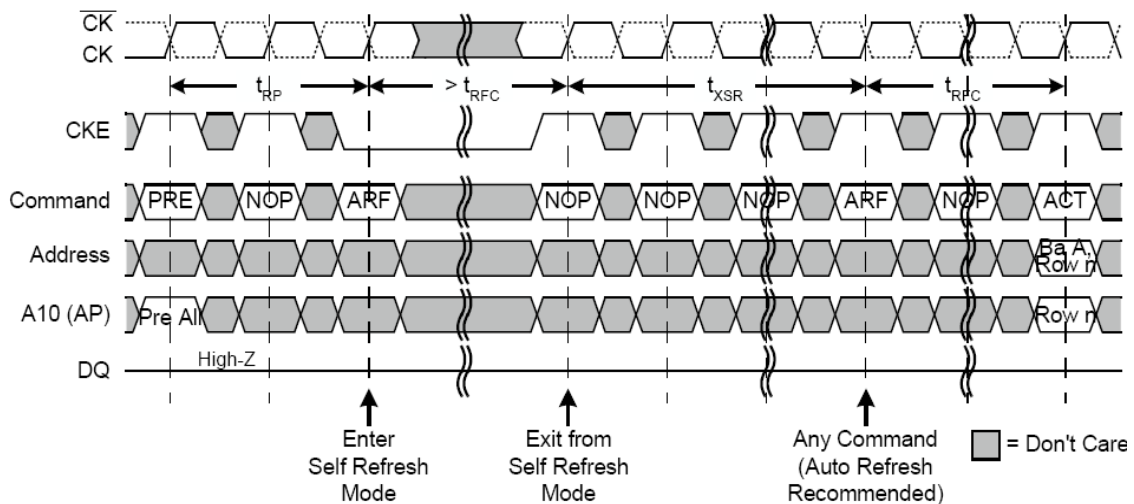


Once the SELF REFRESH command is registered, the external clock can be halted after one clock later. CKE must be held low to keep the device in Self Refresh mode, and internal clock also disabled to save power. The minimum time that the device must remain in Self Refresh mode is  $t_{RFC}$ .

In the Self Refresh mode, two additional power-saving options exist: Temperature Compensated Self Refresh and Partial Array Self Refresh. During this mode, the device is refreshed as identified in the extended mode register. An internal temperature sensor will adjust the refresh rate to optimize device power consumption while ensuring data integrity. During SELF REFRESH operation, refresh intervals are scheduled internally and may vary. These refresh intervals may be different than the specified  $t_{REFI}$  time. For this reason, the SELF REFRESH command must not be used as a substitute for the AUTO REFRESH command.

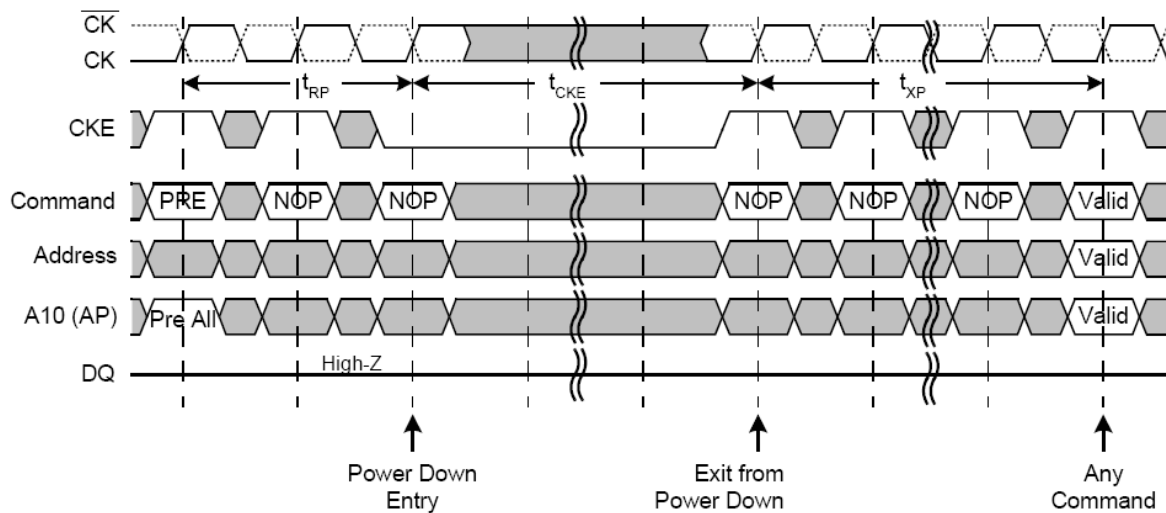
The procedure for exiting SELF REFRESH requires a sequence of commands. First, CK must be stable prior to CKE going back HIGH. When CKE is HIGH, the LPDDR SDRAM must have NOP commands issued for  $t_{XSR}$  time.

The use of Self Refresh mode introduces the possibility that an internally timed refresh event can be missed when CKE is raised for exit from Self Refresh mode. Upon exit from Self Refresh an extra AUTO REFRESH command is recommended.



### Power-Down

Power-down is entered when CKE is registered Low (no accesses can be in progress). If power-down occurs when all banks are idle, this mode is referred to as precharge power-down; if power-down occurs when there is a row active in any bank, this mode is referred to as active power-down. Power-down mode deactivates the input and output buffers, excluding CK,  $\overline{CK}$  and CKE. CKE keep Low to maintain device in the power-down mode, and all other inputs signals are "Don't Care". The minimum power-down duration is specified by  $t_{CKE}$ . The device can not stay in this mode for longer than the refresh requirements of the device, without losing data. The power-down state is synchronously existed when CKE is registered High (along with a NOP or DESELECT command). A valid command can be issued after  $t_{XP}$  after exist from power-down.



Precharge Power-Down mode shown: all banks are idle and  $t_{RP}$  is met when Power-Down Entry command is issued

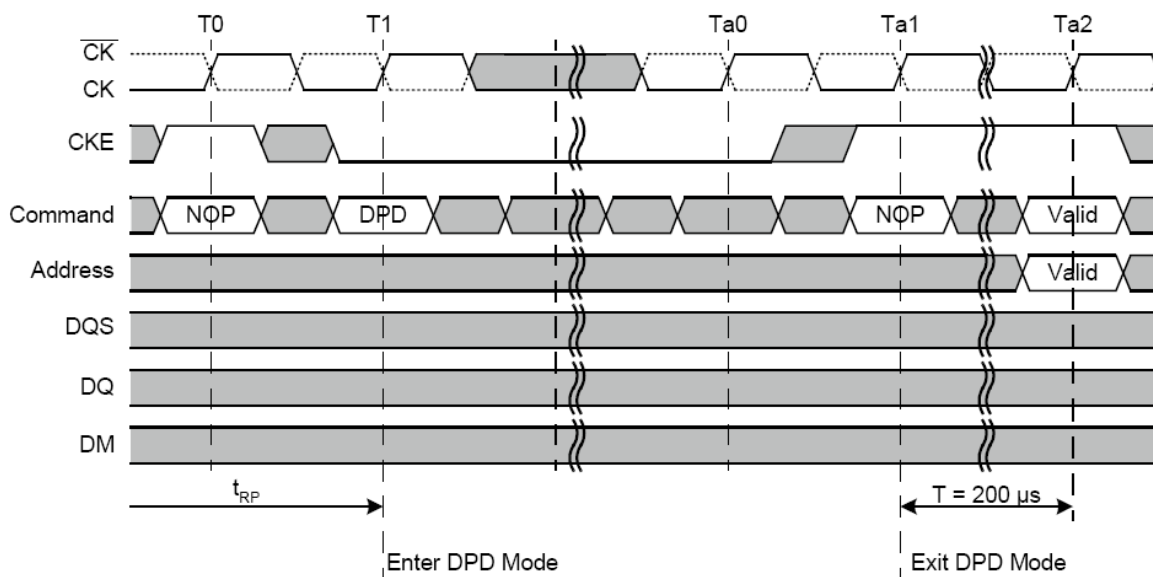
■ = Don't Care

### Deep-Power-Down

The Deep Power-Down (DPD) mode enables very low standby currents. All internal voltage generators inside the LPDDR SDRAM are stopped and all memory data is lost in this mode. All the information in the Mode Register and the Extended Mode Register is lost.

Deep Power-Down is entered using the BURST TERMINATE command except that CKE is registered Low. All banks must be in idle state with no activity on the data bus prior to entering the DPD mode. While in this state, CKE must be held in a constant Low state.

To exit the DPD mode, CKE is taken high after the clock is stable and NOP commands must be maintained for at least 200  $\mu$ s. After 200  $\mu$ s a complete re-initialization is required following steps 4 through 11 as defined for the initialization sequence.



- 1) Clock must be stable before exiting Deep Power-Down mode. That is, the clock must be cycling within specifications by Ta0
  - 2) Device must be in the all banks idle state prior to entering Deep Power-Down mode
  - 3) 200 $\mu$ s is required before any command can be applied upon exiting Deep Power-Down mode
  - 4) Upon exiting Deep Power-Down mode a PRECHARGE ALL command must be issued, followed by two AUTO REFRESH commands and a load mode register sequence
- = Don't Care



### Clock Stop

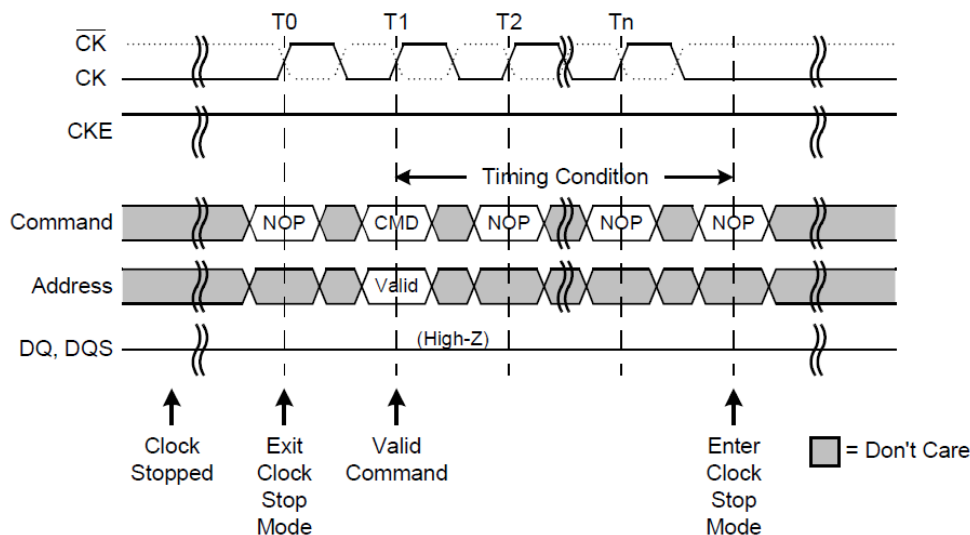
Stopping a clock during idle periods is an effective method of reducing power consumption. The LPDDR SDRAM supports clock stop mode under the following conditions:

- The last command (ACTIVE, READ, WRITE, PRECHARGE, AUTO REFRESH or MODE REGISTER SET) has executed to completion, including any data-out during read bursts; the number of clock pluses per access command depends on the device's AC timing parameters and the clock frequency;
- The related timing condition ( $t_{RCD}$ ,  $t_{WR}$ ,  $t_{RP}$ ,  $t_{RFC}$ ,  $t_{MRD}$ ) has been met;
- CKE is held High.

When all conditions have been met, the device is either in "idle state" or "row active state", and clock stop may be entered with CK held Low and  $\overline{CK}$  held High. Clock Stop mode is exited by restarting the clock. At least one NOP command has to be issued before the next access command may be applied. Additional clock pulses might be required depending on the system characteristics.

### Clock stop mode entry and exit:

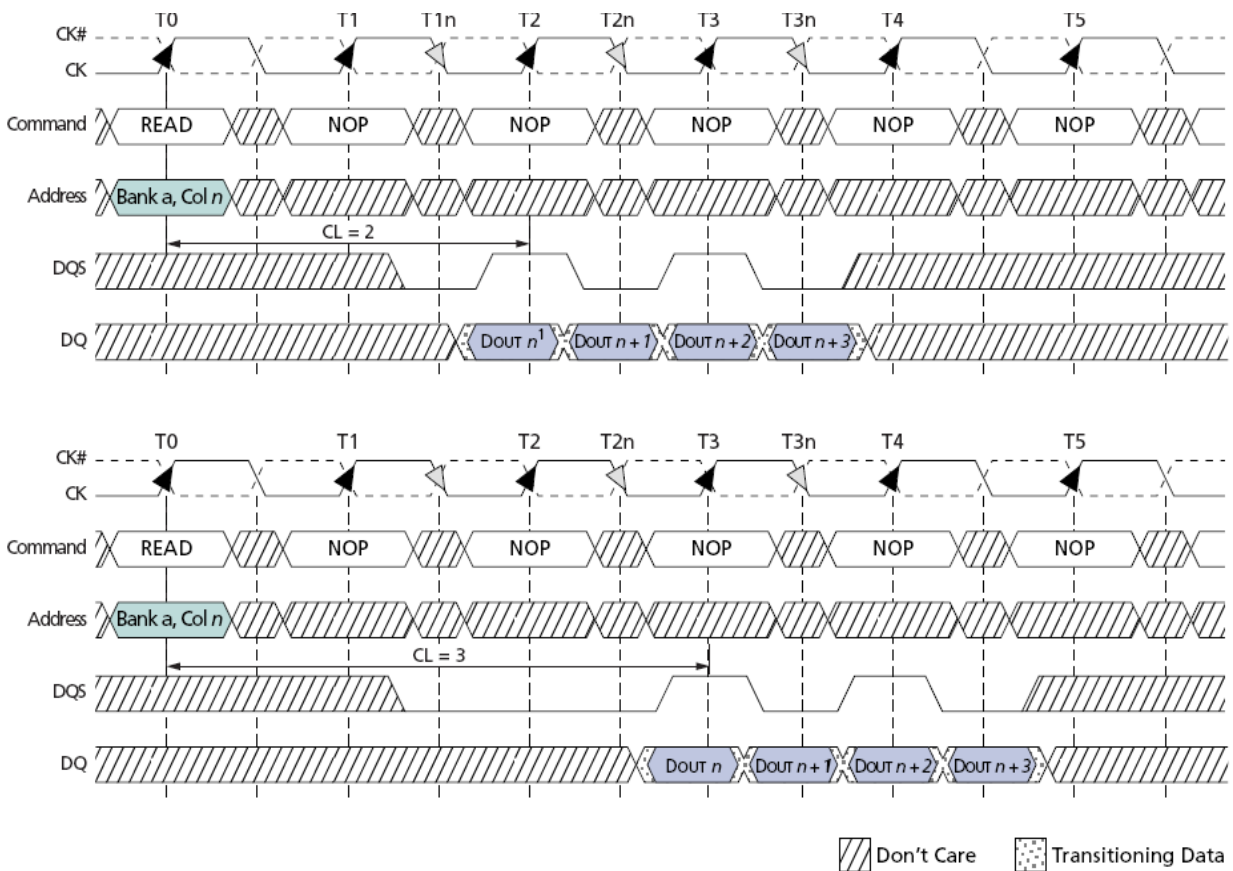
- Initially the device is in clock stop mode
- The clock is restarted with the rising edge of T0 and a NOP on the command inputs
- With T1 a valid access command is latched; this command is followed by NOP commands in order to allow for
- clock stop as soon as this access command is completed
- Tn is the last clock pulse required by the access command latched with T1
- The clock can be stopped after Tn



Timing

READs

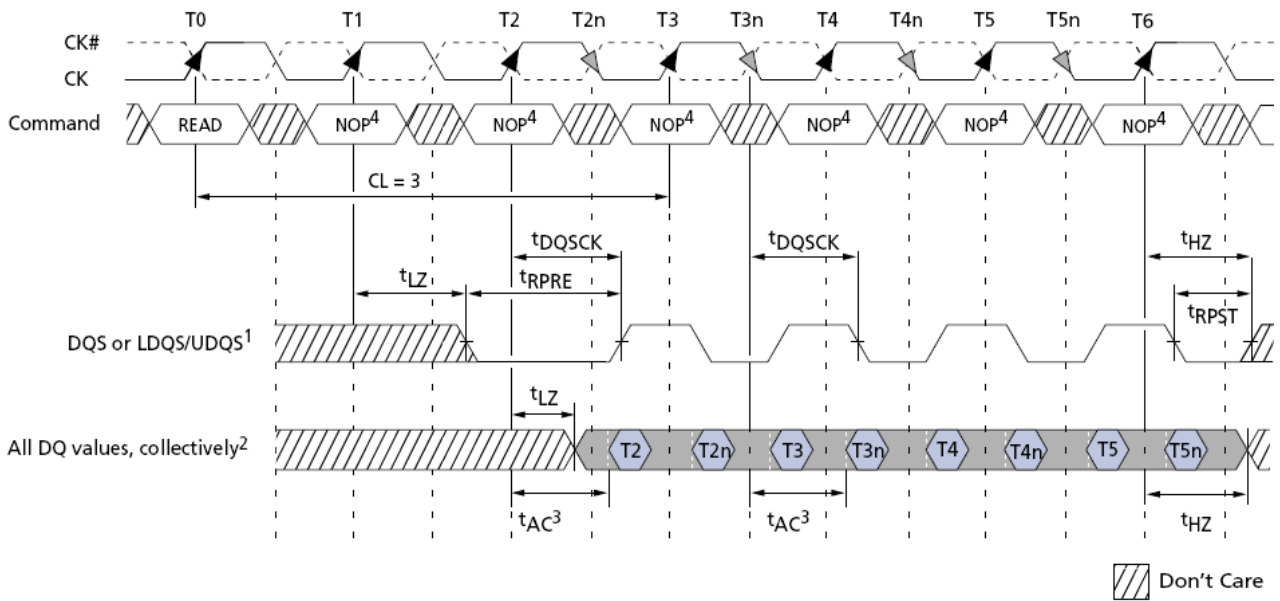
READ burst operations are initiated with a READ command. The starting column and bank addresses are provided with the READ command, and auto precharge is either enabled or disabled for that burst access. If auto precharge is enabled, the row being accessed is precharged at the completion of the burst. During READ bursts, the valid data-out element from the starting column address will be available following the CAS latency after the READ command. The first data-out element is edge aligned with the first rising edge of DQS and the successive data-out elements are edge aligned to successive edges of DQS. DQS is driven by LPDDR SDRAM along with output data. Upon completion of a read burst, assuming no other READ command has been initiated, the DQ will go to High-Z.



Read Burst Operation (BL=4, and CL=2, CL=3)

Notes:

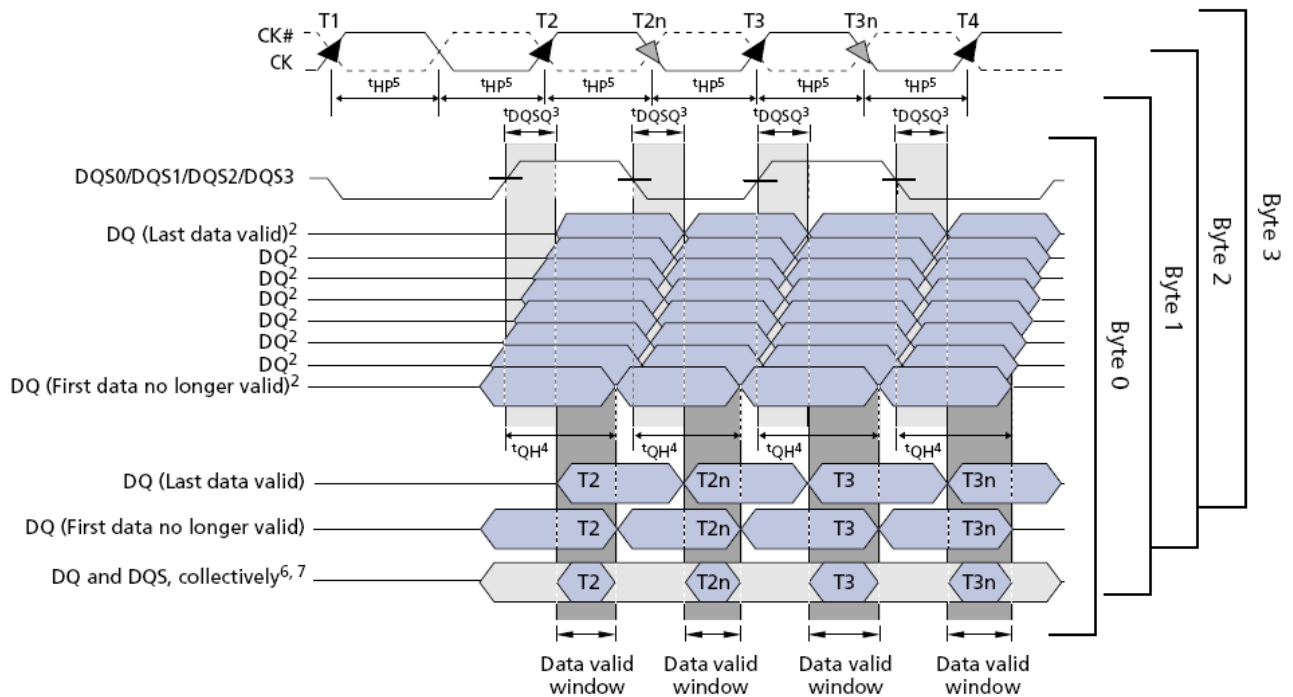
1. Dout n = data-out from column n.
2. Shown with nominal tAC, tDQSK, and tDQSQ.



### Data Output Timing – $t_{AC}$ and $t_{DQSK}$

Notes:

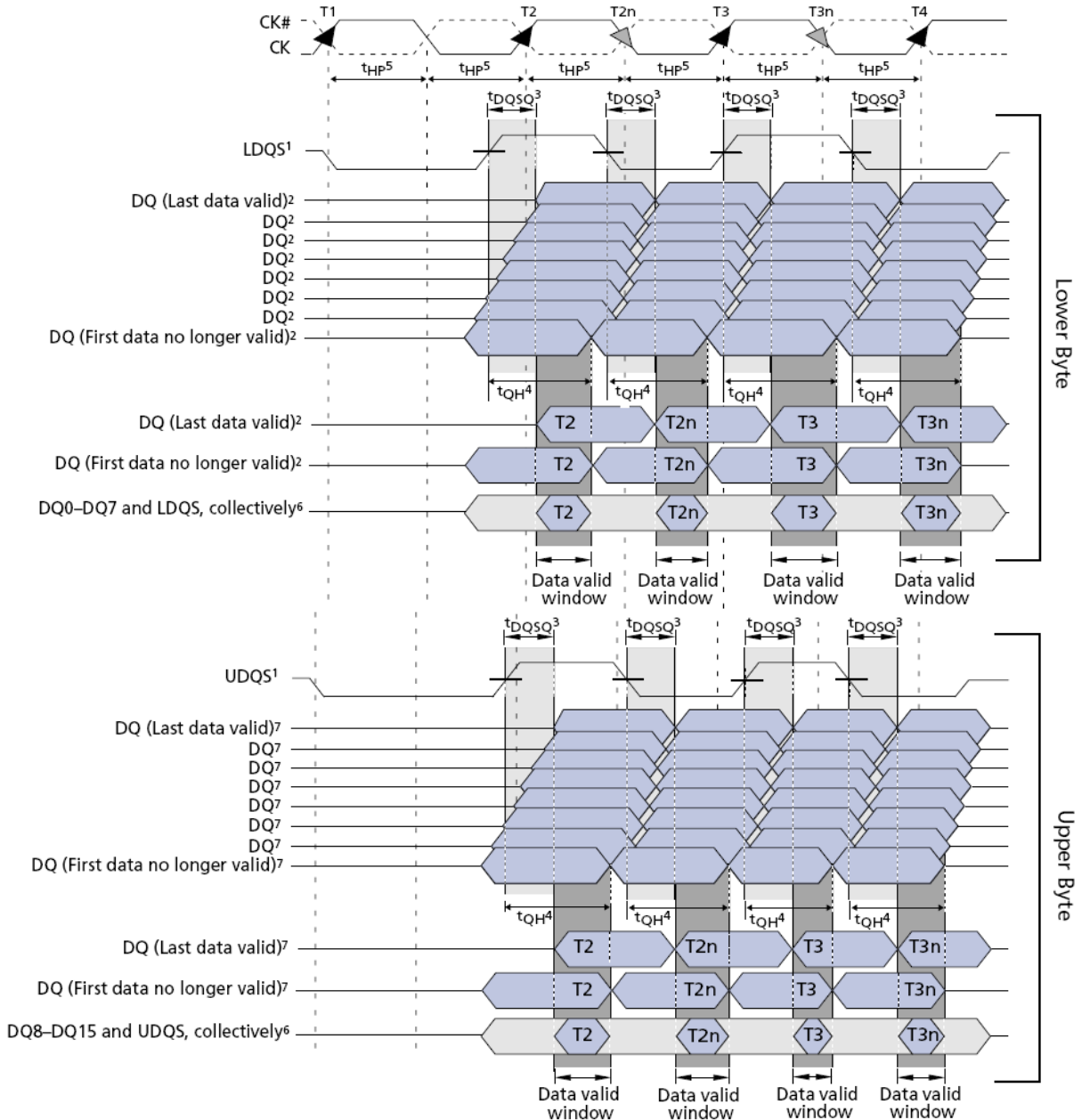
1. DQ transitioning after DQS transitions define  $t_{DQSQ}$  window.
2. All DQ must transition by  $t_{DQSQ}$  after DQS transitions, regardless of  $t_{AC}$ .
3.  $t_{AC}$  is the DQ output window relative to CK and is the "long-term" component of DQ skew.
4. Commands other than NOP may be valid during this cycle.



**Data Output Timing – tDQSQ, tQH and Data Valid Window (x32)**

Notes:

1. DQ transitioning after DQS transitions define tDQSQ window.
2. Byte 0 is DQ0-DQ7, Byte 1 is DQ8-DQ15, Byte 2 is DQ16-DQ23, and Byte 3 is DQ24-DQ31.
3. tDQSQ is derived at each DQS clock edge and is not cumulative over time and begins with DQS transition and ends with the last valid DQ transition .
4. tOH is derived from tHP,  $tOH = tHP - tOHS$ .
5. tOH is the lesser of tCL or tCH clock transition collectively when a bank is active.
6. The data valid window is derived from each DQS transition and is  $tOH - tDQSQ$ .
7. DQ[7:0] and DQS0 for byte 0; DQ[15:8] and DQS1 for byte 1; DQ[23:16] and DQS2 for byte 2; DQ[31:24] and DQS3 for byte 3.



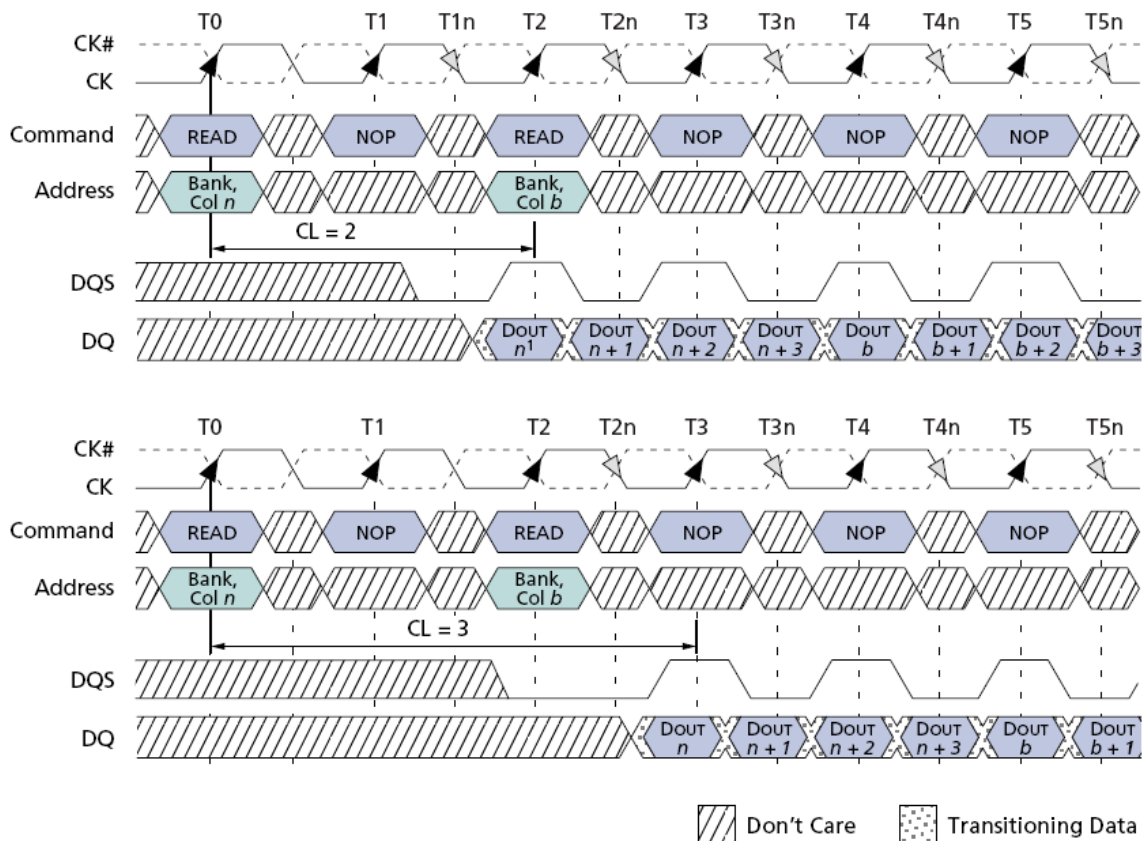
**Data Output Timing –  $t_{DQSQ}$ ,  $t_{QH}$  and Data Valid Window (x16)**

Notes:

1. DQ transitioning after DQS transitions define  $t_{DQSQ}$  window. LDQS defines the lower byte and UDQS defines the upper byte.
2. DQ0, DQ1, DQ2, DQ3, DQ4, DQ5, DQ6, or DQ7.
3.  $t_{DQSQ}$  is derived at each DQS clock edge and is not cumulative over time and begins with DQS transition and ends with the last valid DQ transition .
4.  $t_{OH}$  is derived from  $t_{HP}$ ,  $t_{OH} = t_{HP} - t_{OHS}$ .
5.  $t_{OH}$  is the lesser of  $t_{CL}$  or  $t_{CH}$  clock transition collectively when a bank is active.
6. The data valid window is derived from each DQS transition and is  $t_{OH} - t_{DQSQ}$ .
7. DQ9, DQ10, DQ11, DQ12, DQ13, DQ14, or DQ15.

### READ to READ

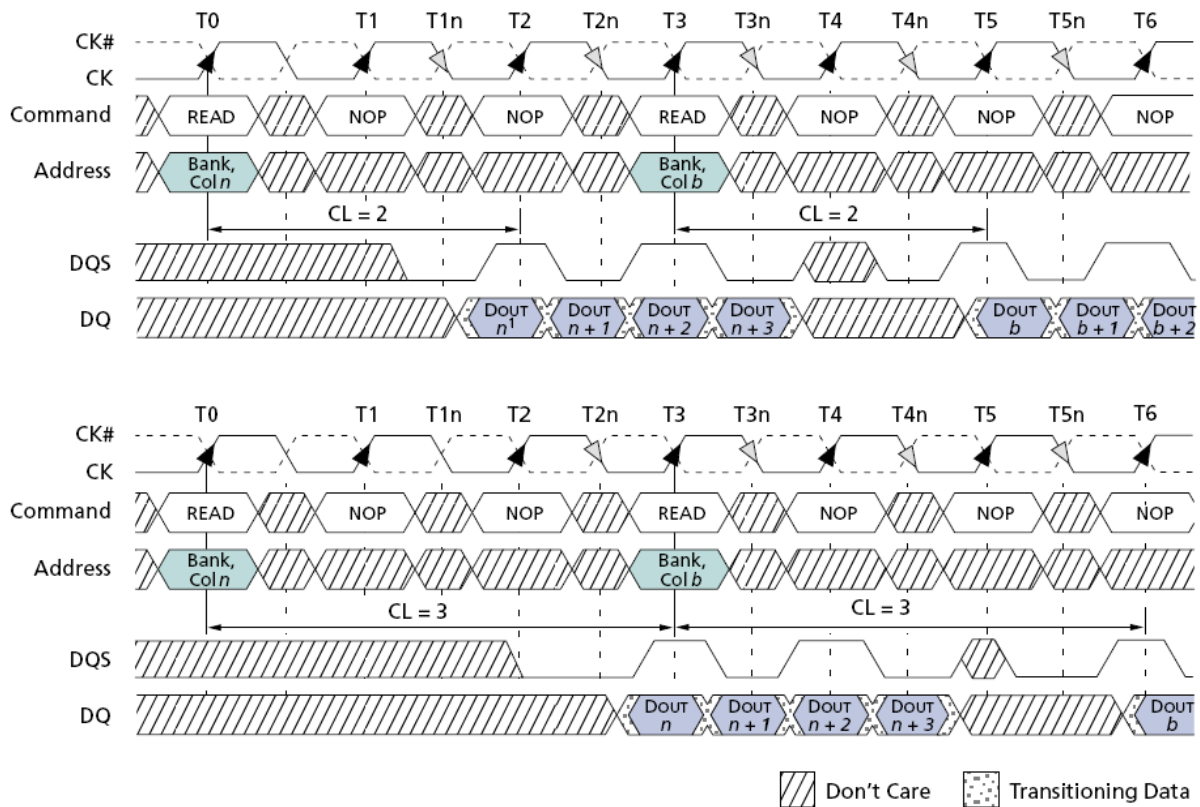
Data from a read burst may be concatenated or truncated by a subsequent READ command. The first data from the new burst follows either the last element of a completed burst or the last desired element of a longer burst that is being truncated. The new READ command should be issued X cycles after the first READ command, where X equals the number of desired data-out element pairs (pairs are required by the 2n prefetch architecture).



#### Consecutive Read Bursts (CL=2 and CL=3)

Notes:

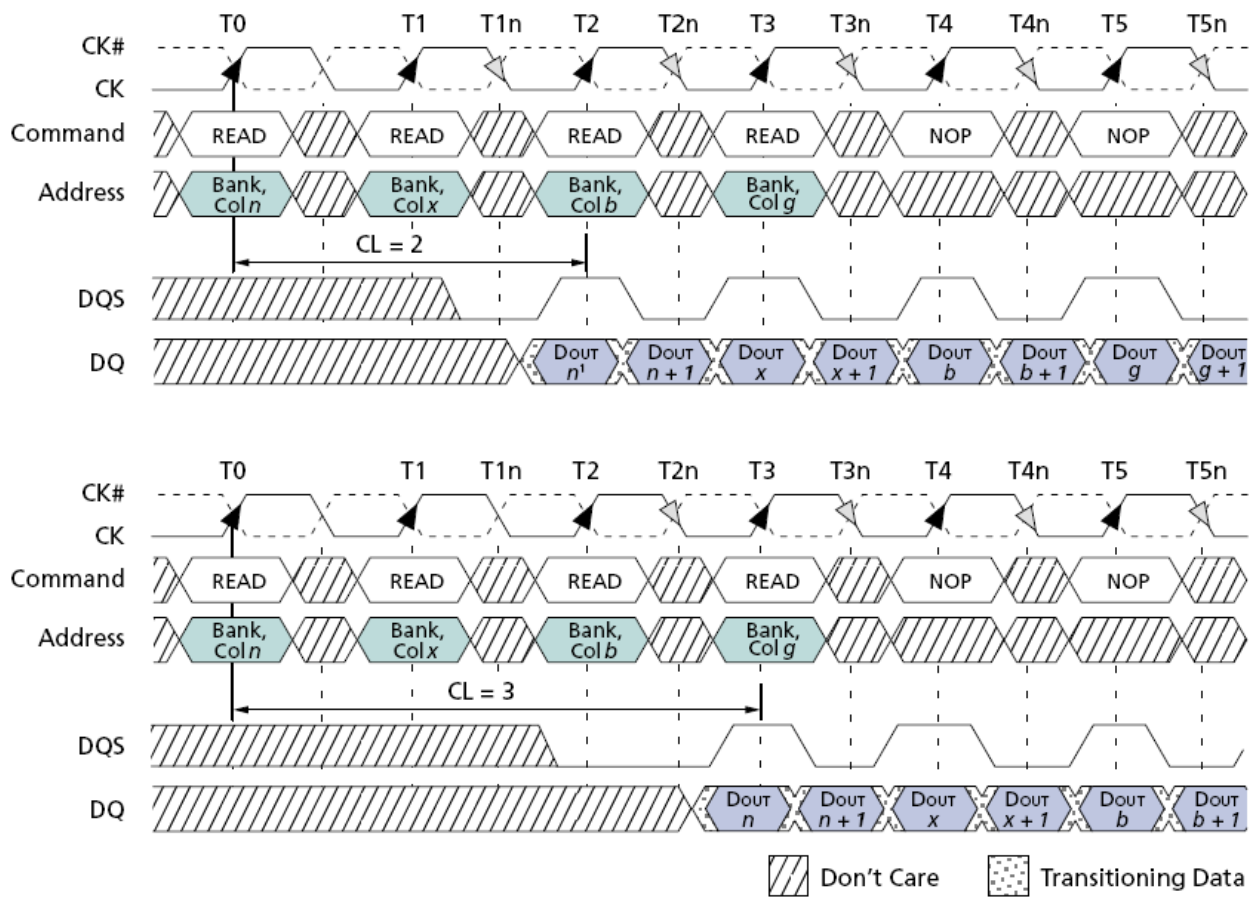
1. Dout n (or b) = data-out from column n (or column b).
2. BL = 4, 8, or 16 (if 4, the bursts are concatenated; if 8 or 16, the second burst interrupts the first).
3. Shown with nominal tAC, tDQSCk, and tDQSQ.



### Nonconsecutive Read Bursts (CL=2 and CL=3)

Notes:

1. DOUT n (or b) = data-out from column n (or column b).
2. BL = 4, 8, or 16 (if 4, the bursts are concatenated; if 8 or 16, the second burst interrupts the first).
3. Shown with nominal tAC, tDQSCK, and tDQSQ.



### Random Read Bursts (CL=2 and CL=3)

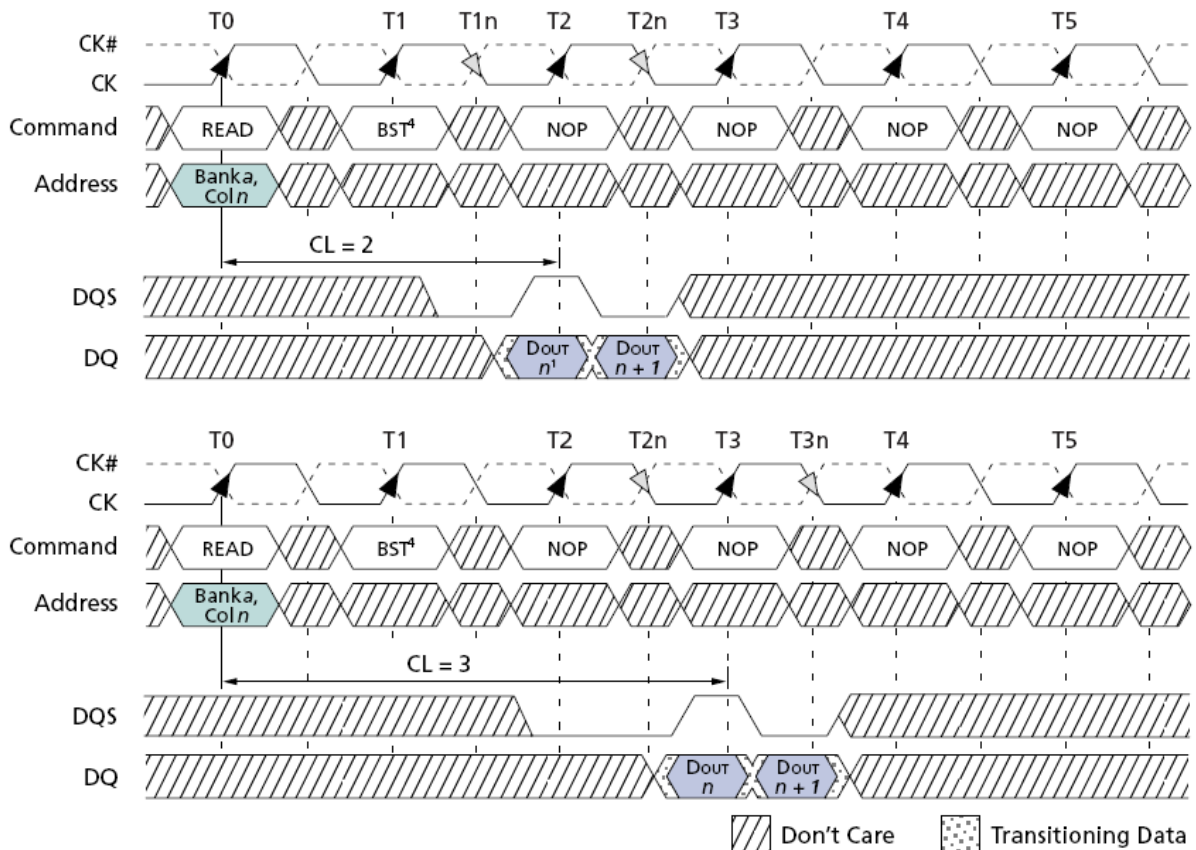
Notes:

1. Dout n (or x, b, g) = data-out from column n (or column x, column b, column g).
2. BL = 2, 4, 8, or 16 (if 4, 8 or 16, the following burst interrupts the previous).
3. Shown with nominal tAC, tDQSCk, and tDQSQ.



### READ BURST TERMINATE

Data from any READ burst may be truncated with a BURST TERMINATE command. The BURST TERMINATE latency is equal to read (CAS) latency, i.e., the BURST TERMINATE command should be issued X cycles after the READ command where X equals the desired data-out element pairs (pairs are required by the 2n-prefetch architecture).



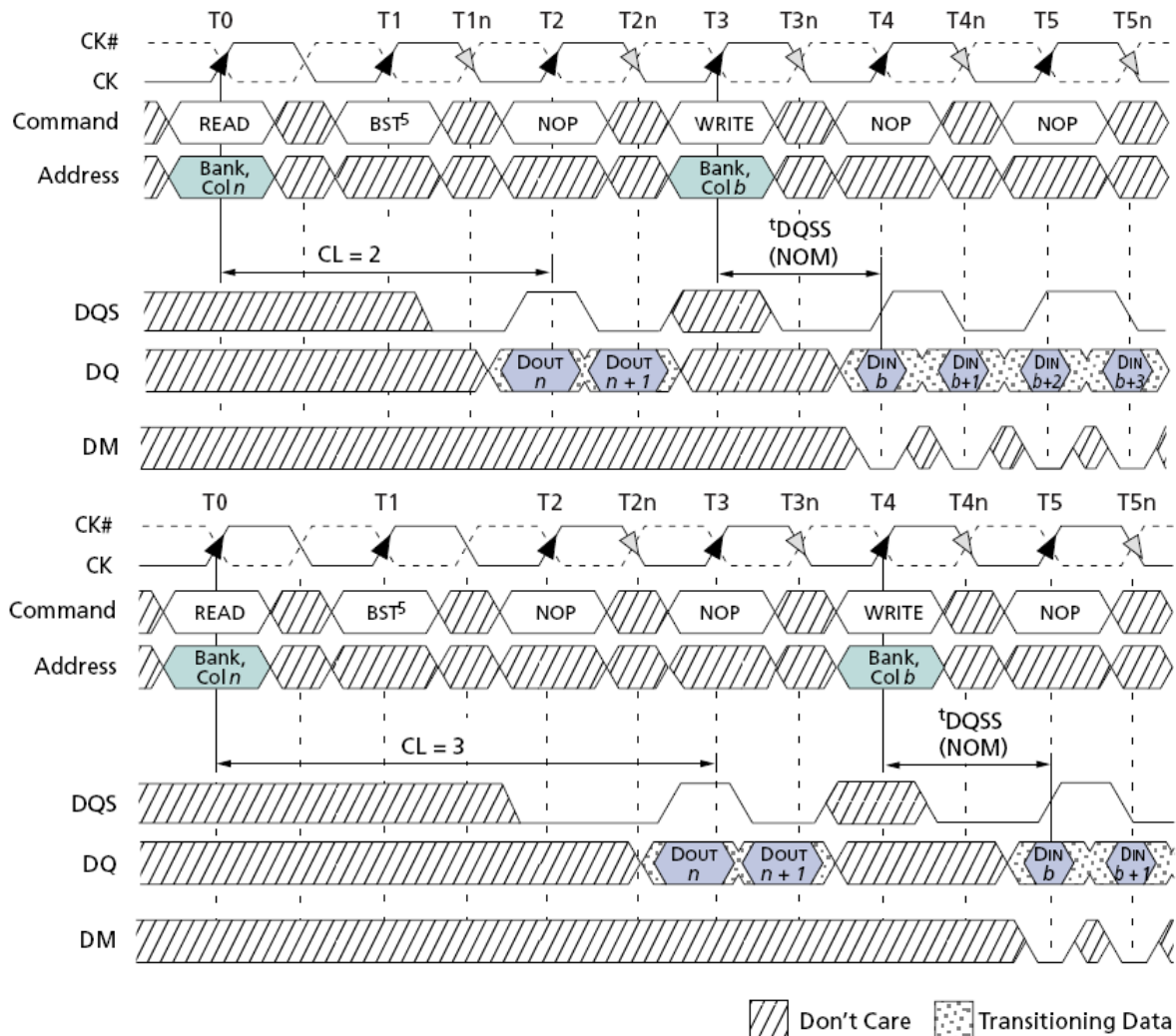
**Terminating a Read Bursts (CL=2 and CL=3)**

Notes:

1. Dout n = data-out from column n.
2. BL = 4, 8, or 16.
3. Shown with nominal tAC, tDQSCK, and tDQSQ.
4. BST = BURST TERMINATE command; page remains open.
5. CKE = HIGH.

### READ to WRITE

Data from READ burst must be completed or truncated before a subsequent WRITE command can be issued. If truncation is necessary, the BURST TERMINATE command must be used.



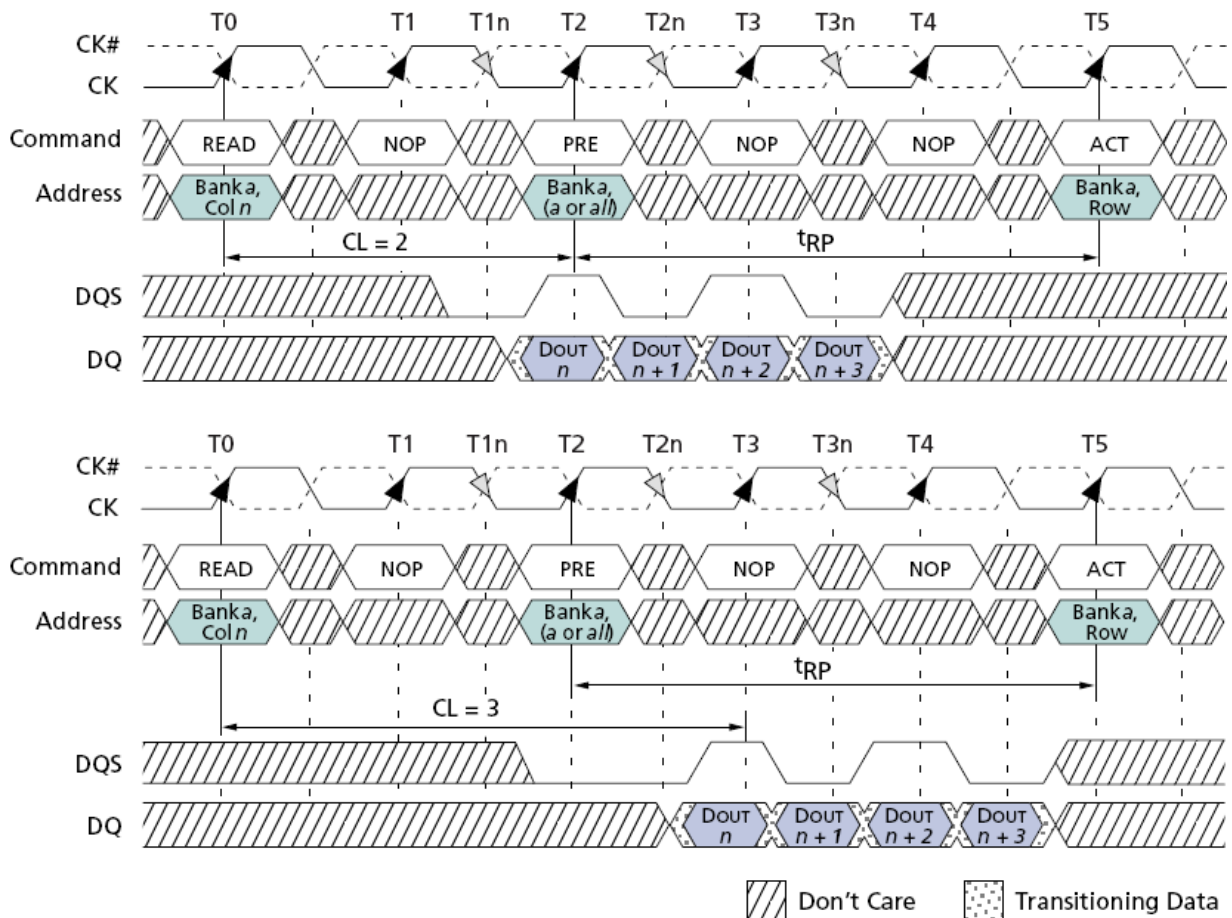
**Read to Write (CL=2 and CL=3)**

Notes:

1. Dout n = data-out from column n.
2. BL = 4, 8, or 16.
3. Shown with nominal tAC, tDQSK, and tDQSQ.
4. BST = BURST TERMINATE command; page remains open.
5. CKE = HIGH.

### READ to Precharge

A READ burst may be followed by, or truncated with, a PRECHARGE command to the same bank. The PRECHARGE command should be issued X cycles after the READ command, where X equals the number of desired data element pairs. Following the PRECHARGE command, a subsequent command to the same bank can not be issued until  $t_{RP}$  is met. Part of the row precharge time is hidden during the access of the last data element. In the case of a READ being executed to completion, a PRECHARGE command issued at optimum time provides the same operation as READ with AP. The disadvantage of PRECHARGE command is that the command and address buses be available at the appropriate time to issue the command. The advantage of the PRECHARGE command is that can be used to truncate bursts.



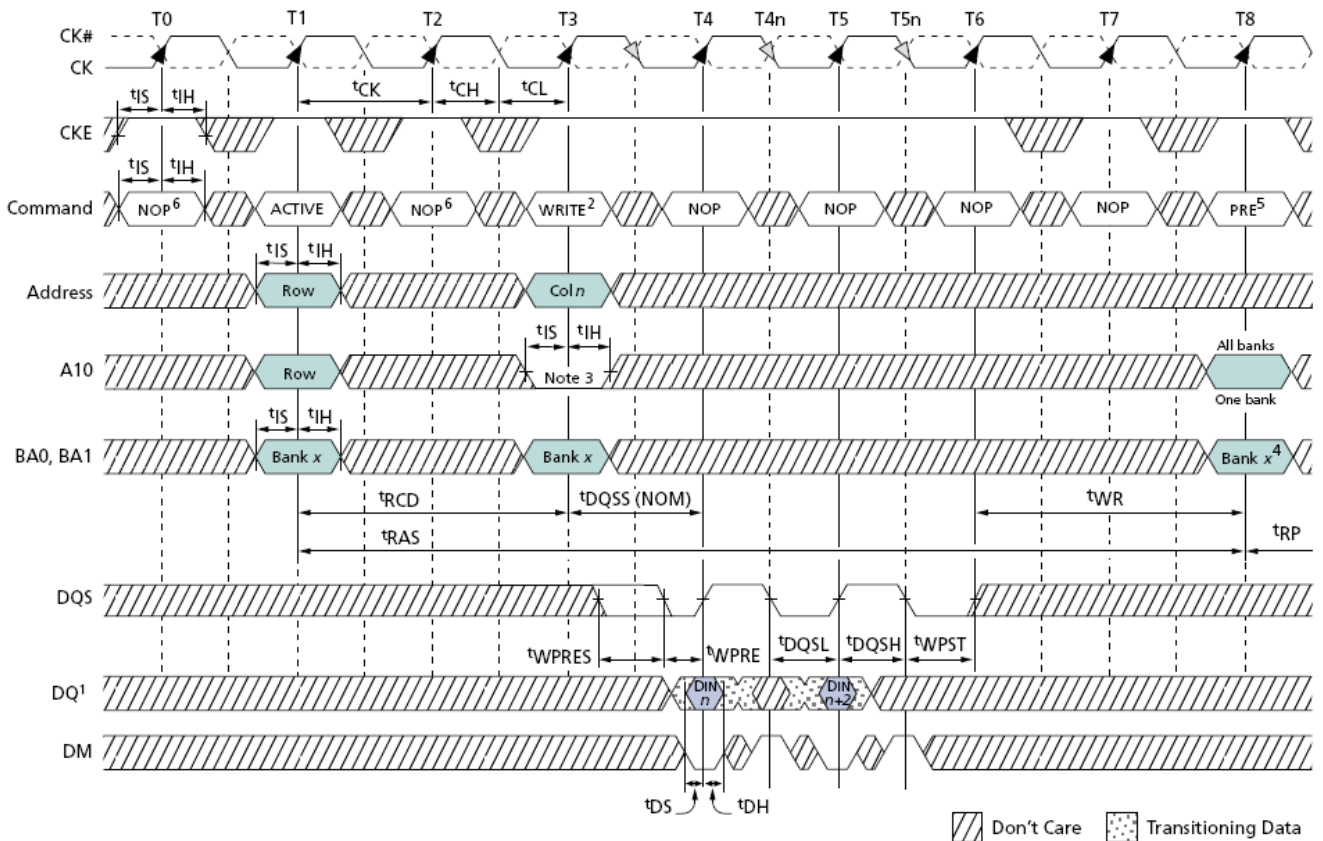
### Read to Precharge (CL=2 and CL=3)

#### Notes:

1. DOUT n = data-out from column n.
2. BL = 4, or an interrupted burst 8 or 16.
3. Shown with nominal tAC, tDQSCK, and tDQSQ.
4. READ-to-PRECHARGE equals 2 clocks, which enables 2 data pairs of data-out. A READ command with auto precharge enabled, provided tRAS (min) is met, would cause a precharge to be performed at X number of clock cycles after the READ command, where  $x = BL/2$ .
5. PRE = PRECHARGE command; ACT = ACTIVE command.

### WRITEs

WRITE burst operations are initiated with a WRITE command. The starting column and bank addresses are provided with the WRITE command, and auto precharge is either enabled or disabled for that burst access. If auto precharge is enabled, the row being accessed is precharged at the completion of the burst. During WRITE bursts, the first valid data-in element will be registered on the first rising edge of DQS following the WRITE command, and subsequent data elements will be registered on successive edges of DQS. Input data appearing on the data bus is written to the memory array subject to the state of the data mask DM inputs coincident with the data.



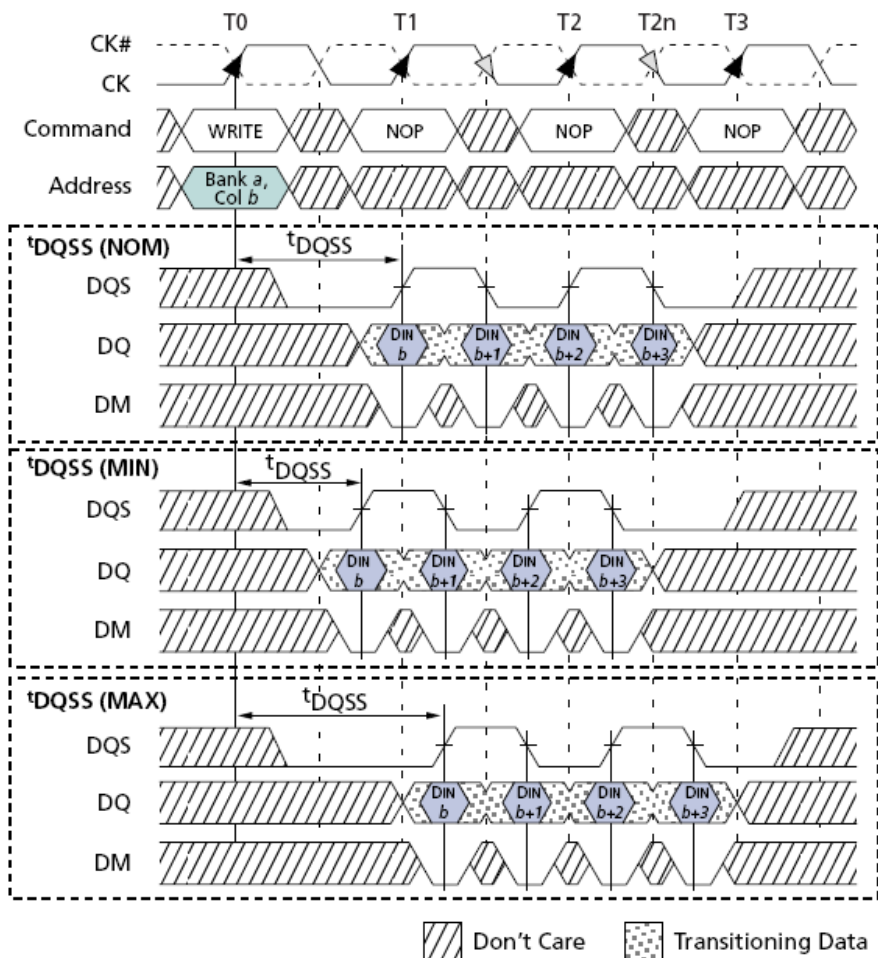
**Write – DM Operation (CL=2 and CL=3)**

Notes:

1. Din n = data-in from column n.
2. BL = 4 in the case shown.
3. Disable auto precharge.
4. Bank x at T8 is "Don't Care", if A10 is HIGH at T8.
5. PRE = PRECHARGE command.
6. NOP commands are shown for ease of illustration; other commands may be valid at these time.

### WRITE Burst

The time between the WRITE command and the first corresponding rising edge of DQS ( $t_{DQSS}$ ) is specified with a relatively wide range (from 75% to 125% of one clock cycle). All of the WRITE diagrams show the nominal case, and where the two extreme cases (that is,  $t_{DQSS}(\min)$  and  $t_{DQSS}(\max)$ ) might not be intuitive, they have also been included. Upon completion of the burst, assuming no other commands have been initiated, the DQs will remain High-Z and any additional input data will be ignored.



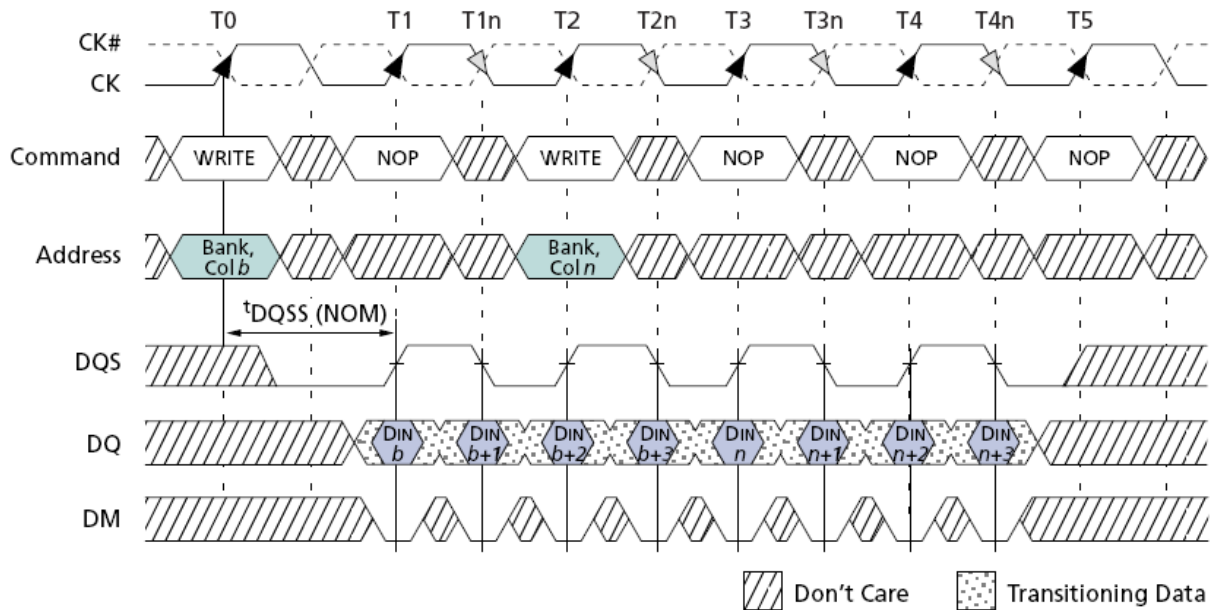
Write burst (nominal,  $t_{DQSS}(\min)/(\max)$ , BL=4)

Notes:

1. Din b = data-in from column b.
2. An uninterrupted burst of 4 is shown.
3. A10 is LOW with the WRITE command (Auto Precharge is disabled).

### WRITE to WRITE

Data for any WRITE burst may be concatenated with or truncated by a subsequent WRITE command. In either case, a continuous flow input data can be maintained. The new WRITE command can be issued on any positive edge of the clock following the previous WRITE command. The first data-in element from the new burst is applied after either the last element of a completed burst or the last desired data element of the longer burst which is being truncated. The new WRITE command should be issued X cycles after the first WRITE command, where X equals the number of desired data-in element pairs (pairs are required by the 2n-prefetch architecture).

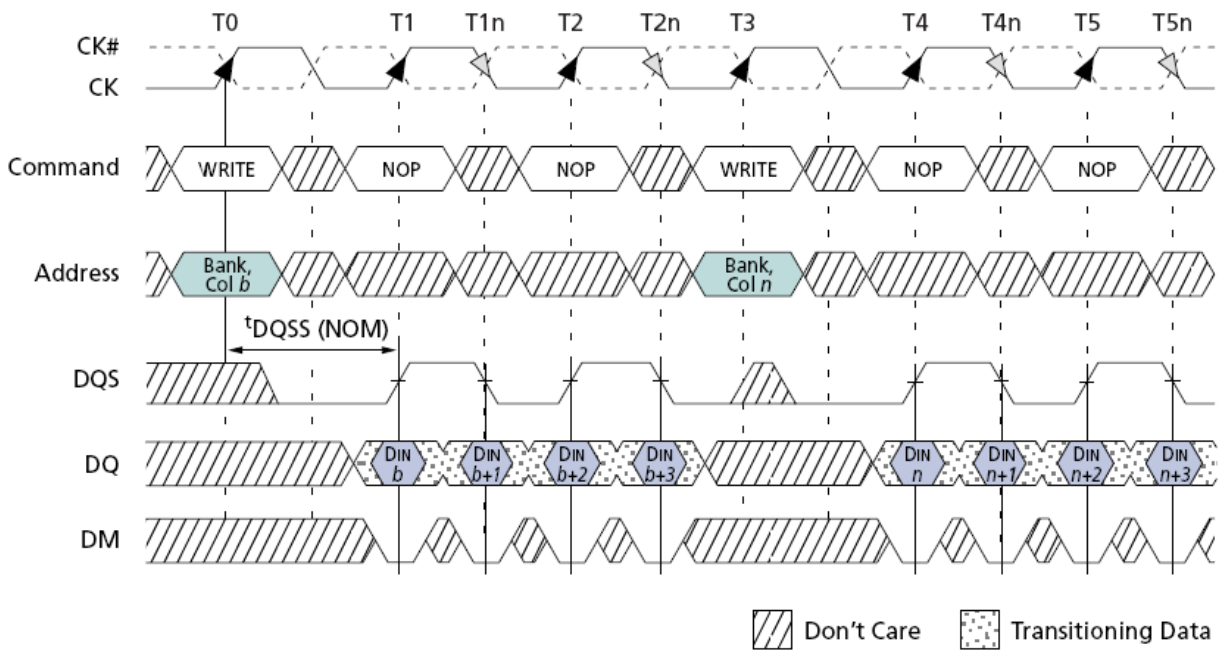


**Consecutive WRITE-to-WRITE (BL=4)**

**Notes:**

1. Din b (n) = data-in from column b (n).
2. An uninterrupted burst of 4 is shown.
3. Each WRITE command may be to any bank.



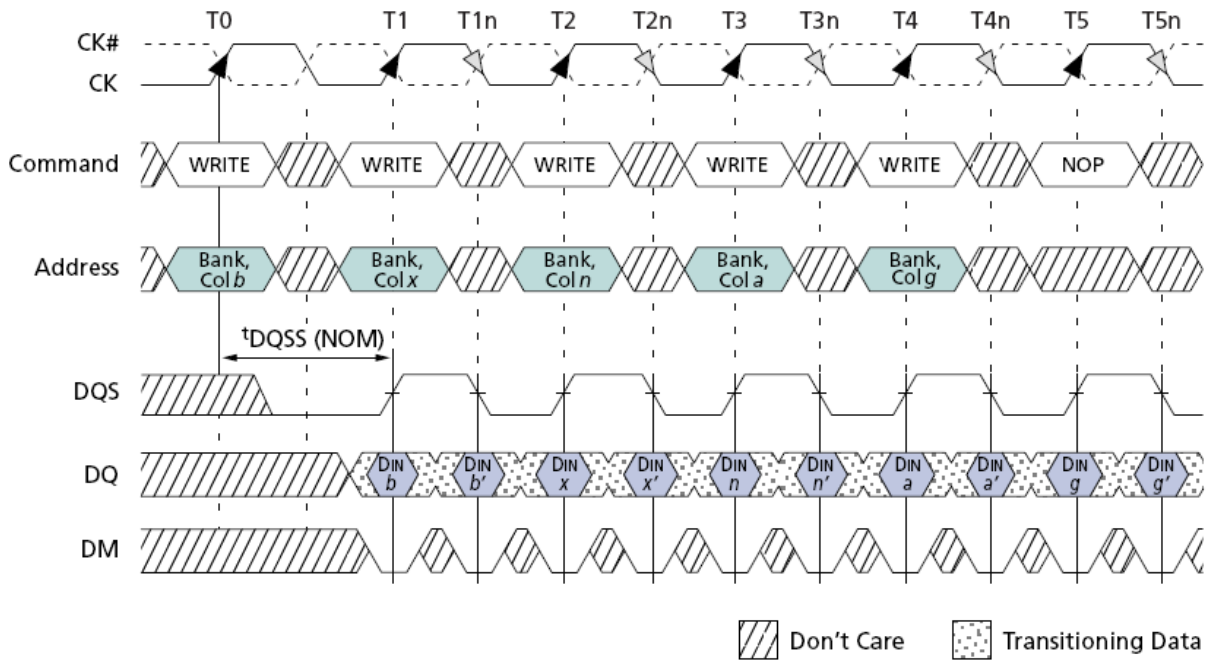


### Nonconsecutive WRITE-to-WRITE (BL=4)

Notes:

1. Din b (n) = data-in from column b (n).
2. An uninterrupted burst of 4 is shown.
3. Each WRITE command may be to any bank.





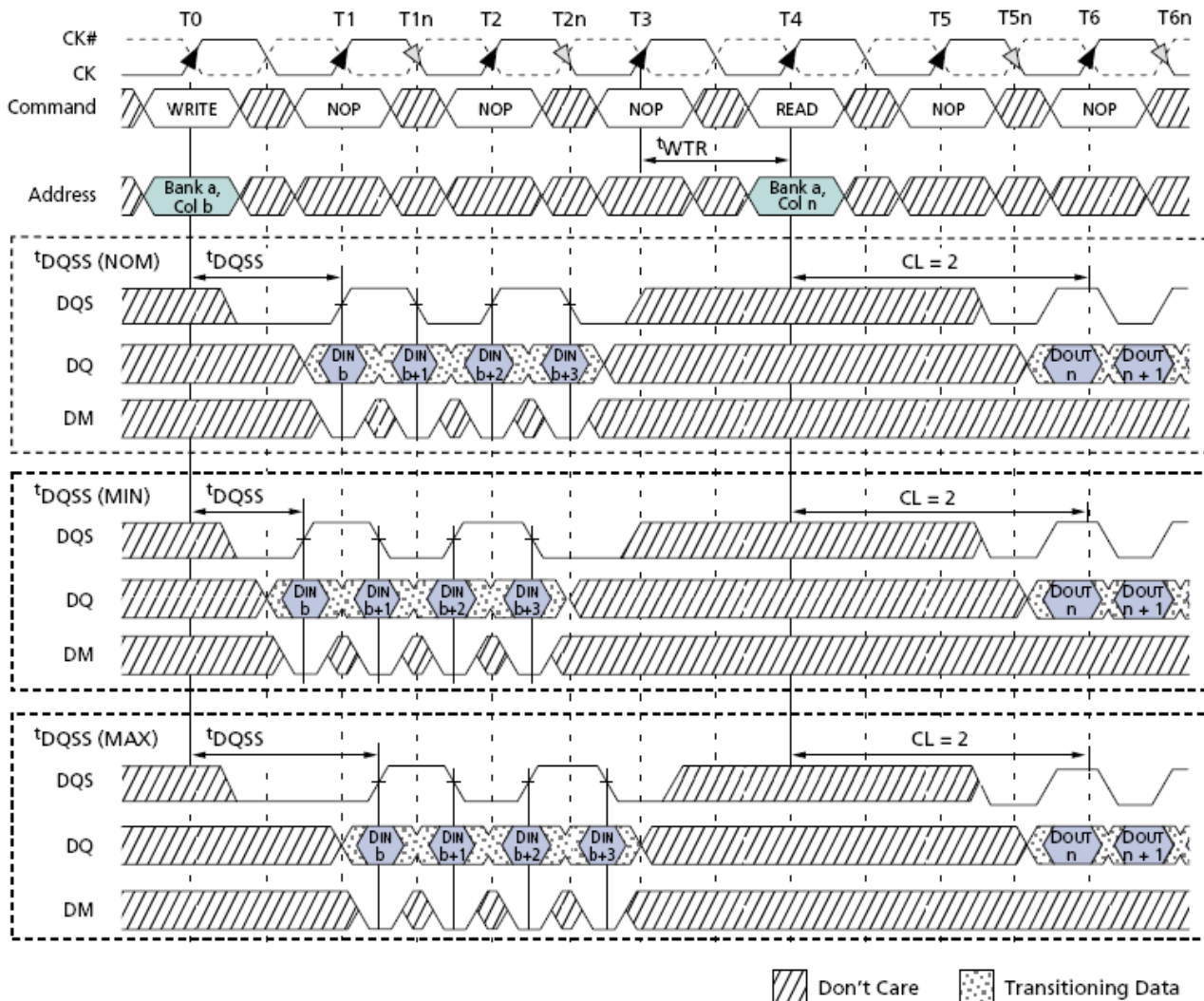
### Random Write Cycles

Notes:

1. Din b (or x, n, a, g) = data-in from column b (or x, n, a, g).
2. b' (or x', n', a', g') = the next data-in following Din b (x, n, a, g) according to the programmed burst order.
3. Programmed BL = 2, 4, 8, or 16 in cases shown.
4. Each WRITE command may be to any bank.

### WRITE to READ

Data for any Write burst may be followed by a subsequent READ command. To follow a Write without truncating the write burst,  $t_{WTR}$  should be met as shown in Figure.

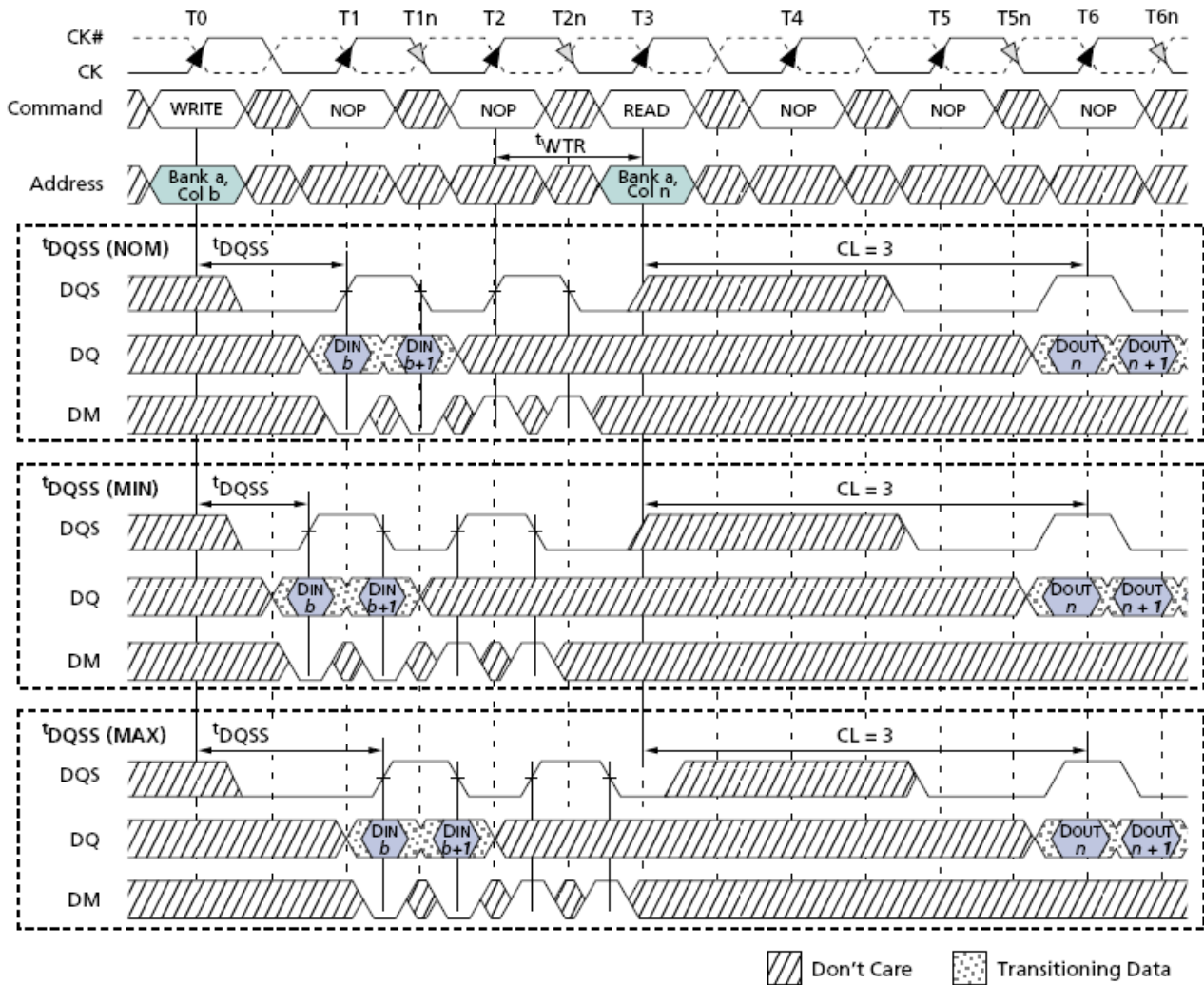


**Non-Interrupting Write-to-Read (nominal,  $t_{DQSS}(\min)/(\max)$ , BL=4)**

Notes:

1. Din b = data-in from column b; Dout n = data-out for column n.
2. An uninterrupted burst of 4 is shown.
3.  $t_{WTR}$  is referenced from the first positive CK edge after the last data-in pair.
4. The READ and WRITE commands are to the same device. However, the READ and WRITE commands may be to different devices. In which case  $t_{WTR}$  is not required and the READ command could be applied earlier.
5. A10 is LOW with the WRITE command (auto precharge is disabled).

Data for any Write burst may be truncated by a subsequent READ command as shown in Figure. Note that the only data-in pairs that are registered prior to the  $t_{WTR}$  period are written to the internal array, and any subsequent data-in must be masked with DM.



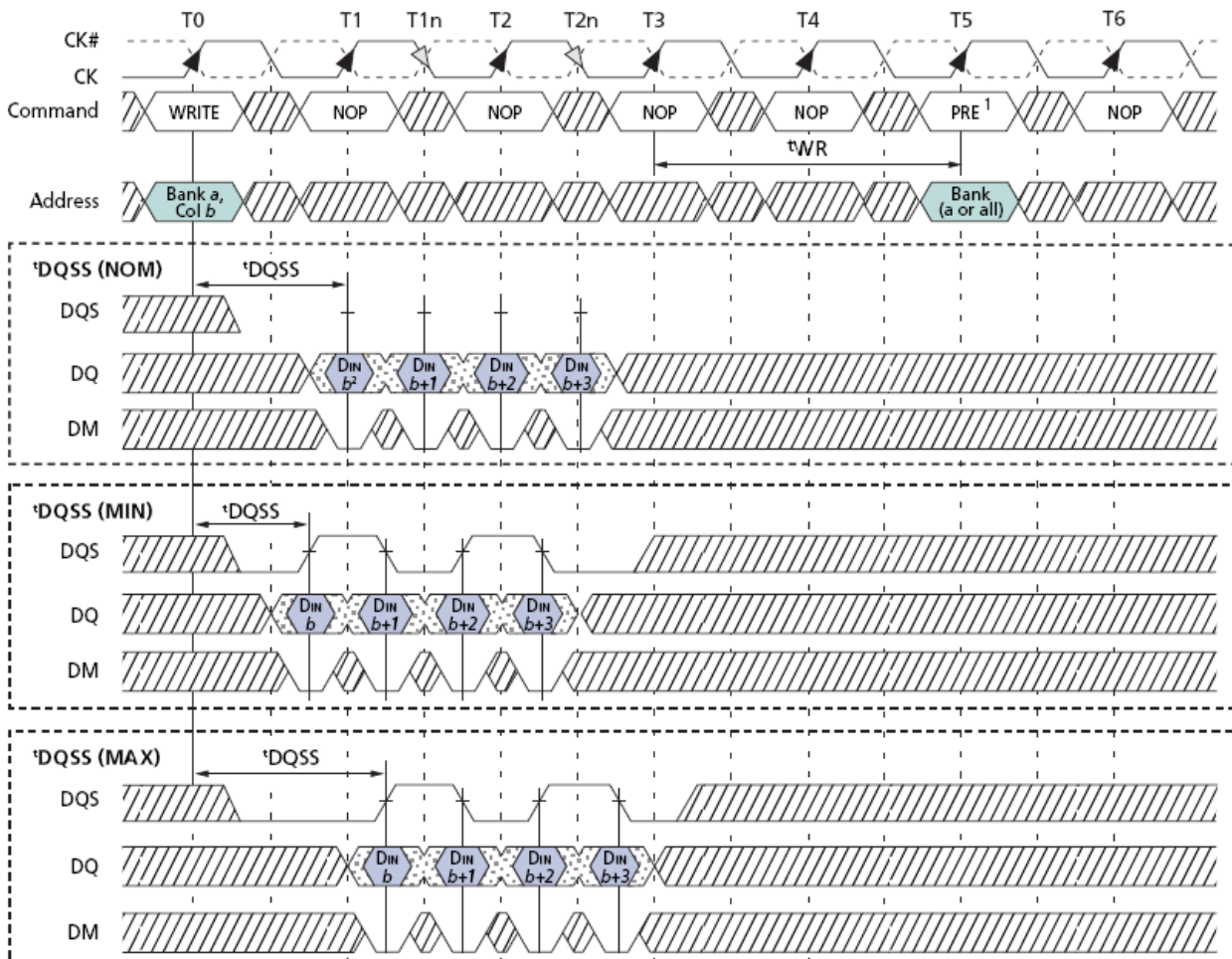
**Interrupting Write-to-Read (nominal,  $t_{DQSS}(\min)$ / $(\max)$ , BL=4)**

Notes:

1. Din b = data-in from column b; Dout n = data-out for column n.
2. An uninterrupted burst of 4 is shown; two data elements are written.
3.  $t_{WTR}$  is referenced from the first positive CK edge after the last data-in pair.
4. A10 is LOW with the WRITE command (auto precharge is disabled).
5. DQS is required at T2 and T2n (nominal case) to register DM.

### WRITE to PRECHARGE

Data for any WRITE burst may be followed by a subsequent PRECHARGE command. To follow a WRITE without truncating the WRITE burst,  $t_{WR}$  should be met.

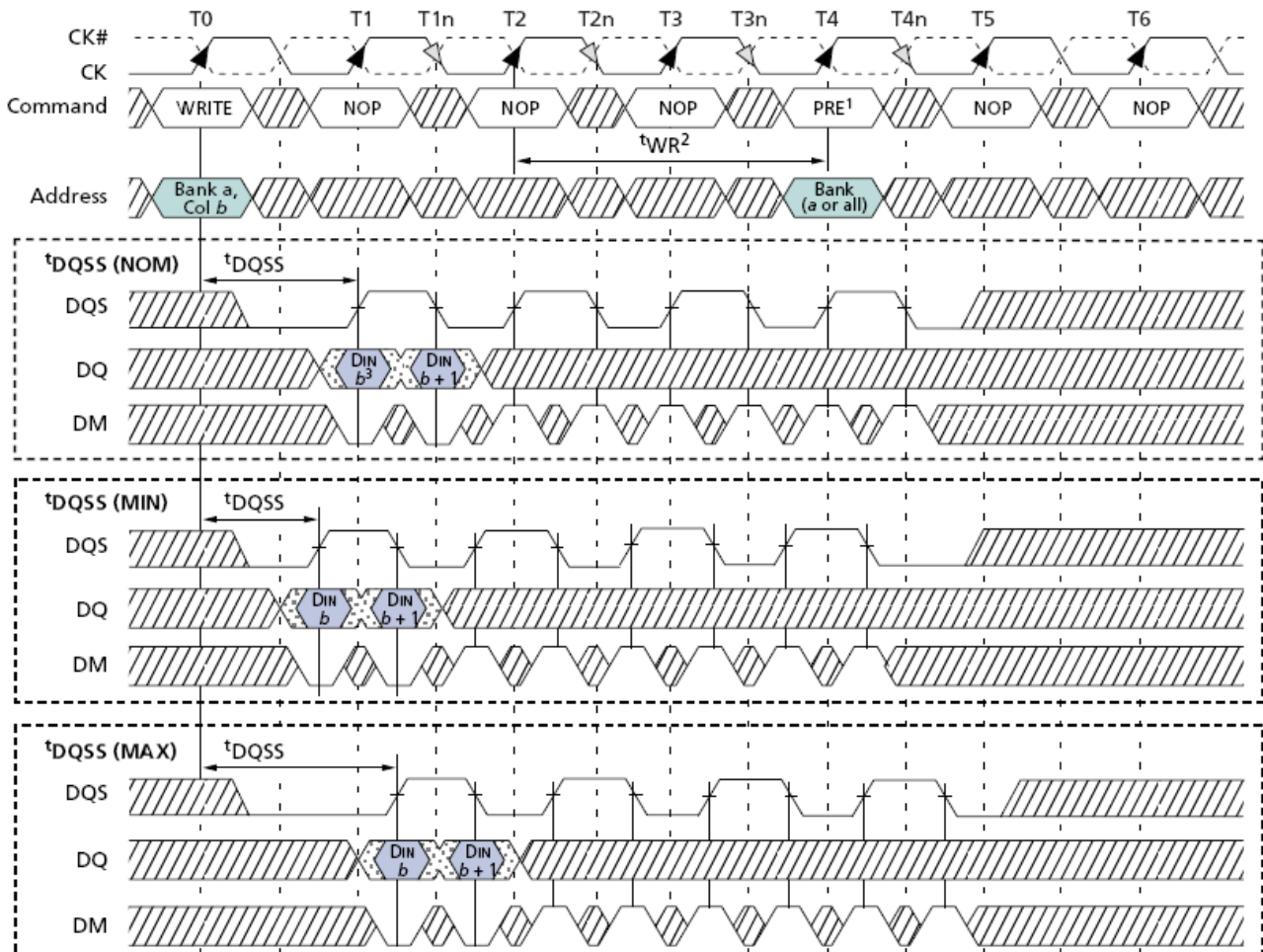


**Non-Interrupting Write-to-Precharge (nominal,  $t_{DQSS}(\min)/(\max)$ , BL=4)**

Notes:

1. PRE = PRECHARGE.
2. Din b = data-in from column b.
3. An uninterrupted burst of 4 is shown.
4. A10 is LOW with the WRITE command (auto precharge is disabled).
5.  $t_{WTR}$  is referenced from the first positive CK edge after the last data-in pair.
6. The PRECHARGE and WRITE commands are to the same device. However, the PRECHARGE and WRITE commands can be to different devices; in this case,  $t_{WR}$  is not required and the PRECHARGE command can be applied earlier.

Data for any Write burst may be truncated by a subsequent PRECHARGE command as shown in Figure. Note that the only data-in pairs that are registered prior to the  $t_{WR}$  period are written to the internal array, and any subsequent data-in must be masked with DM. After the PRECHARGE command, a subsequent command to the same bank can not be issued until  $t_{RP}$  is met.



**Interrupting Write-to-Precharge (nominal,  $t_{DQSS}(\min)/(\max)$ , BL=8)**

Notes:

1. PRE = PRECHARGE.
2.  $t_{WR}$  is referenced from the first positive CK edge after the last data-in pair.
3. Din b = data-in from column b.
4. An interrupted burst of 8 is shown; two data elements are written.
5. A10 is LOW with the WRITE command (auto precharge is disabled).
6. DQS is required at T4 and T4n to register DM.



### PRECHARGE

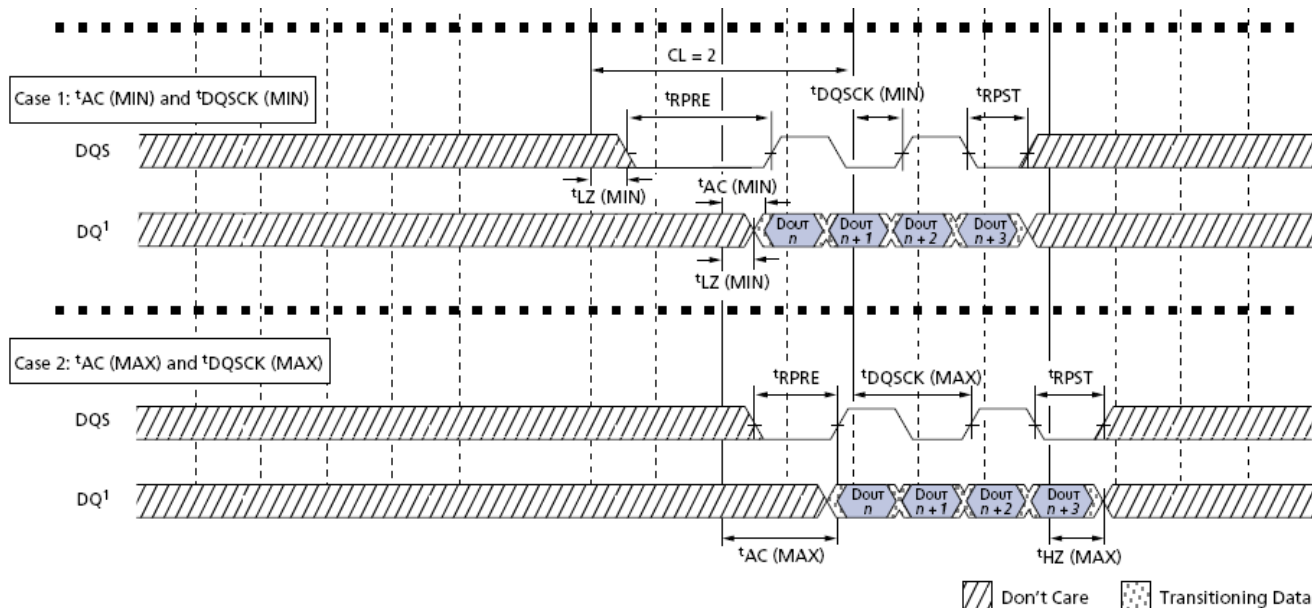
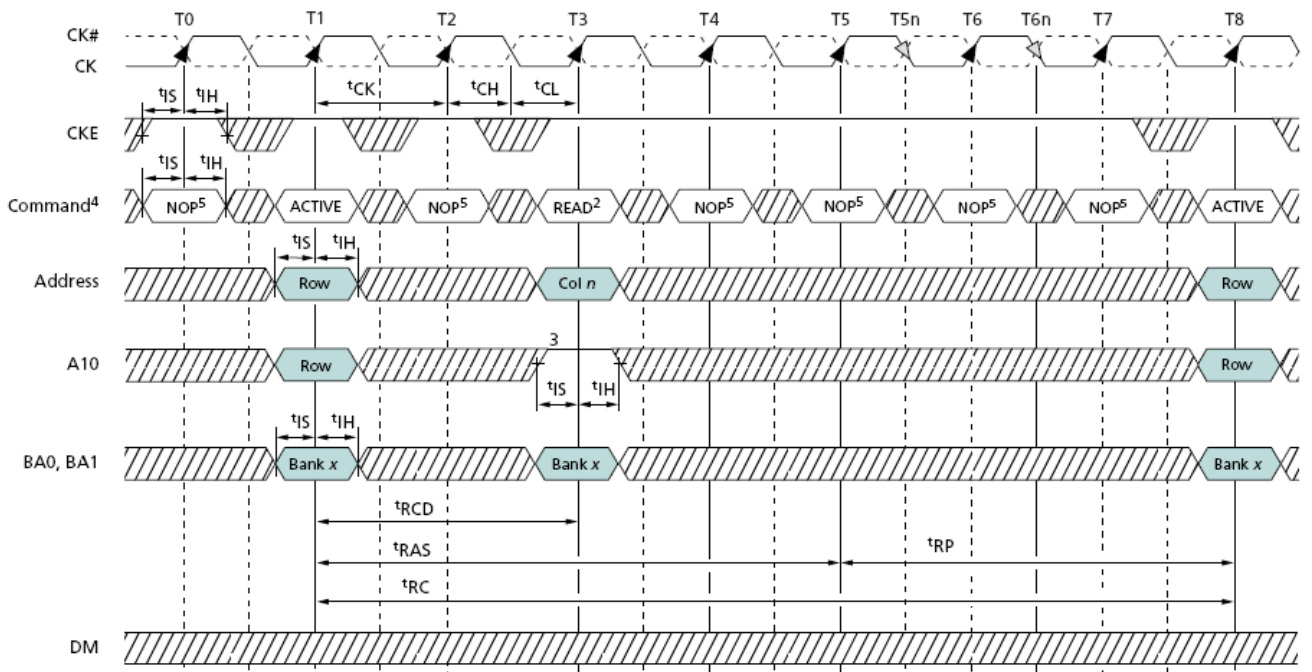
The PRECHARGE command is used to deactivate the open row in a particular bank or the open row in all banks. The bank(s) will be available for a subsequent row access some specified time ( $t_{RP}$ ) after the PRECHARGE command is issued. Input A10 determines whether one or all banks are to be precharged. In case where only one bank is to be precharged (A10=LOW), inputs BA0, BA1 select the bank. Otherwise BA0, BA1 are treated as “Don’t Care”. Once a bank has been precharged, it is in the idle state and must be activated prior to any READ or WRITE commands being issued to that bank. A PRECHARGE command will be treated as a NOP if there is no open row in that bank (idle state), or if the previously open row is already in the process of precharging.

### AUTO PRECHARGE

Auto Precharge is a feature which performs the same individual bank precharge function described previously, but without requiring an explicit command. This is accomplished by using A10 (A10=High), to enable Auto Precharge in conjunction with a specific READ or WRITE command. A precharge of the bank / row that is addressed with the READ or WRITE command is automatically performed upon completion of the read or write burst. Auto precharge is non-persistent in that it is either enabled or disabled for each individual READ or WRITE command. Auto precharge ensures that a precharge is initiated at the earliest valid stage within a burst. The “earliest valid stage” is determined as if an explicit PRECHARGE command was issued at the earliest possible time, without violating  $t_{RAS(min)}$ . The READ with auto precharge enabled or WRITE with auto precharge enabled states can each be broken into two parts: the access period and the precharge period. The access period starts with registration of the command and ends where the precharge period (or  $t_{RP}$ ) begins. For READ with auto precharge, the precharge period is defined as if the same burst was executed with auto precharge disabled and then followed with the earliest possible PRECHARGE command that still accesses all the data in the burst. For WRITE with auto precharge, the precharge period begins when  $t_{WR}$  ends, with  $t_{WR}$  measured as if auto precharge was disabled. In addition, during a WRITE with auto precharge, at least one clock is required during  $t_{WR}$  time. During the precharge period, the user must not issue another command to the same bank until  $t_{RP}$  is satisfied. This device supports  $t_{RAS}$  lock-out. In the case of a single READ with auto-precharge or a single WRITE with auto-precharge issued at  $t_{RCD(min)}$ , the internal precharge will be delayed until  $t_{RAS(min)}$  has been satisfied.

### Concurrent AUTO PRECHARGE

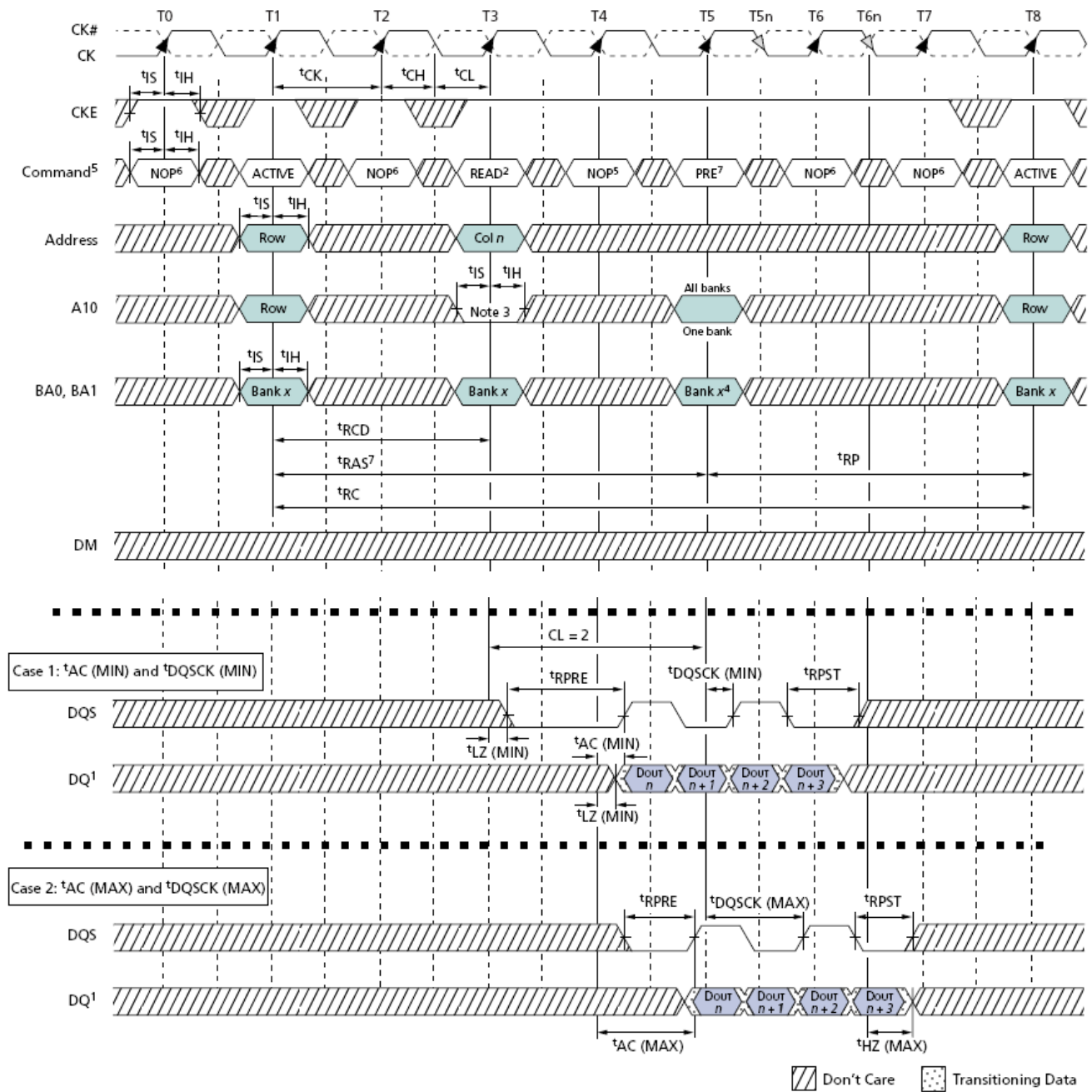
This device supports concurrent auto precharge such that when a READ with auto precharge is enabled or a WRITE with auto precharge is enabled, any command to another bank is supported, as long as that command does not interrupt the read or write data transfer already in process. This feature enables the precharge to complete in the bank in which the READ or WRITE with auto precharge was executed, without requiring an explicit PRECHARGE command, thus freeing the command bus for operations in other banks. During the access period of a READ or a WRITE with auto precharge, only ACTIVE and PRECHARGE commands may be applied to other banks. During the precharge period, ACTIVE, PRECHARGE, READ, and WRITE commands may be applied to other banks. In either situation, all other related limitations apply.



### Bank Read with Auto precharge ( $t_{AC}$ , $t_{DQSK}(\text{min})/(\text{max})$ , BL=4)

Notes:

1. Din n = data-out from column n.
2. BL = 4 in the case shown.
3. Enable auto precharge.
4. NOP commands are shown for ease of illustration; other commands may be valid at these times.

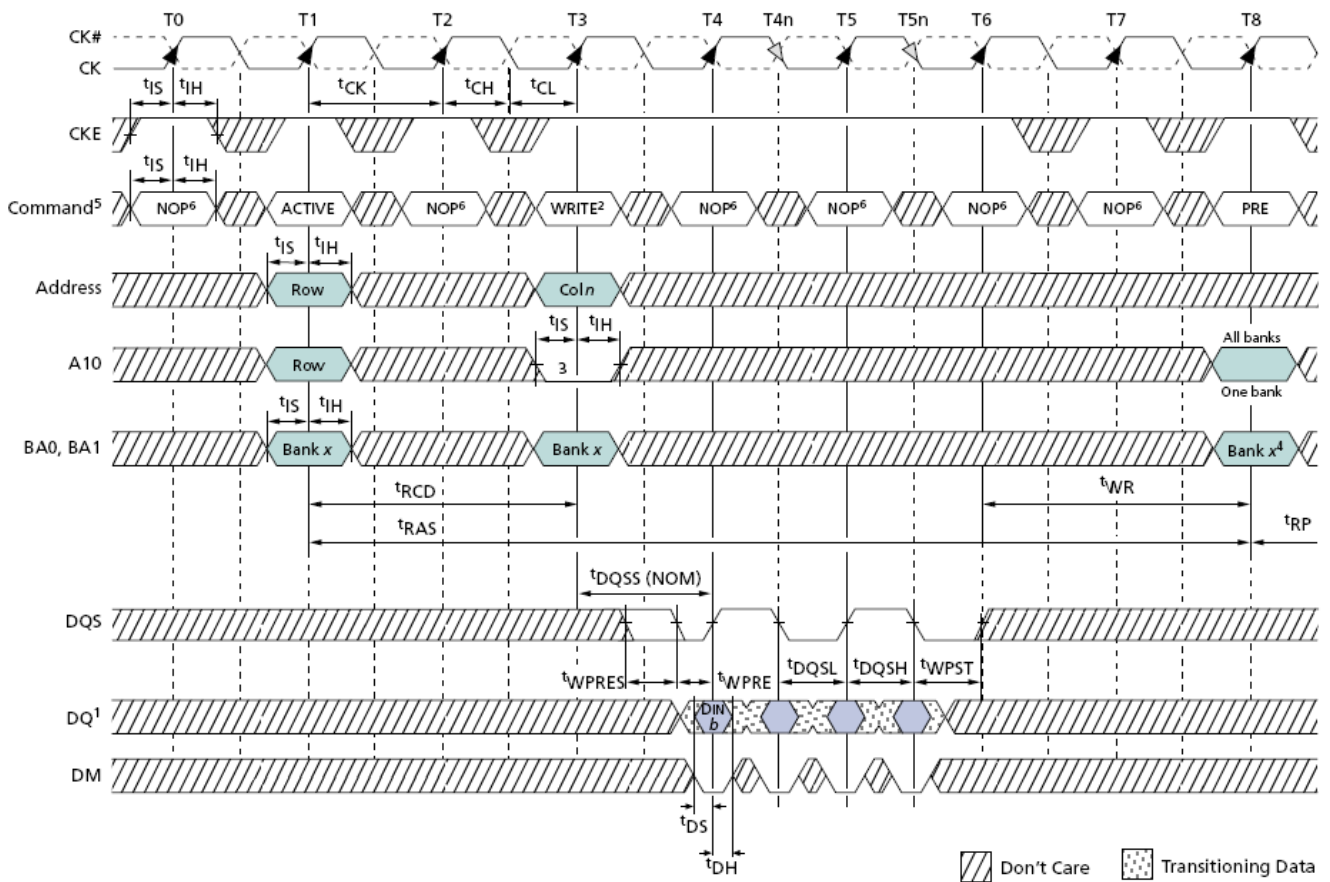


### Bank Read without Auto precharge ( $t_{AC}$ , $t_{DQSK}(\min)/(\max)$ , $BL=4$ )

Notes:

1. Din n = data-out from column n.
2. BL = 4 in the case shown.
3. Disable auto precharge.
4. BANK x at T5 is "Don't Care", if A10 is HIGH at T5.
5. PRE = PRECHARGE.
6. NOP commands are shown for ease of illustration; other commands may be valid at these times.
7. The PRECHARGE command can only be applied at T5, if  $t_{RAS}(\min)$  is met.

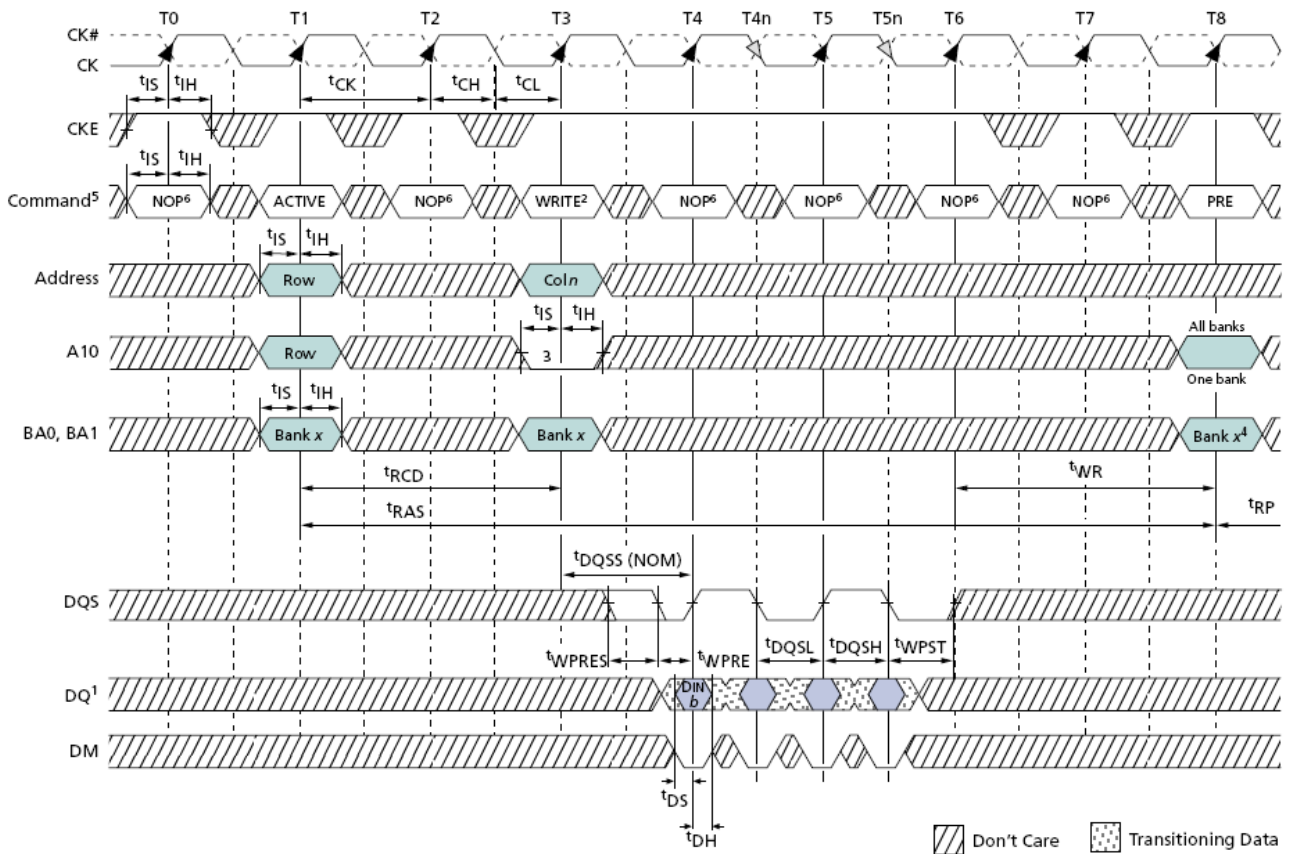




### Bank Write with Auto precharge (BL=4)

**Notes:**

1. Din n = data-out from column n.
2. BL = 4 in the case shown.
3. Enable auto precharge.
4. NOP commands are shown for ease of illustration; other commands may be valid at these times.



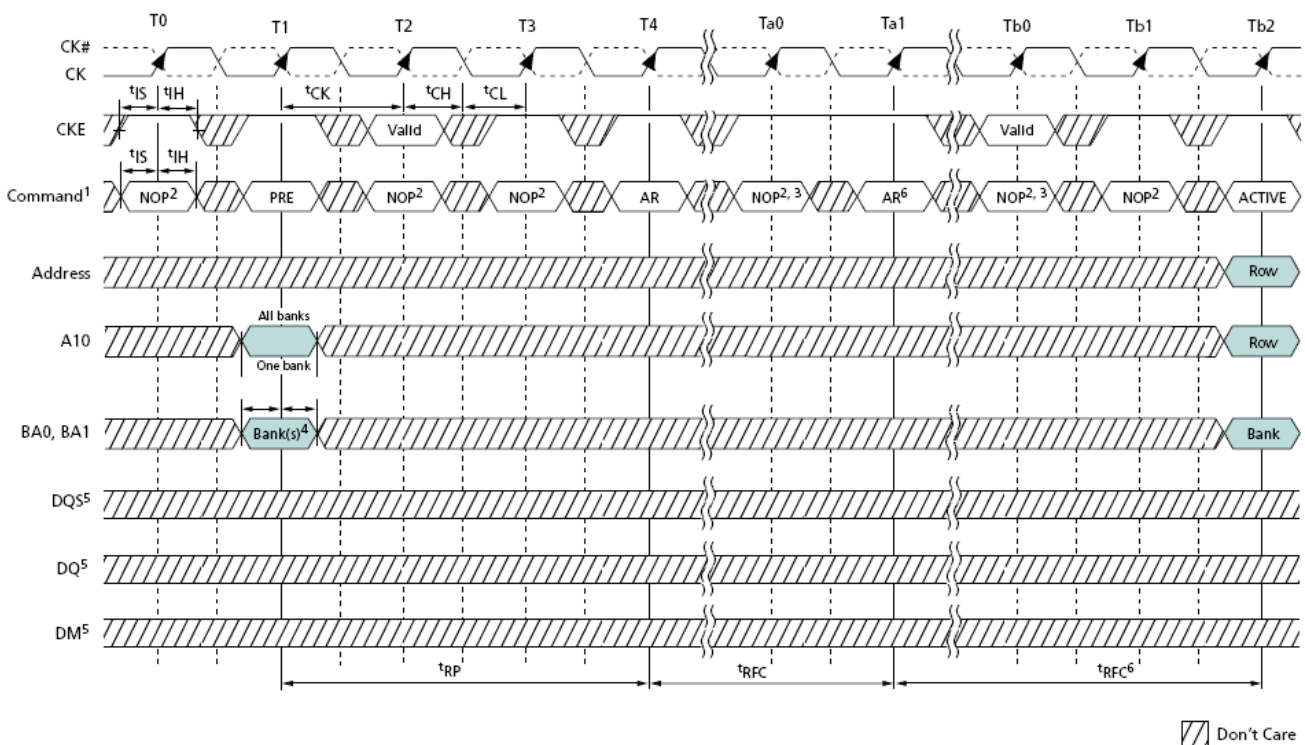
### Bank Write without Auto precharge (BL=4)

Notes:

1. Din n = data-out from column n.
2. BL =4 in the case shown.
3. Disable auto precharge.
4. Bank x at T8 is "Don't Care", if A10 is HIGH at T8.
5. PRE = PRECHARGE.
6. NOP commands are shown for ease of illustration; other commands may be valid at these times.

### AUTO REFRESH

AUTO REFRESH command is used during normal operation of the LPDDR SDRAM, and is analogous to  $\overline{\text{CAS-BEFORE-RAS}}$  (CBR) Refresh in the FPM/EDO DRAMs. The Auto Refresh is non-persistent, so it must be issued each time a refresh is required. The refresh addressing is generated by the internal refresh controller. The address bits become “Don’t Care” during AUTO REFRESH. The LPDDR SDRAM requires AUTO REFRESH commands at an average periodic interval of  $t_{\text{REFI}}$ . To provide improved efficiency in scheduling and switching between tasks, some flexibility in the absolute refresh interval is provided. Although it is not a JEDEC requirement, CKE must be active (HIGH) during the auto refresh period to provide support for future functional features. The auto refresh period begins when the AUTO REFRESH command is registered and ends  $t_{\text{RFC}}$  later.



**Notes:**

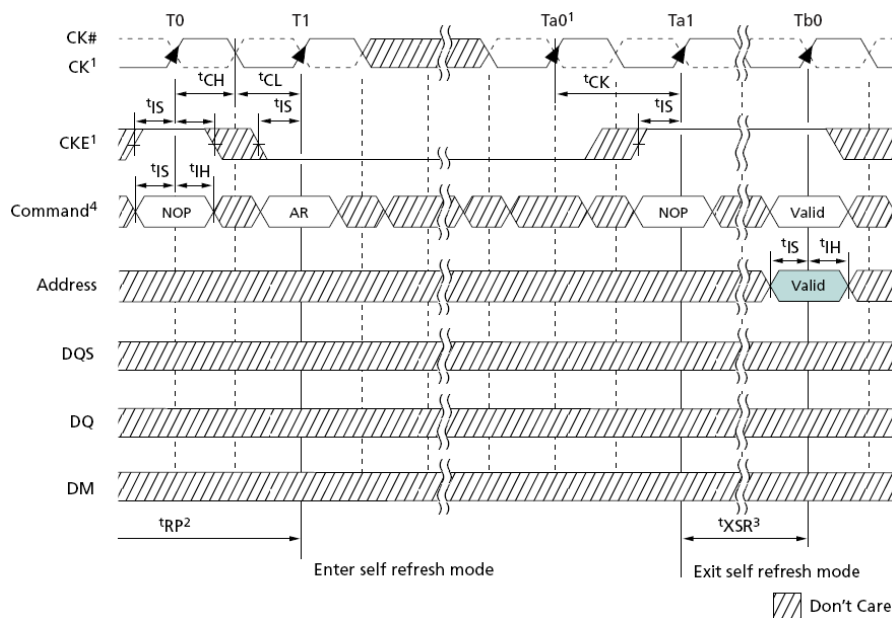
1. PRE = PRECHARGE; AR = AUTO REFRESH.
2. NOP commands are shown for ease of illustration; other commands may be valid at these times. CKE must be active during clock positive transitions.
3. NOP or COMMAND INHIBIT are the only commands supported until after  $t_{\text{RFC}}$  time; CKE must be active during clock positive transitions.
4. Bank x at T1 is “Don’t Care”, if A10 is HIGH at this point; A10 must be HIGH if more than one bank is active.
5. DM, DQ, and DQS signals are all “Don’t Care”, High-Z for operations shown.
6. The second AUTO PRECHARGE is not required and is only shown as an example of two back-to-back AUTO REFRESH commands.

### SELF REFRESH

SELF REFRESH command can be used to retain data in the LPDDR SDRAM, even if the rest of the system is powered down. When in the self refresh mode, the LPDDR SDRAM retains data without external clocking. The LPDDR SDRAM device has a built-in timer to accommodate Self Refresh operation. The SELF REFRESH command is initiated like an AUTO REFRESH command, except CKE is LOW. Input signals except CKE are “Don’t Care” during Self Refresh. During SELF REFRESH, the device is refreshed as identified in the extended mode register. Once the SELF REFRESH command is registered, the external clock can be halted after one clock later. CKE must be held low to keep the device in Self Refresh mode, and internal clock also disabled to save power. The minimum time that the device must remain in Self Refresh mode is  $t_{RFC}$ .

In the Self Refresh mode, two additional power-saving options exist: Temperature Compensated Self Refresh and Partial Array Self Refresh. During this mode, the device is refreshed as identified in the extended mode register. An internal temperature sensor will adjust the refresh rate to optimize device power consumption while ensuring data integrity. During SELF REFRESH operation, refresh intervals are scheduled internally and may vary. These refresh intervals may be different than the specified  $t_{REFI}$  time. For this reason, the SELF REFRESH command must not be used as a substitute for the AUTO REFRESH command.

The procedure for exiting SELF REFRESH requires a sequence of commands. First, CK must be stable prior to CKE going back HIGH. When CKE is HIGH, the LPDDR SDRAM must have NOP commands issued for  $t_{XSR}$  time to complete any internal refresh already in progress.

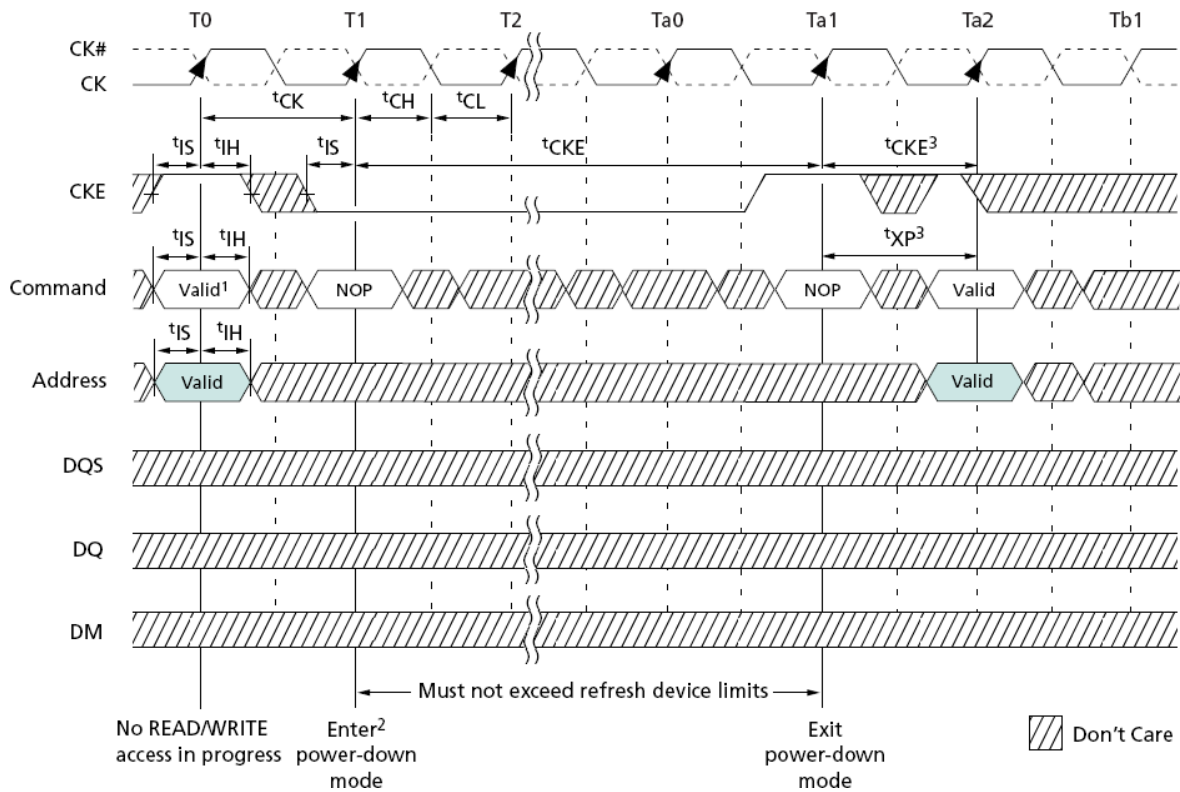


**Notes:**

1. Clock must be stable, cycling within specifications by  $Ta0$ , before exiting self refresh mode.
2. Device must be in the all banks idle state prior to entering self refresh mode.
3. NOPs or DESELECTs is required for  $t_{XSR}$  time with at least two clock pulses.
4. AR = AUTO REFRESH.
5. CKE must remain LOW to remain in self refresh.

**Power-Down**

Power-down is entered when CKE is registered Low (no accesses can be in progress). If power-down occurs when all banks are idle, this mode is referred to as precharge power-down; if power-down occurs when there is a row active in any bank, this mode is referred to as active power-down. Power-down mode deactivates all input and output buffers, excluding CK,  $\overline{CK}$  and CKE. CKE keep Low to maintain device in the power-down mode, and all other inputs signals are “Don’t Care”. The minimum power-down duration is specified by  $t_{CKE}$ . The device can not stay in this mode for longer than the refresh requirements of the device, without losing data. The power-down state is synchronously existed when CKE is registered High (along with a NOP or DESELECT command). A valid command can be issued after  $t_{XP}$  after exist from power-down.

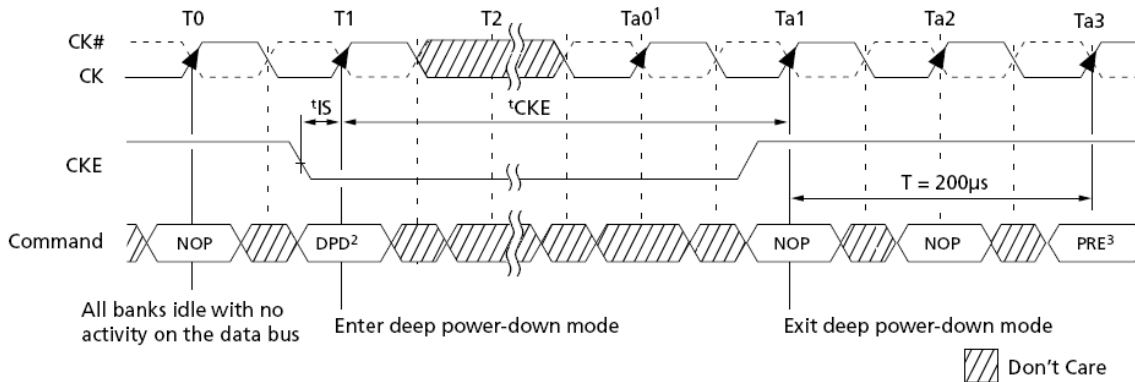


Notes:

1. If this command is a PRECHARGE (or if the device is already in the idle state), then the power-down mode shown is precharge power-down. If this command is an ACTIVE (or if at least one row is already active), then the power-down mode is active power-down.
2. No column accesses can be in progress, when power-down is entered.
3.  $t_{CKE}$  applies if CKE goes LOW at Ta2 (entering power-down);  $t_{XP}$  applies if CKE remains HIGH at Ta2 (exit power-down).

### Deep-Power-Down

The Deep Power-Down (DPD) mode is an operating mode used to achieve maximum power reduction by eliminating the power of the memory array. All internal voltage generators inside the LPDDR SDRAM are stopped and all memory data, MRS and EMRS information is lost in this mode. The DPD command is the same as a BURST TERMINATE command with CKE LOW. All banks must be in idle state with no activity on the data bus prior to entering the DPD mode. While in this mode, CKE must be held in a constant Low state. To exit the DPD mode, CKE is taken high after the clock is stable and NOP commands must be maintained for at least 200us. After 200us a complete re-initialization is required.



#### Notes:

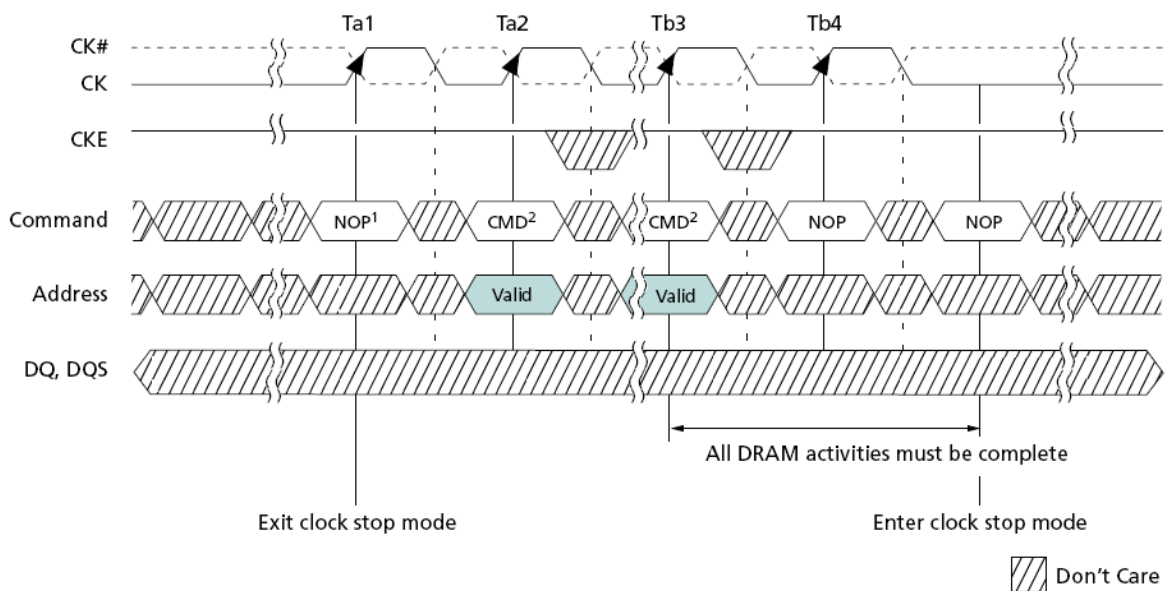
1. Clock must be stable prior to CKE going HIGH.
2. DPD = Deep Power-Down.
3. Upon exit of power-down mode, a full DRAM initialization sequence is required.

### Clock Stop

One method of controlling the power efficiency in applications is to throttle the clock that controls the LPDDR SDRAM. The clock may be controlled in two ways:

- Change the clock frequency.
- Stop the clock.

The LPDDR SDRAM enables the clock to change frequency during operation only if all the timing parameters are met, and all refresh requirements are satisfied. The clock can be stopped altogether if there are no DRAM operations in progress that would be affected by this change. Any DRAM operation already in process must be completed before entering clock stop mode; this includes the following timings:  $t_{RCD}$ ,  $t_{RP}$ ,  $t_{RFC}$ ,  $t_{MRD}$ ,  $t_{WR}$ , and  $t_{RPST}$ . In addition, any READ or WRITE burst in progress must complete. CKE must be held HIGH, with  $CK=LOW$  and  $\overline{CK}=HIGH$ , for the full duration of the clock stop mode. One clock cycle and at least one NOP or DESELECT is required after the clock is restarted before a valid command can be issued.



#### Notes:

1. Prior to  $Ta1$ , the device is in clock stop mode. To exit, at least one NOP is required before any valid command.
2. Any valid command is supported; device is not in clock suspend mode.

### Revision History

Rev	Page	Modified	Description	Released
0.1	-	Preliminary Release	1 <sup>st</sup> version released.	03/13
1.0	-	Official Release	Official version released.	05/13
1.1	P5	Ball assignment	Remove unused pin, A13, which is for 1G device above.	06/13
	P6	Ball assignment	1. A13's background is colored by light blue. 2. Add Note: A13 is only available for reduced page-size configuration.	
	P16-19	Electrical Characteristics and Recommended AC Operating Conditions	1. Add Notes: 1-9 apply to all of the parameters in this table	
1.2	All	-	1. Renew it. 2. VDD/VDDQ Spec change: 1.7 to 1.95V (was: 1.7 to 1.9V)	08/13
1.3	P1	Density & Addressing	Correct typo: 1. 333 Mbps / 3-3-3 (was: 333 Mbps / 2.5-3-3) 2. 110ns (was:78ns). Align the spec on page 18	09/13
	P14-16	IDD spec	1. Add IDD8 spec @ 85C. 2. IDD6 Spec: typical (was: Max)	
	P27	Initialization Sequence Diagram with temperature output signal	Add the information about temperature sensor's output signal	





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